CMOS 8-Bit Microcontroller

TMP87C814N/F, TMP87CH14N/F, TMP87CK14N/F, TMP87CM14N/F

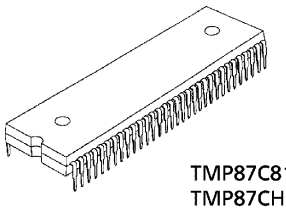
The 87C814/H14/K14/M14 are the high speed and high performance 8-bit single chip microcomputers. These MCU contain 8-bit A/D conversion inputs and a VFT (Vacuum Fluorescent Tube) driver on a chip.

Part No.	ROM	RAM	Package	OTP MCU
TMP87C814N/F	8 K × 8-bit	512 × 8-bit	P-SDIP64-750-1.78	TMP87PM14N/F
TMP87CH14N/F	16 K × 8-bit			
TMP87CK14N/F	24 K × 8-bit	1024 × 8-bit	P-QFP64-1420-1.00A	
TMP87CM14N/F	32 K × 8-bit			

Features

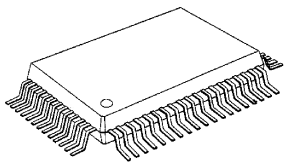
- ◆8-bit single chip microcomputer TLCS-870 Series
- ◆Instruction execution time: 0.5 μs (at 8 MHz), 122 μs (at 32.768 kHz)
- ◆412 basic instructions
  - Multiplication and Division (8 bits × 8 bits , 16 bits ÷ 8 bits)
  - Bit manipulations (Set/Clear/Complement/Move/Test/Exclusive or)
  - 16-bit data operations
  - 1-byte jump/subroutine-call (Short relative jump/ Vector call)
- ◆13 interrupt sources (External: 5, Internal: 8)
  - All sources have independent latches each, and nested interrupt control is available.
  - 3 edge-selectable external interrupts with noise reject
  - High-speed task switching by register bank changeover
- ◆8 Input/Output ports (55 pins)
  - Input/Output: 8 ports (55 pins)
- ◆Two 16-bit Timer/Counters
  - Timer, Event counter, Programmable pulse generator output, Pulse width measurement, External trigger timer, Window modes.
- ◆Two 8-bit Timer/Counters
  - Timer, Event counter, PWM output, Programmable divider output modes
- ◆Time Base Timer (Interrupt frequency: 1 Hz to 16 kHz)
- ◆Divider output function (frequency: 1 kHz to 8 kHz)
- ◆Watchdog Timer
  - Interrupt source/reset output (programmable)
- ◆8-bit Serial Interface: 1 channel
  - With 8 bytes transmit/receive data buffer
  - Internal/external serial clock, and 4/8-bit mode
- ◆8-bit successive approximate type A/D converter with sample and hold
  - 8 analog inputs
  - Conversion time: 23 μs at 8 MHz
- ◆D/A conversion (Pulse Width Modulation) output
  - 14-bit resolution (1 channel)

P-SDIP64-750-1.78



TMP87C814N
TMP87CH14N
TMP87CK14N
TMP87CM14N
TMP87PM14N

P-QFP64-1420-1.00A



TMP87C814F
TMP87CH14F
TMP87CK14F
TMP87CM14F
TMP87PM14F

980910EBP1

- For a discussion of how the reliability of microcontrollers can be predicted, please refer to Section 1.3 of the chapter entitled Quality and Reliability Assurance / Handling Precautions.
- TOSHIBA is continually working to improve the quality and the reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to observe standards of safety, and to avoid situations in which a malfunction or failure of a TOSHIBA product could cause loss of human life, bodily injury or damage to property. In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent products specifications. Also, please keep in mind the precautions and conditions set forth in the TOSHIBA Semiconductor Reliability Handbook.
- The products described in this document are subject to the foreign exchange and foreign trade laws.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA CORPORATION for any infringements of intellectual property or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any intellectual property or other rights of TOSHIBA CORPORATION or others.
- The information contained herein is subject to change without notice.

◆Vacuum Fluorescent Tube Driver (automatic display)
- High breakdown voltage ports (max. 40 V × 24 bits)

◆Dual clock operation
- Single/Dual-clock mode (option)

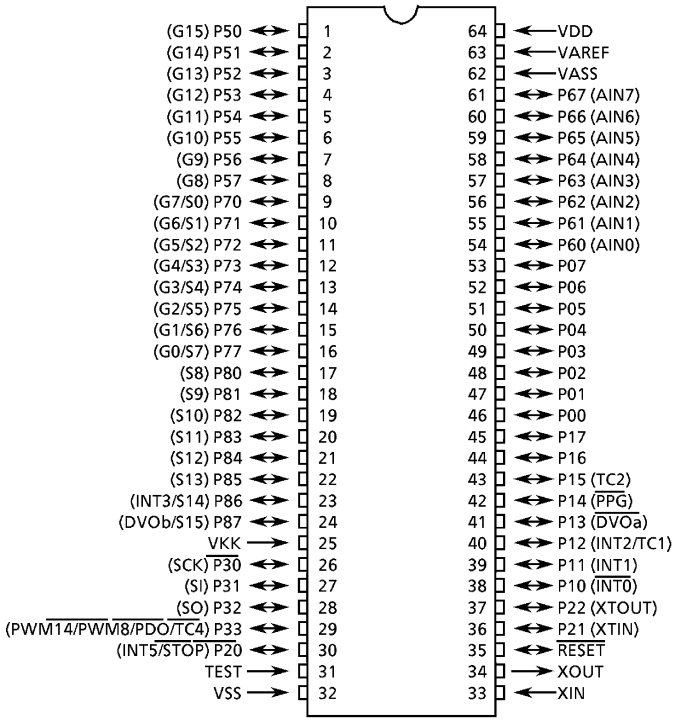
◆Five Power saving operating modes
- STOP mode: Oscillation stops. Battery / Capacitor back-up. Port output hold/High-impedance.
- SLOW mode: Low power consumption operation using low-frequency clock (32.768 kHz).
- IDLE1 mode: CPU stops, and Peripherals operate using high-frequency clock. Release by interrupts.
- IDLE2 mode: CPU stops, and Peripherals operate using high-and low-frequency clock. Release by interrupts.
- SLEEP mode: CPU stops, and Peripherals operate using low-frequency clock. Release by interrupts.

◆Wide operating voltage: 2.7 to 5.5 V at 32.768 kHz, 4.5 to 5.5 V at 8 MHz / 32.768 kHz

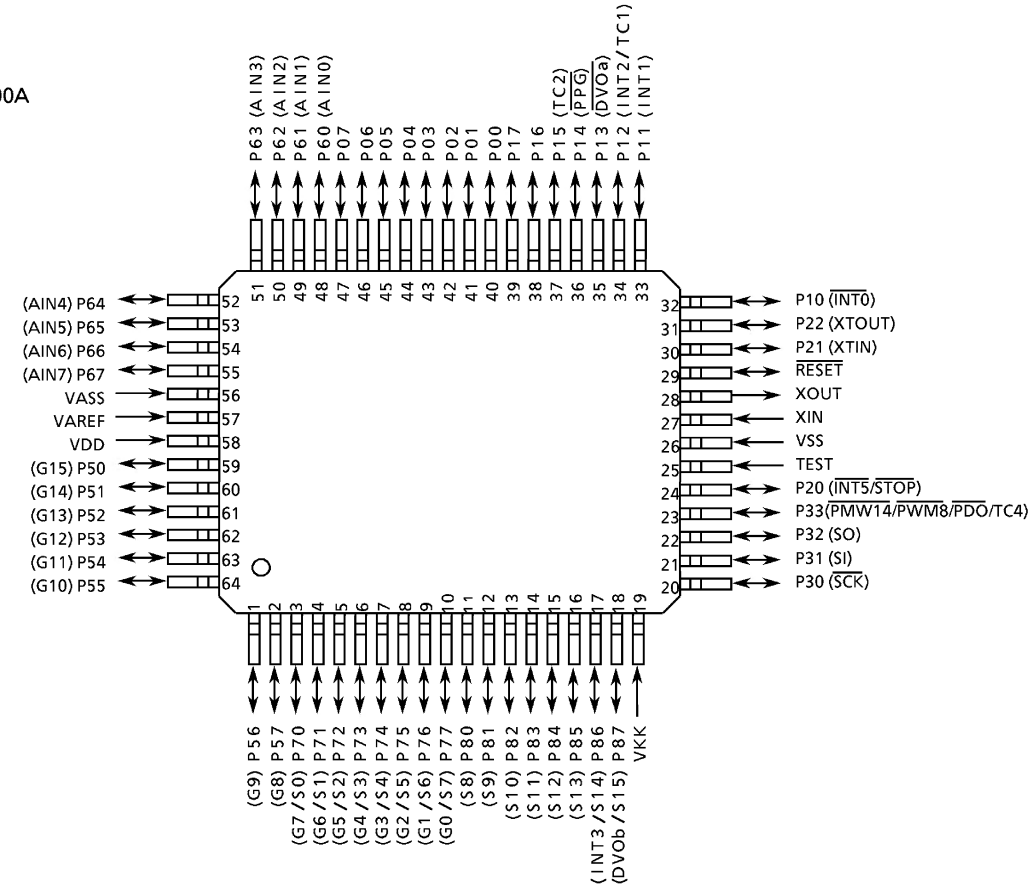
◆Emulation Pod: BM87CM14N0A

Pin Assignments (Top View)

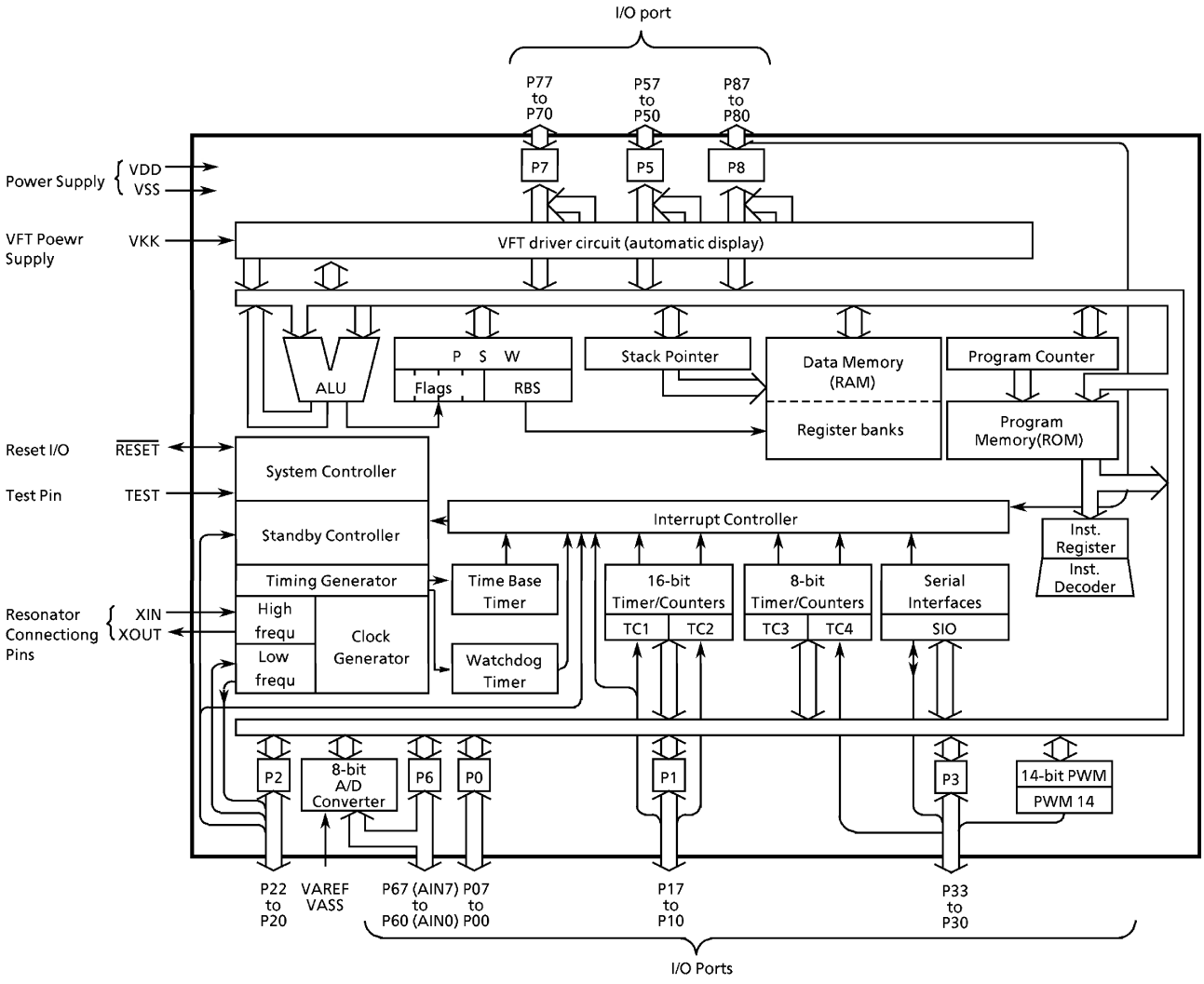
(1) P-SDIP64-750-1.78



(2) P-QFP64-1420-1.00A



Block Diagram



Pin Function

Pin Name	Input / Output	Function	
P07 to P00	I/O	Two 8-bit programmable input/output ports (tri-state). Each bit of these ports can be individually configured as an input or an output under software control. During reset, all bits are configured as inputs. When used as a divider output, the latch must be set to "1".	
P17, P16	I/O		
P15 (TC2)	I/O (Input)		Timer/Counter 2 input
P14 ($\overline{PPG}$)	I/O (Output)		Programmable pulse generator output
P13 ($\overline{DVOa}$)	I/O (Output)		Divider output a
P12 (INT2 / TC1)	I/O (Input)		External interrupt input 2 or Timer/Counter 1 input
P11 (INT1)			External interrupt input 1
P10 ($\overline{INT0}$)			External interrupt input 0
P22 (XTOUT)	I/O (Output)	3-bit input/output port with latch. When used as an input port, the latch must be set to "1".	Resonator connecting pins (32.768 kHz).
P21 (XTIN)	I/O (Input)		For inputting external clock, XTIN is used and XTOUT is opened.
P20 ($\overline{INT5}$/$\overline{STOP}$)			External interrupt input 5 or STOP mode release signal input
P33 ($\overline{PWM14}$ / $\overline{PWM8}$ / $\overline{PDO}$ / TC4)	I/O (I/O)	4-bit input/output port with latch. When used as an input port, a SIO input/output, a timer/counter input, or an interrupt input, the latch must be set to "1".	14-bit PWM output or 8-bit PWM output or 8-bit programmable divider output or Timer/Counter 4 input
P32 (SO)	I/O (Output)		SIO serial data Output
P31 (SI)	I/O (Input)		SIO serial data Input
P30 ($\overline{SCK}$)	I/O (I/O)		SIO serial clock input/output
P57 (G8) to P50 (G15)	I/O (Output)	8-bit high breakdown voltage linput/output ports with the latch. When used as a VFT driver output, the latch must be cleared to"0".	VFT digit driver outputs
P67 (AIN7) to P60 (AIN0)	I/O (Input)	8-bit programmable input/output port (tri-state).Each bit of the port can be individually configured as an input or an output under software control.	A/D converter analog inputs
P77 (S7/G0) to P70 (S0 / G7)	I/O (Output)	Two 8-bit high breakdown voltage output ports with the latch. When used as a VFT driver output, the latch must be cleared to "0".	VFT digit/segment driver outputs
P87 (DVOb / S15)	I/O (Output)	Two 8-bit high breakdown voltage output ports with the latch. When used as a VFT driver output, the latch must be cleared to "0".	Divider out put b or VFT segment driver output
P86 (INT3 / S14)	I/O (I/O)		External interrupt input 3 or VFT segment driver output
P85 (S13) to P80 (S8)	I/O (Output)		VFT segment output
XIN, XOUT	Input, Output	Resonator connecting pins for high-frequency clock. For inputting external clock, XIN is used and XOUT is opened.	
$\overline{RESET}$	I/O	Reset signal input or wachdog timer output/address-trap-reset output/system-clock-reset putput.	
TEST	Input	Test pin for out-going test. Be tied to low	
VDD, VSS	Power Supply	+ 5 V, 0 V (GND)	
VKK		VFT driver power supply	
VAREF, VASS		Analog reference voltage inputs (High, Low)	

OPERATIONAL DESCRIPTION

1. CPU CORE FUNCTIONS

The CPU core consists of a CPU, a system clock controller, an interrupt controller, and a watchdog timer. This section provides a description of the CPU core, the program memory (ROM), the data memory (RAM), and the reset circuit.

1.1 Memory Address Map

The TLCS-870 Series is capable of addressing 64K bytes of memory. Figure 1-1 shows the memory address maps of the 87C814/H14/K14/M14. In the TLCS-870 Series, the memory is organized 4 address spaces (ROM, RAM, SFR, and DBR). It uses a memory mapped I/O system, and all I/O registers are mapped in the SFR/DBR address spaces. There are 16 banks of general-purpose registers. The register banks are also assigned to the first 128 bytes of the RAM address space.

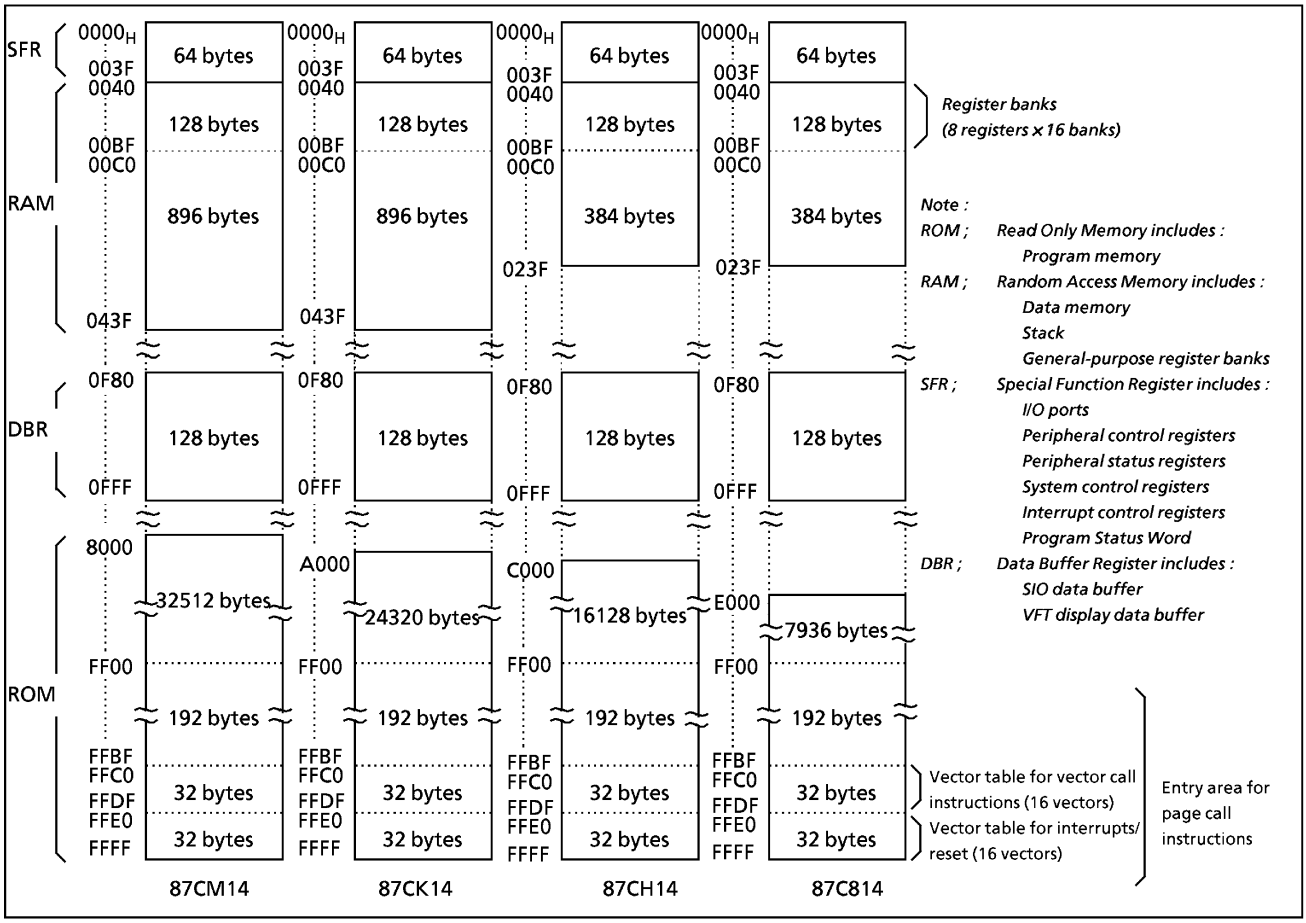


Figure 1-1. Memory Address Maps

1.2 Program Memory (ROM)

The 87C814 has a 8K × 8-bit (addresses $E000_H$-$FFFF_H$), the 87CH14 has a 16K × 8-bit (addresses $C000_H$-$FFFF_H$), the 87CK14 has a 24K × 8-bit (addresses $A000_H$-$FFFF_H$), and the 87CM14 has a 32K × 8-bit (address 8000_H-$FFFF_H$) of program memory (mask programmed ROM).
Addresses $FF00_H$-$FFFF_H$ in the program memory can also be used for special purposes.

(1) **Interrupt / Reset** vector table (addresses $FFE0_H$-$FFFF_H$)
This table consists of a reset vector and 15 interrupt vectors (2 bytes/vector). These vectors store a reset start address and interrupt service routine entry addresses.

(2) Vector table for **vector call** instructions (addresses $FFC0_H$-$FFDF_H$)
This table stores call vectors (subroutine entry address, 2 bytes/vector) for the vector call instructions [CALLV n]. There are 16 vectors. The CALLV instruction increases memory efficiency when utilized for frequently used subroutine calls (called from 3 or more locations).

(3) Entry area (addresses $FF00_H$-$FFFF_H$) for **page call** instructions
This is the subroutine entry address area for the page call instructions [CALLP n]. Addresses $FF00_H$-$FFBF_H$ are normally used because address $FFC0_H$-$FFFF_H$ are used for the vector tables.

Programs and fixed data are stored in the program memory. The instruction to be executed next is read from the address indicated by the current contents of the program counter (PC). There are relative jump and absolute jump instructions. The concepts of page or bank boundaries are not used in the program memory concerning any jump instruction.

Example: The relationship between the jump instructions and the PC.

① 5-bit PC-relative jump [JRS cc, $ + 2 + d]
E8C4H: JRS T, $ + 2 + 08H
When JF = 1, the jump is made to $E8CE_H$, which is 08_H added to the contents of the PC. (The PC contains the address of the instruction being executed + 2; therefore, in this case, the PC contents are $E8C4_H + 2 = E8C6_H$.)

② 8-bit PC-relative jump [JR cc, $ + 2 + d]
E8C4H : JR Z, $ + 2 + 80H
When ZF = 1, the jump is made to $E846_H$, which is $FF80_H$ (– 128) added to the current contents of the PC.

③ 16-bit absolute jump [JP a]
E8C4H : JP 0C235H
An unconditional jump is made to address $C235_H$. The absolute jump instruction can jump anywhere within the entire 64K-byte space.

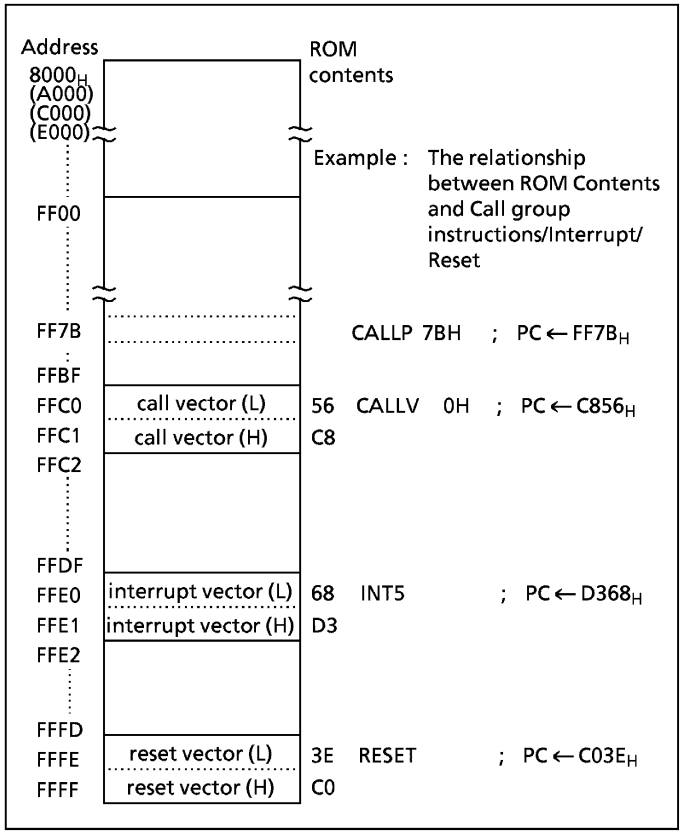


Figure 1-2. Program Memory Map

In the TLCS-870 Series, the same instruction used to access the data memory (e. g. [LD A, (HL)]) is also used to read out fixed data (ROM data) stored in the program memory. The register-offset-PC-relative addressing (PC + A) instructions can also be used, and the code conversion, table look-up and n-way multiple-direction jump processing can easily be programmed.

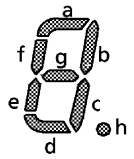
Example 1 : Loads the ROM contents at the address specified by the HL register pair contents into the accumulator (87CK14 : HL≧A000$_H$):

```
LD      A, (HL)          ; A←ROM (HL)
```

Example 2 : Converts BCD to 7-segment code (common anode LED). When A = 05$_H$, 92$_H$ is output to port P3 after executing the following program:

```
        ADD     A, TABLE – $ – 4     ; P3 ←ROM (TABLE + A)
        LD      (P3), (PC + A)
        JRS     T, SNEXT
TABLE : DB      0C0H, 0F9H, 0A4H, 0B0H, 99H, 92H, 82H, 0D8H, 80H, 98H
SNEXT :
```

Notes : "$" is a header address of ADD instruction.
DB is a byte data difinition instruction.



Example 3 : N-way multiple jump in accordance with the contents of accumulator (0≦A≦3):

```
        SHLC    A            ; if A = 00H then PC←C234H
        JP      (PC + A)       if A = 01H then PC←C378H
                               if A = 02H then PC←DA37H
                               if A = 03H then PC←E1B0H
        DW      0C234H, 0C378H, 0DA37H, 0E1B0H
```

Note : DW is a word data definition instruction.

SHLC A
JP (PC + A)
34
C2
78
C3
37
DA
B0
E1

1.3 Program Counter (PC)

The program counter (PC) is a 16-bit register which indicates the program memory address where the instruction to be executed next is stored. After reset, the user defined reset vector stored in the vector table (addresses FFFF$_H$ and FFFE$_H$) is loaded into the PC ; therefore, program execution is possible from any desired address. For example, when C0$_H$ and 3E$_H$ are stored at addresses FFFF$_H$ and FFFE$_H$, respectively, the execution starts from address C03E$_H$ after reset.

The TLCS-870 Series utilizes pipelined processing (instruction pre-fetch); therefore, the PC always indicates 2 addresses in advance. For example, while a 1-byte instruction stored at address C123$_H$ is being executed, the PC contains C125$_H$.

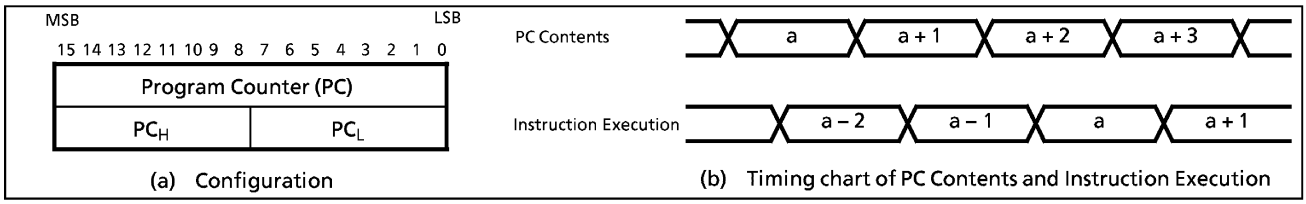


Figure 1-3. Program Counter

1.4 Data Memory (RAM)

The 87C814/H14 have a 512 × 8-bits (addresses 0040$_H$ to 023F$_H$), and the 87CK14/CM14 have a 1K × 8bit (address 0040$_H$ to 043F$_H$) of data memory (static RAM). Figure 1-4 shows the data memory map.

Addresses 0000$_H$-00FF$_H$ are used as a direct addressing area to enhance instructions which utilize this addressing mode; therefore, addresses 0040$_H$-00FF$_H$ in the data memory can also be used for user flags or user counters. General-purpose register banks (8 registers × 16 banks) are also assigned to the 128 bytes of addresses 0040$_H$-00BF$_H$. Access as data memory is still possible even when being used for registers. For example, when the contents of the data memory at address 0040$_H$ is read out, the contents of the accumulator in the bank 0 are also read out. The stack can be located anywhere within the data memory except the register bank area. The stack depth is limited only by the free data memory size. For more details on the stack, see section "1.7 Stack and Stack Pointer".

The 87C814/H14/K14/M14 cannot execute programs placed in the data memory. When the program counter indicates a data memory address, a bus error occurs and an address-trap-reset applies. The $\overline{\text{RESET}}$ pin goes low during the address-trap-reset.

Example 1 : If bit 2 at data memory address $00C0_H$ is "1", 00_H is written to data memory at address $00E3_H$; otherwise, FF_H is written to the data memory at address $00E3_H$:

```
        TEST    (00C0H).2       ;  if (00C0H) 2 = 0 then jump
        JRS     T,SZERO
        CLR     (00E3H)         ;  (00E3H) ← 00H
        JRS     T,SNEXT
SZERO : LD      (00E3H), 0FFH   ;  (00E3H) ← FFH
SNEXT :
```

Example 2 : Increments the contents of data memory at address $00F5_H$, and clears to 00_H when 10_H is exceeded:

```
        INC     (00F5H)         ;  (00F5H) ← (00F5H) + 1
        AND     (00F5H), 0FH    ;  (00F5H) ← (00F5H)∧0FH
```

The data memory contents become unstable when the power supply is turned on; therefore, the data memory should be initialized by an initialization routine. Note that the general-purpuse registers are mapped in the RAM ; therefore, *do not clear RAM at the current bank addresses*.

Example1 : Clears RAM to "00_H" except the bank 0: (87C814/CH14)

```
        LD      HL, 0048H       ;  Sets start address to HL register pair
        LD      A, H            ;  Sets initial data (00H) to A register
        LD      BC, 01F7H       ;  Sets number of byte to BC register pair
SRAMCLR : LD    (HL+), A
        DEC     BC
        JRS     F, SRAMCLR
```

Example2 : Clears RAM to "00_H" except the bank 0: (87CK14/CM14)

```
        LD      HL, 0048H       ;  Sets start address to HL register pair
        LD      A, H            ;  Sets initial data (00H) to A register
        LD      BC, 03F7H       ;  Sets number of byte to BC register pair
SRAMCLR : LD    (HL+), A
        DEC     BC
        JRS     F, SRAMCLR
```

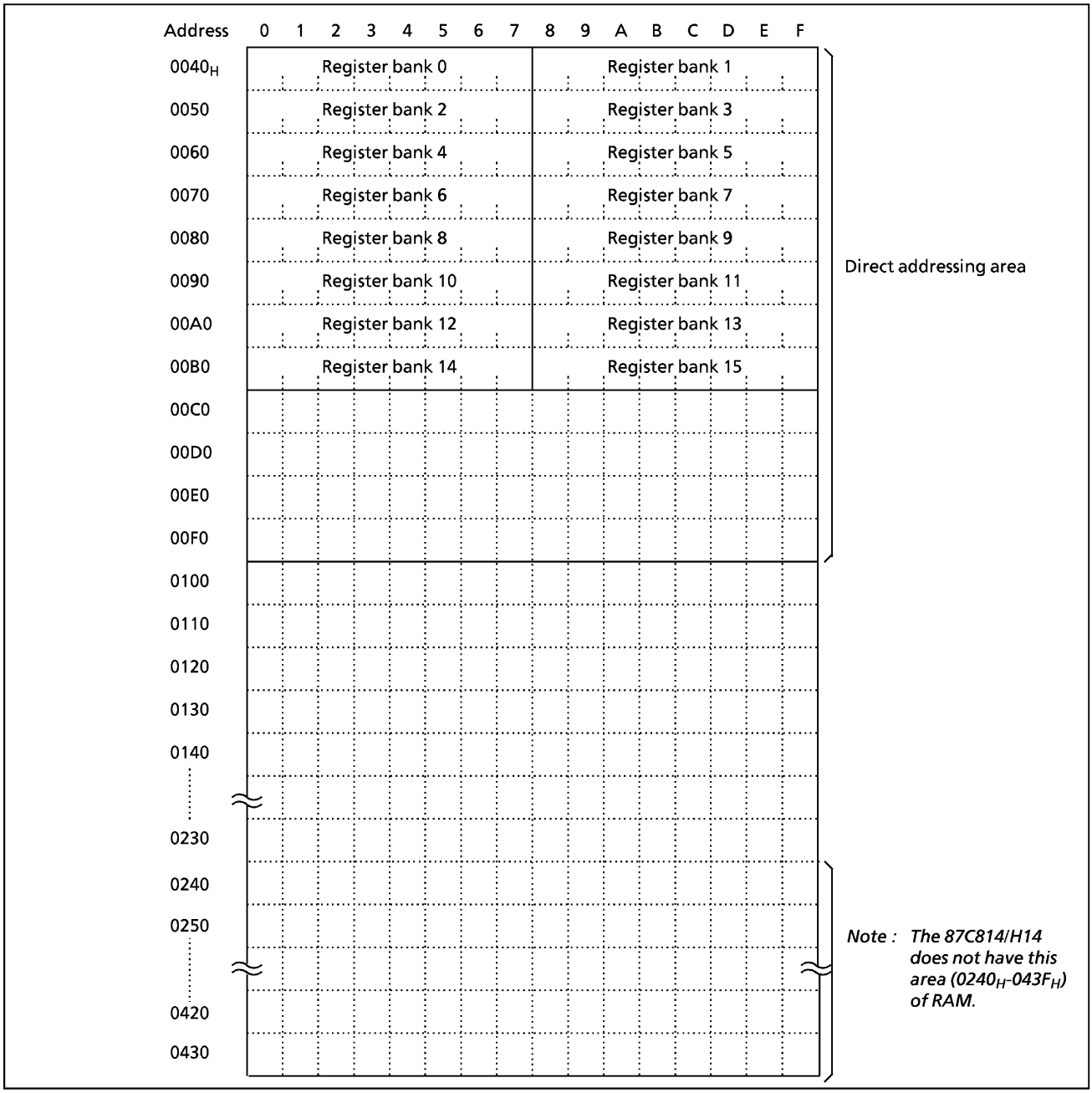


Figure 1-4. Data Memory Map

1.5 General-purpose Register Banks

The general-purpose registers are mapped into addresses 0040_H-$00BF_H$ in the data memory as shown in Figure 1-4. There are 16 register banks, and each bank contains eight 8-bit registers W, A, B, C, D, E, H, and L. Figure 1-5 shows the general-purpose register bank configuration.

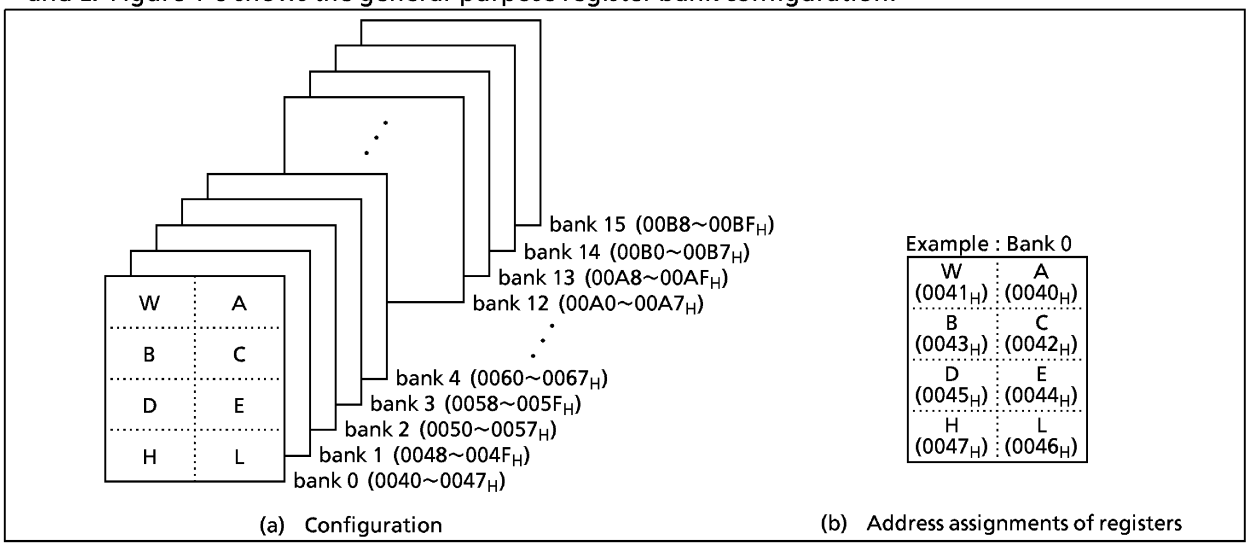


Figure 1-5. General-purpose Register Banks

In addition to access in 8-bit units, the registers can also be accessed in 16-bit units as the register pairs WA, BC, DE, and HL. Besides its function as a general-purpose register, the register also has the following functions:

(1) **A, WA**

The A register functions as an 8-bit accumulator and WA the register pair functions as a 16-bit accumulator (W is high byte and A is low byte). Registers other than A can also be used as accumulators for 8-bit operations.

```
Examples :  ①   ADD  A, B        ; Adds B contents to A contents and stores the result into A.
            ②   SUB  WA, 1234H   ; Subtracts 1234H from WA contents and stores the result into WA.
            ③   SUB  E, A        ; Subtracts A contents from E contents, and stores the result into E.
```

(2) **HL, DE**

The HL and DE specify a memory address. The HL register pair functions as data pointer (HL) / index register (HL + d) / base register (HL + C), and the DE register pair function as a data pointer (DE). The HL also has an auto-post- increment and auto-pre-decrement functions. This function simplifies multiple digit data processing, software LIFO (last-in first-out) processing, etc.

```
Example 1 :  ①   LD   A, (HL)       ; Loads the memory contents at the address specified by HL into A.
             ②   LD   A, (HL + 52H)  ; Loads the memory contents at the address specified by the value
                                     obtained by adding 52H to HL contents into A.
             ③   LD   A, (HL + C)    ; Loads the memory contents at the address specified by the value
                                     obtained by adding the register C contents to HL contents into A.
             ④   LD   A, (HL + )     ; Loads the memory contents at the address specified by HL into A.
                                     Then increments HL.
             ⑤   LD   A, ( – HL)     ; Decrements HL. Then loads the memory contents at the address
                                     specified by new HL into A.
```

The TLCS-870 Series can transfer data directly memory to memory, and operate directly between memory data and memory data. This facilitates the programming of block processing.

Example 2 : Block transfer

```
        LD    B, m          ; m = n – 1 (n : Nunber of bytes to transfer)
        LD    HL, DSTA      ; Sets destination address to HL
        LD    DE, SRCA      ; Sets source address to DE
SLOOP : LD    (HL), (DE)    ; (HL) ← (DE)
        INC   HL
        INC   DE
        DEC   B
        JRS   F, SLOOP
```

(3) **B, C, BC**

Registers B and C can be used as 8-bit buffers or counters, and the BC register pair can be used as a 16-bit buffer or counter. The C register functions as an offset register for register-offset index addressing (refer to example 1 ③ above) and as a divisor register for the division instruction [DIV gg, C].

Example 1 : Repeat processing

```
        LD        B, n        ; Sets n as the number of repetitions to B
SREPEAT : processing            (n + 1 times processing)
        DEC       B
        JRS       F, SREPEAT
```

Example 2 : Unsigned integer division (16-bit ÷ 8-bit)

```
        DIV       WA, C       ; Divides the WA contents by the C contents, places the
                                quotient in A and the remainder in W.
```

The general-purpose register banks are selected by the 4-bit register bank selector (RBS). During reset, the RBS is initialized to "0". The bank selected by the RBS is called the current bank.
Together with the flag, the RBS is assigned to address $003F_H$ in the SFR as the program status word (PSW). There are 3 instructions [LD RBS, n], [PUSH PSW] and [POP PSW] to access the PSW. The PSW can be also operated by the memory access instruction.

Example 1 : Incrementing the RBS

```
        INC       (003FH)     ; RBS ← RBS + 1
```

Example 2 : Reading the RBS

```
        LD        A, (003FH)  ; A ← PSW (A3-0 ← RBS, A7-4 ← Flags)
```

Highly efficient programming and high-speed task switching are possible by using bank changeover to save registers during interrupt and to transfer parameters during subroutine processing.
During interrupt, the PSW is automatically saved onto the stack. The bank used before the interrupt was accepted is restored automatically by executing an interrupt return instruction [RETI]/[RETN] ; therefore, there is no need for the RBS save/restore software processing.
The TLCS-870 Series supports a maximum of 15 interrupt sources. One bank is assigned to the main program, and one bank can be assigned to each source. Also, to increase the efficiency of data memory usage, assign the same bank to interrupt sources which are not nested.

Example: Saving /restoring registers during interrupt task using bank changeover.

```
PINT1 : LD        RBS, n      ; RBS ← n (Bank changeover)
        Interrupt processing
        RETI                  ; Maskable interrupt return (Bank restoring)
```

1.6 Program Status Word (PSW)

The program status word (PSW) consists of a register bank selector (RBS) and four flags, and the PSW is assigned to address $003F_H$ in the SFR.

The RBS can be read and written using the memory access instruction (e. g. [LD A, (003FH)], [LD (003FH), A], however the flags can only be read. When writing to the PSW, the change specified by the instruction is made without writing data to the flags. For example, when the instruction [LD (003FH), 05H] is executed, "5" is written to the RBS and the JF is set to "1", but the other flags are not affected.

[PUSH PSW] and [POP PSW] are the PSW access instructions.

1.6.1 Register Bank Selector (RBS)

The register bank selector (RBS) is a 4-bit register used to select general-purpose register banks. For example, when RBS = 2, bank 2 is currently selected. During reset, the RBS is initialized to "0".

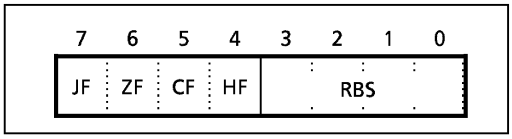


Figure 1-6. PSW (Flags, RBS) Configuration

1.6.2 Flags

The flags are configured with the upper 4 bits : a zero flag, a carry flag, a half carry flag and a jump status flag. The flags are set or cleared under conditions specified by the instruction. These flags except the half carry flag are used as jump condition "cc" for conditional jump instructions [JR cc, $ + 2 + d]/[JRS cc, $ + 2 + d]. After reset, the jump status flag is initialized to "1", other flags are not affected.

(1) **Zero flag (ZF)**

The ZF is set to "1" if the operation result or the transfer data is 00_H (for 8-bit operations and data transfers)/0000_H (for 16-bit operations); otherwise the ZF is cleared to "0".

During the bit manipulation instruction [SET, CLR, and CPL], the ZF is set to "1" if the contents of the specified bit is "0"; otherwise the ZF is cleared to "0".

This flag is set to "1" when the upper 8 bits of the product are 00_H during the multiplication instruction [MUL], and when 00_H for the remainder during the division instruction [DIV]; otherwise it is cleared to "0".

(2) **Carry flag (CF)**

The CF is set to "1" when a carry out of the MSB (most significant bit) of the result occurred during addition or when a borrow into the MSB of the result occurred during subtraction; otherwise the CF is cleared to "0". During division, this flag is set to "1" when the divisor is 00_H (divided by zero error), or when the quotient is 100_H or higher (overflow error); otherwise it is cleared. The CF is also affected during the shift/rotate instructions [SHLC, SHRC, ROLC, and RORC]. The data shifted out from a register is set to the CF.

This flag is also a 1-bit register (a boolean accumulator) for the bit manipulation instructions.

Set/clear/complement are possible with the CF manipulation instructions.

Example1 : Bit manipulation

```
LD      CF, (0007H) . 5      ;  (0001H)2 ← (0007H)5 ∀ (009AH)0
XOR     CF, (009AH) . 0
LD      (0001H) . 2, CF
```

Example2 : Arithmetic right shift

```
LD      CF, A . 7
RORC    A
```

(3) **Half carry flag (HF)**

The HF is set to "1" when a carry occurred between bits 3 and 4 of the operation result during an 8-bit addition, or when a borrow occurred from bit 4 into bit 3 of the result during an 8-bit subtraction; otherwise the HF is cleared to "0". This flag is useful in the decimal adjustment for BCD operations (adjustments using the [DAA r], or [DAS r] instructions).

Example : BCD operation
(The A becomes 47_H after executing the following program when A = 19_H, B = 28_H)

```
ADD     A, B        ; A ← 41H, HF ← 1
DAA     A           ; A ← 41H + 06H = 47H (decimal-adjust)
```

(4) **Jump status flag (JF)**

Zero or carry information is set to the JF after operation (e. g. INC, ADD, CMP, TEST).
The JF provides the jump condition for conditional jump instructions [JRS T/F, $ + 2 + d], [JR T/F, $ + 2 + d] (T or F is a condition code). Jump is performed if the JF is "1" for a true condition (T), or the JF is "0" for a false condition (F).
The JF is set to "1" after executing the load/exchange/swap/nibble rotate/jump instruction, so that [JRS T, $ + 2 + d] and [JR T, $ + 2 + d] can be regarded as an unconditional jump instruction.

Example : Jump status flag and conditional jump instruction

```
INC     A
JRS     T, SLABLE1      ; Jump when a carry is caused by the immediately
 ⋮                        preceding operation instruction.
LD      A, (HL)
JRS     T, SLABLE2      ; JF is set to "1" by the immediately preceding
 ⋮                        instruction, making it an unconditional jump
                          instruction.
```

Example : The accumulator and flags will become as shown below after executing the following instructions when the WA register pair, the HL register pair, the data memory at address $00C5_H$, the carry flag and the half carry flag contents being "$219A_H$", "$00C5_H$", "$D7_H$", "1" and "0", respectively.

Instruction	Acc. after execution	Flag after execution			
		JF	ZF	CF	HF
ADDC A, (HL)	72	1	0	1	1
SUBB A, (HL)	C2	1	0	1	0
CMP A, (HL)	9A	0	0	1	0
AND A, (HL)	92	0	0	1	0
LD A, (HL)	D7	1	0	1	0
ADD A, 66H	00	1	1	1	1

Instruction	Acc. after execution	Flag after execution			
		JF	ZF	CF	HF
INC A	9B	0	0	1	0
ROLC A	35	1	0	1	0
RORC A	CD	0	0	0	0
ADD WA, 0F508H	16A2	1	0	1	0
MUL W, A	13DA	0	0	1	0
SET A.5	BA	1	1	1	0

1.7 Stack and Stack Pointer

1.7.1 Stack

The stack provides the area in which the return address or status, etc. are saved before a jump is performed to the processing routine during the execution of a subroutine call instruction or the acceptance of an interrupt. On a subroutine call instruction, the contents of the PC (the return address) is saved; on an interrupt acceptance, the contents of the PC and the PSW are saved (the PSW is pushed first, followed by PC_H and PC_L). Therefore, a subroutine call occupies two bytes on the stack; an interrupt occupies three bytes.
When returning from the processing routine, executing a subroutine return instruction [RET] restores the contents to the PC from the stack; executing an interrupt return instruction [RETI] / [RETN] restores the contents to the PC and the PSW (the PC_L is popped first, followed by PC_H and PSW).
The stack can be located anywhere within the data memory space except the register bank area, therefore the stack depth is limited only by the free data memory size.

1.7.2 Stack Pointer (SP)

The stack pointer (SP) is a 16-bit register containing the address of the next free locations on the stack.

The SP is postdecremented when a subroutine call or a push instruction is executed, or when an interrupt is accepted; and the SP is preincremented when a return or a pop instruction is executed. Figure 1-8 shows the stacking order.

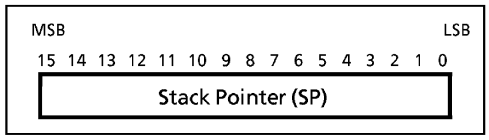


Figure 1-7. Stack Pointer

The SP is not initialized hardware-wise but requires initialization by an initialize routine (sets the highest stack address). [LD SP, mn], [LD SP, gg] and [LD gg, SP] are the SP access instructions (mn ; 16-bit immediate data, gg ; register pair).

Example 1 : To initialize the SP

```
LD      SP, 023FH       ; SP←023F_H
```

Example 2 : To read the SP

```
LD      HL, SP          ; HL←SP
```

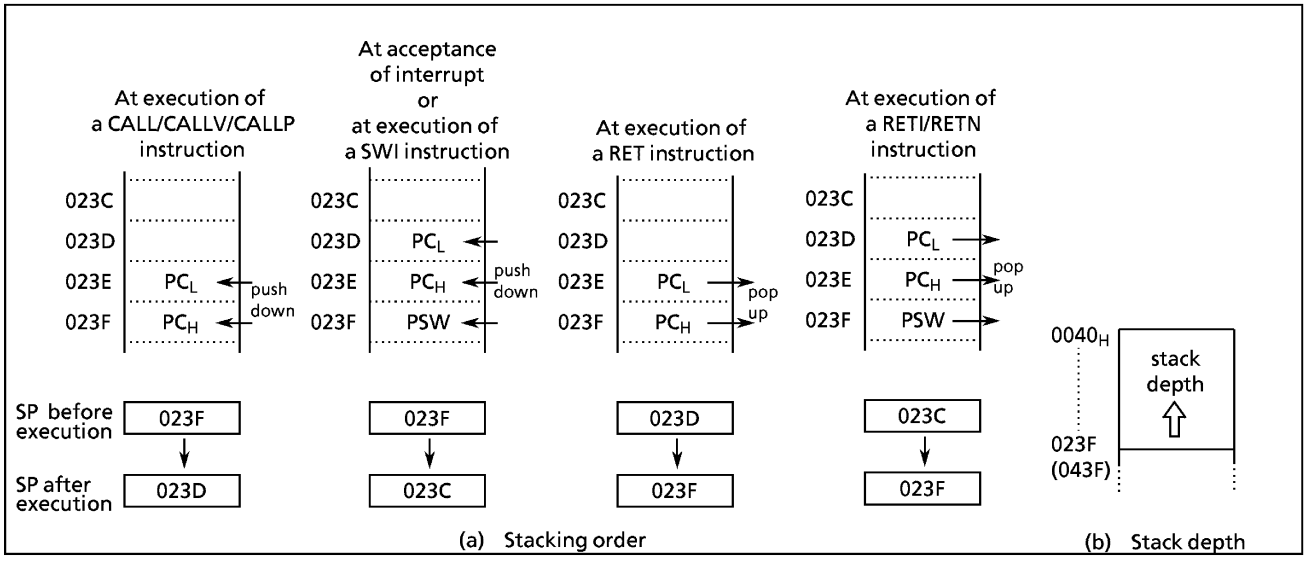


Figure 1-8. Stack

1.8 System Clock Controller

The system clock controller consists of a clock generator, a timing generator, and a stand-by controller.

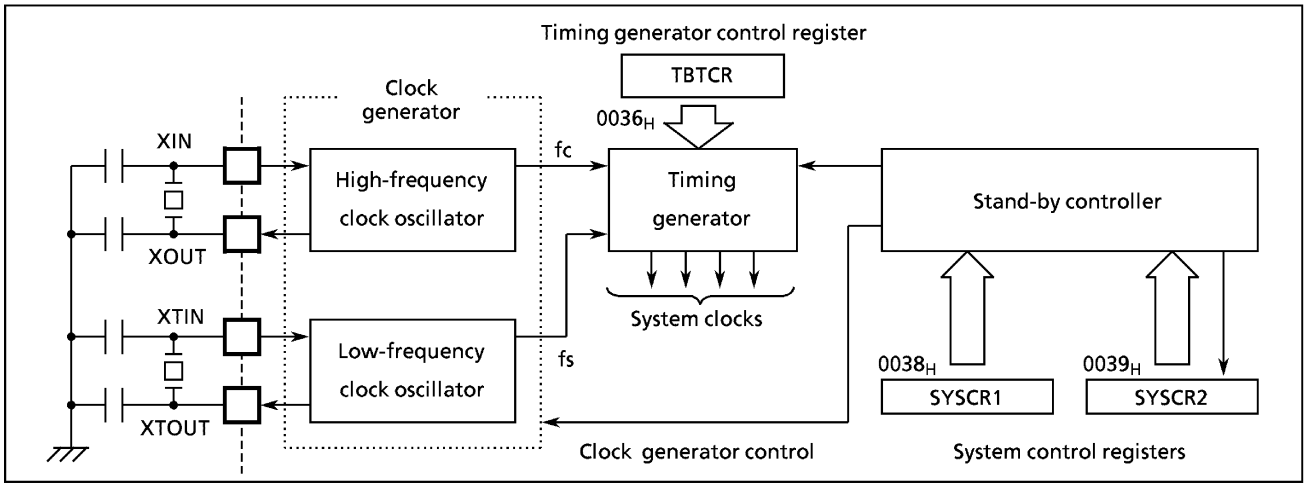


Figure 1-9. System Clock Controller

1.8.1 Clock Generator

The clock generator generates the basic clock which provides the system clocks supplied to the CPU core and on-chip peripheral hardware. It contains two oscillation circuits: one for the high-frequency clock and one for the low-frequency clock. Power consumption can be reduced by switching of the system clock controller to low-power operation based on the low-frequency clock.

The high-frequency (fc) and low-frequency (fs) clocks can be easily obtained by connecting a resonator between the XIN/XOUT and XTIN/XTOUT pins, respectively. Clock input from an external oscillator is also possible. In this case, external clock is applied to the XIN/XTIN pin with the XOUT/XTOUT pin not connected.

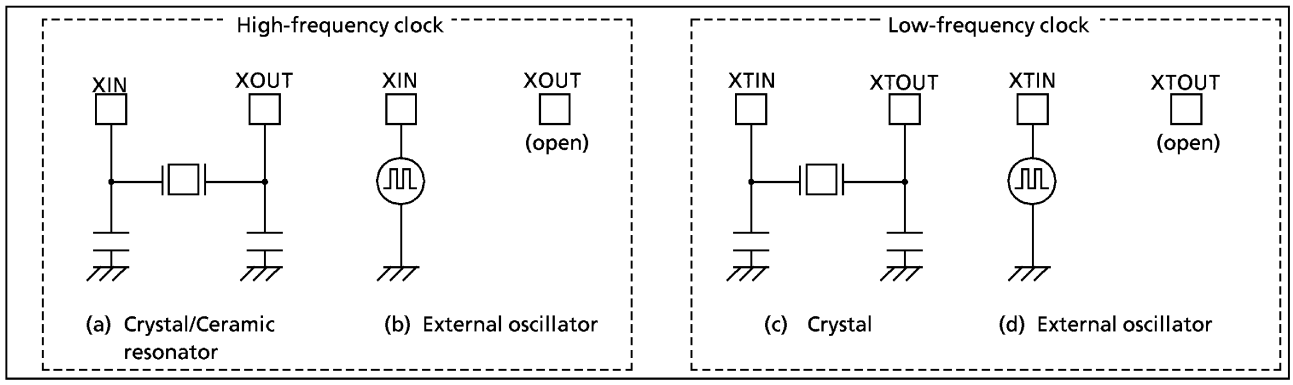
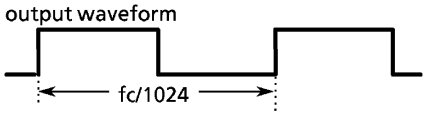


Figure 1-10. Examples of Resonator Connection

Note : Accurate Adjustment of the Oscillation Frequency:
Although no hardware to externally and directly monitor the basic clock pulse is not provided, the oscillation frequency can be adjusted by providing a program to output fixed frequency pulses to the port while disabling all interrupts and monitoring this pulse. With a system requiring adjustment of the oscillation frequency, the adjusting program must be created beforehand.

Example: To output the high-frequency oscillation frequency adjusting monitor pulse to P13 ($\overline{DVO}$a) pin.

```
SFCCHK:  LD   (P1CR),   00001000B  ;  Configures port P13 as an output
         SET  (P1).3                ;  P13 output latch ← 1
         LD   (TBTCR),  11100000B  ;  Enables divider output
         JRS  T, $                  ;  Loops endless
```



1.8.2 Timing Generator

The timing generator generates from the basic clock the various system clocks supplied to the CPU core and peripheral hardware. The timing generator provides the following functions :

① Generation of main system clock
② Generation of divider output ($\overline{DVO}$) pulses
③ Generation of source clocks for time base timer
④ Generation of source clocks for watchdog timer
⑤ Generation of internal source clocks for timer/counters TC1 – TC4
⑥ Generation of internal clocks for serial interface SIO
⑦ Generation of source clocks for VFT driver circuit
⑧ Generation of warm-up clocks for releasing STOP mode
⑨ Generation of a clock for releasing reset output

(1) Configuration of Timing Generator

The timing generator consists of a 21-stage divider with a divided-by-4 prescaler, a main system clock generator, and machine cycle counters. An input clock to the 7th stage of the divider depends on the operating mode and DV7CK (bit 4 in TBTCR) shown in Figure 1-11 as follows.

During reset and upon releasing STOP mode, the divider is cleared to "0", however, the prescaler is not cleared.

① In the single-clock mode
A divided-by-256 of high-frequency clock (fc/2^8) is input to the 7th stage of the divider.

② In the dual-clock mode
During NORMAL2 or IDLE2 mode (SYSCK = 0), an input clock to the 7th stage of the divider can be selected either "fc/2^8" or "fs" with DV7CK.
During SLOW or SLEEP mode (SYSCK = 1), "fs" is automatically input to the 7th stage. To input clock to the 1st stage is stopped ; output from the 1st to 6th stages is also stopped.

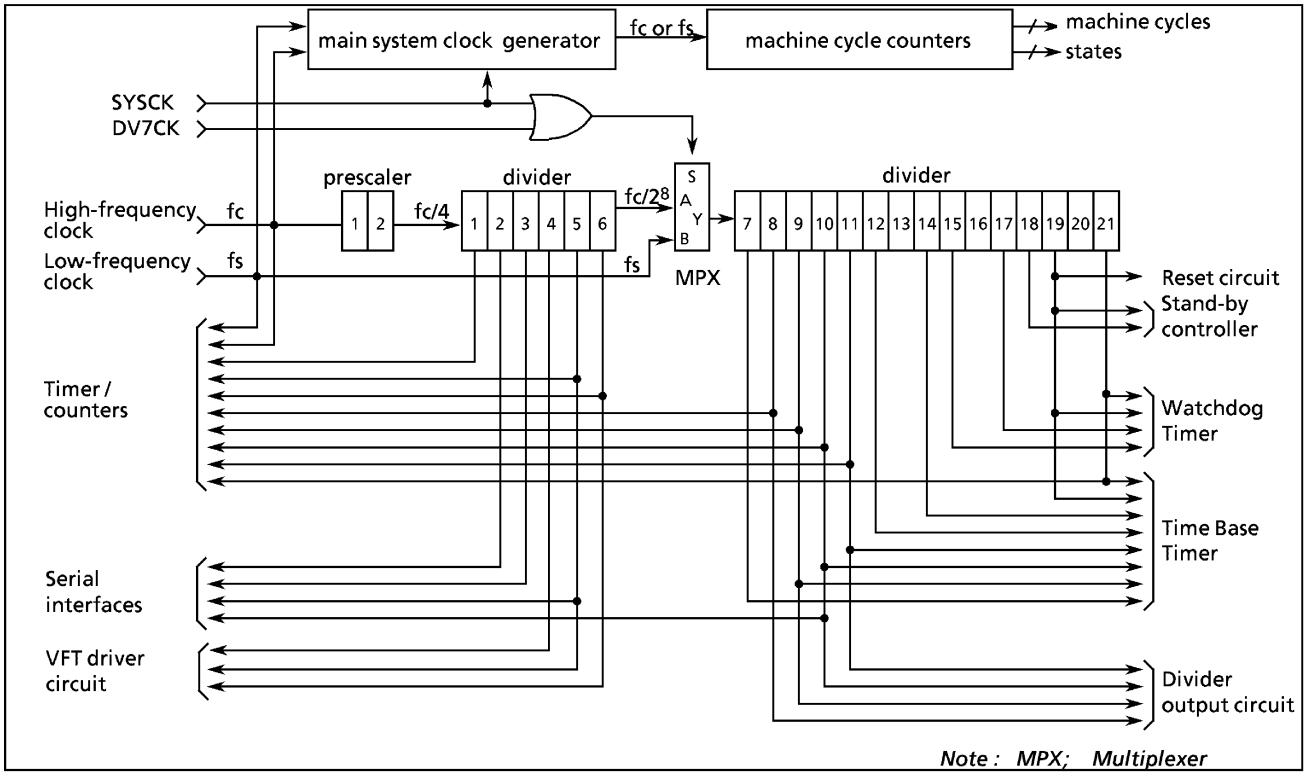


Figure 1-11. Configuration of Timing Generator

TBTCR (0036_H)

7	6	5	4	3	2	1	0
(DVOEN)	(DVOCK)		DV7CK	(TBTEN)	(TBTCK)		

(Initial value: 0**0 0***)

DV7CK	Selection of input clock to the 7th stage of the divider	0 : fc/2^8 [Hz] 1 : fs	R/W

*Note1 : fc ; high-frequency clock [Hz], fs ; low-frequency clock [Hz], * ; don't care*
Note 2 : Do not set DV7CK to "1" in the single-clock mode.
Note 3 : Do not set DV7CK to "1" before low-frequency clock is stable in the dual-clock mode.

Figure 1-12. Timing Generator Control Register

(2) Machine Cycle

Instruction execution and on-chip peripheral hardware operation are synchronized with the main system clock. The minimum instruction execution unit is called an "machine cycle". There are a total of 10 different types of instructions for the TLCS-870 Series: ranging from 1-cycle instructions which require one machine cycle for execution to 10-cycle instructions which require 10 machine cycles forexecution.

A machine cycle consists of 4 states (S0 - S3), and each state consists of one main system clock.

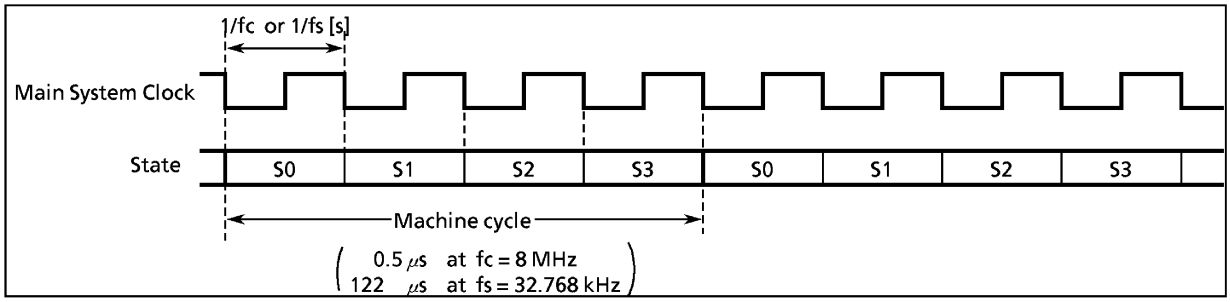


Figure 1-13. Machine Cycle

1.8.3 Stand-by Controller

The stand-by controller starts and stops the oscillation circuits for the high-frequency and low-frequency clocks, and switches the main system clock. There are two operating modes: single-clock and dual-clock. These modes are controlled by the system control registers (SYSCR1, SYSCR2).

Figure 1-14 shows the operating mode transition diagram and Figure 1-15 shows the system control registers. Either the single-clock or the dual-clock mode can be selected by an option during reset.

(1) **Single-clock mode**

Only the oscillation circuit for the high-frequency clock is used, and P21 (XTIN) and P22 (XTOUT) pins are used as input/output ports. In the single-clock mode, the machine cycle time is 4/fc [s] (0.5 μs at fc = 8 MHz).

① NORMAL1 mode

In this mode, both the CPU core and on-chip peripherals operate using the high-frequency clock. In the case where the single-clock mode has been selected as an option, the 87C814/H14/K14/M14 are placed in this mode after reset.

② IDLE1 mode

In this mode, the internal oscillation circuit remains active. The CPU and the watchdog timer are halted; however, on-chip peripherals remain active (operate using the high-frequency clock). IDLE1 mode is started by setting IDLE bit in the system control register 2 (SYSCR2), and IDLE1 mode is released to NORMAL1 mode by an interrupt request from the on-chip peripherals or external interrupt inputs. When IMF (interrupt master enable flag) is "1" (interrupt enable), the execution will resume upon acceptance of the interrupt, and the operation will return to normal after the interrupt service is completed. When IMF is "0" (interrupt disable), the execution will resume with the instruction which follows IDLE mode start instruction.

③ STOP1 mode

In this mode, the internal oscillation circuit is turned off, causing all system operations to be halted. The internal status immediately prior to the halt is held with the lowest power consumption during this mode. The output status of all output ports can be set to either output hold or high-impedance under software control.

STOP1 mode is started by setting STOP bit in the system control register 1 (SYSCR1), and STOP1 mode is released by an input (either level-sensitive or edge-sensitive can be programmably selected) to the $\overline{\text{STOP}}$ pin. After the warming-up period is completed, the execution resumes with the next instruction which follows the STOP mode start instruction.

(2) **Dual-clock mode**

Both the high-frequency and low-frequency oscillation circuits are used in this mode. P21 (XTIN) and P22 (XTOUT) pins cannot be used as input/output ports. The main system clock is obtained from the high-frequency clock in NORMAL2 and IDLE2 modes, and is obtained from the low-frequency clock in SLOW and SLEEP modes. The machine cycle time is 4/fc [s] (0.5 μs at fc = 8 MHz) in NORMAL2 and IDLE2 modes, and 4/fs [s] (122 μs at fs = 32.768 kHz) in SLOW and SLEEP modes. Note that *the 87PM14 is placed in the single-clock mode during reset.* To use the dual-clock mode, the low-frequency oscillator should be turned on by executing [SET (SYSCR2).XTEN] instruction.

① NORMAL2 mode

In this mode, the CPU core is operated using the high-frequency clock. The on-chip peripherals are operated on the high-frequency clock and/or low-frequency clock. In case that the dual-clock mode has been selected as an option, the 87C814/H14/K14/M14 are placed in this mode after reset.

② SLOW mode

This mode can be used to reduce power-consumption by turning off oscillation of the high-frequency clock. The CPU core and on-chip peripherals are operated using the low-frequency clock.

Switching back and forth between NORMAL2 and SLOW modes is performed by the system control register 2.

③ IDLE2 mode

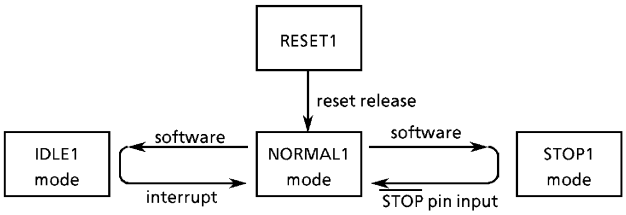
In this mode, the internal oscillation circuits remain active. The CPU and the watchdog timer are halted; however, on-chip peripherals operate using the high-frequency clock and/or the low-frequency clock. Starting and releasing of IDLE2 mode are the same as for IDLE1 mode, except that operation returns to NORMAL2 mode.

④ SLEEP mode

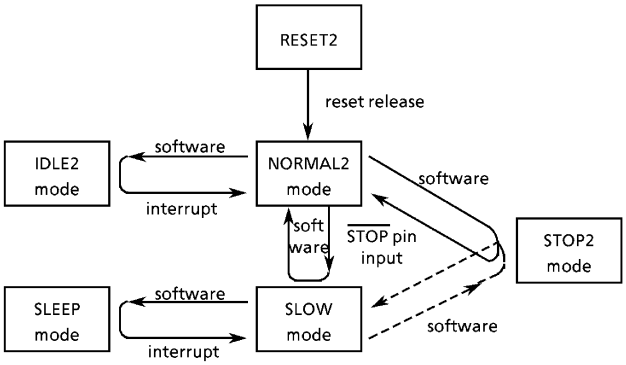
In this mode, the internal oscillation circuit of the low-frequency clock remains active. The CPU, the watchdog timer, and the internal oscillation circuit of the high- frequency clock are halted; however, on-chip peripherals operate using the low-frequency clock. Starting and releasing of SLEEP mode is the same as for IDLE1 mode, except that operation returns to SLOW mode.

⑤ STOP2 mode

As in STOP1 mode, all system operations are halted in this mode.



(a) Single-clock mode



(b) Dual-clock mode

Note 1 : NORMAL1 and NORMAL2 modes are generically called NORMAL; STOP1 and STOP2 are called STOP; and IDLE1, IDLE2 and SLEEP are called IDLE.

Note 2 : There is no RESET2 mode in 87PM14.

Operating mode		Frequency		CPU core	On-chip Peripherals	Machine cycle time
		High-frequency	Low-frequency			
Single-Clock	RESET1	turning on oscillation	turning off oscillation	reset	reset	4/fc [s]
	NORMAL1			operate	operate (Note 1)	
	IDLE1			halt		
	STOP1	turning off oscillation			halt	—
Dual-Clock	RESET2	turning on oscillation	turning on oscillation	reset	reset	4/fc [s]
	NORMAL2			High-frequency	operate (High and/or Low) (Note 1)	
	IDLE2			halt		
	SLOW	turning off oscillation		Low-frequency	Low-frequency (Note 2)	4/fs [s]
	SLEEP			halt		
	STOP2		turning off oscillation		halt	—

Note 1: The Vacuum Fluorescent Tube (VFT) driver circuit are halted.

Figure 1-14. Operating Mode Transition Diagram

System Control Register 1

SYSCR1 (0038_H)

7	6	5	4	3	2	1	0
STOP	RELM	RETM	OUTEN	WUT			

(Initial value: 0000 00**)

STOP	STOP mode start	0 : CPU core and peripherals remain active 1 : CPU core and peripherals are halted (start STOP mode)	R/W
RELM	Release method for STOP mode	0 : Edge-sensitive release 1 : Level-sensitive release	
RETM	Operating mode after STOP mode	0 : Return to NORMAL mode 1 : Return to SLOW mode	
OUTEN	Port output control during STOP mode	0 : High-impedance 1 : Remain unchanged	
WUT	Warming-up time at releasing STOP mode	00 : 3×2^{19}/fc or 3×2^{13}/fs [s] 01 : 2^{19}/fc or 2^{13}/fs 1* : Reserved	

Note 1 : *Always set RETM to "0" when transiting from NORMAL1 mode to STOP1 mode and from Normal2 mode to STOP2 mode.*
Always set RETM to "1" when transiting from SLOW mode to STOP2 mode.

Note 2 : *When STOP mode is released with $\overline{\text{RESET}}$ pin input, a return is made to NORMAL mode regardless of the RETM contents.*

Note 3 : fc ; *high-frequency clock* [Hz]
fs ; *low-frequency clock* [Hz]
* ; *don't care*

Note 4 : *Bits 1 and 0 in SYSCR1 are read in as undefined data when a read instruction is executed.*

Note 5 : *When the STOP mode is started by specifying OUTEN = "0", the internal input is fixed to "0" and the interrupt of the falling edge may be set.*

System Control Register 2

SYSCR2 (0039_H)

7	6	5	4	3	2	1	0
XEN	XTEN	SYSCK	IDLE				

(Initial value: 10/100 ****)

XEN	High-frequency oscillator control	0 : Turn off oscillation 1 : Turn on oscillation	R/W
XTEN	Low-frequency oscillator control	0 : Turn off oscillation 1 : Turn on oscillation	
SYSCK	Main system clock select (write)/main system clock monitor (read)	0 : High-frequency clock 1 : Low-frequency clock	
IDLE	IDLE mode start	0 : CPU and watchdog timer remain active 1 : CPU and watchdog timer are stopped (start IDLE mode)	

Note 1 : *A reset is applied ($\overline{\text{RESET}}$ pin output goes low) if both XEN and XTEN are cleared to "0".*

Note 2 : *Do not clear XEN to "0" when SYSCK = 0, and do not clear XTEN to "0" when SYSCK = 1.*

Note 3 : *WDT; watchdog timer, * ; don't care*

Note 4 : *Bits 3 - 0 in SYSCR2 are always read in as "1" when a read instruction is executed.*

Note 5 : *An optional initial value can be selected for XTEN. Always specify when ordering ES (engineering sample).*

Note 6 : *The instruction for specifying Masking Option (Operating Mode) in ES Order Sheet is described in ADDITIONAL INFORMATION "Notice for Masking Option of TLCS-870 and TLCS-870/X series" section 8.*

XTEN	operating mode after reset
0	Single-clock mode (NORMAL1)
1	Dual-clock mode (NORMAL2)

Figure 1-15. System Control Registers

1.8.4 Operating Mode Control

(1) **STOP** mode (STOP1, STOP2)

STOP mode is controlled by the system control register 1 (SYSCR1) and the $\overline{\text{STOP}}$ pin input. The $\overline{\text{STOP}}$ pin is also used both as a port P20 and an $\overline{\text{INT5}}$ (external interrupt input 5) pin. STOP mode is started by setting STOP (bit 7 in SYSCR1) to "1". During STOP mode, the following status is maintained.

① Oscillations are turned off, and all internal operations are halted.

② The data memory, registers (except for DBR) and port output latches are all held in the status in effect before STOP mode was entered. The port output can be select either output hold or high-impedance by setting OUTEN (bit 4 in SYSCR1).

③ The divider of the timing generator is cleared to "0".

④ The program counter holds the address of the instruction following the instruction which started STOP mode.

STOP mode includes a level-sensitive release mode and an edge-sensitive release mode, either of which can be selected with RELM (bit 6 in SYSCR1).

a. Level-sensitive release mode (RELM = 1)

In this mode, STOP mode is released by setting the $\overline{\text{STOP}}$ pin high. This mode is used for capacitor back-up when the main power supply is cut off and for long term battery back-up.

When the $\overline{\text{STOP}}$ pin input is high, executing an instruction which starts the STOP mode will not place in the STOP mode but instead will immediately start the release sequence (warm-up). Thus, to start the STOP mode in the level-sensitive release mode, it is necessary for the program to first confirm that the $\overline{\text{STOP}}$ pin input is low. The following two methods can be used for confirmation:

- Using an external interrupt input $\overline{\text{INT5}}$ ($\overline{\text{INT5}}$ is a falling edge-sensitive input).

Example : Starting STOP mode with an INT5 interrupt.

```
PINT5 :   TEST   (P2) . 0                  ;  To reject noise, STOP mode does not start if
                                              port P20 is at high
          JRS    F, SINT5
          LD     (SYSCR1), 01000000B       ;  Sets up the level-sensitive release mode.
          SET    (SYSCR1) . 7              ;  Starts STOP mode
          LDW    (IL), 1111011101010111B   ;  IL11, 7, 5, 3 ← 0 (Clears interrupt latches)
SINT5 :   RETI
```

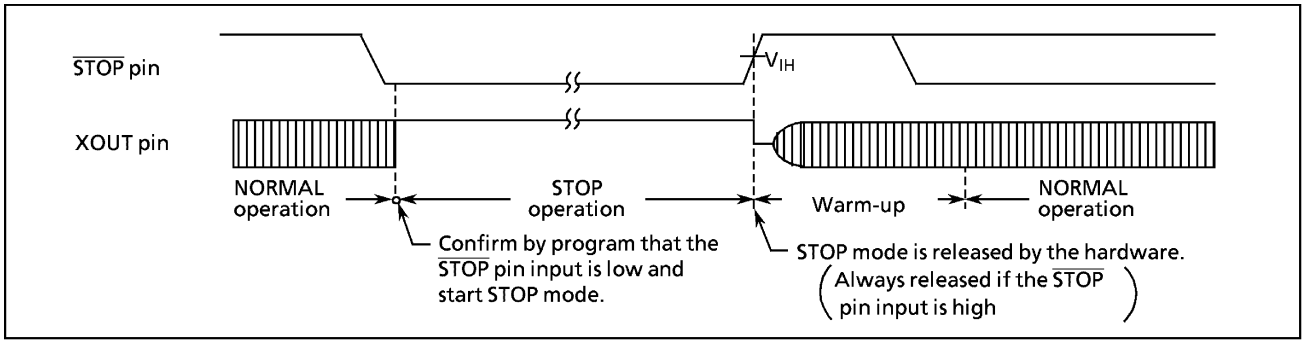


Figure 1-16. Level-sensitive Release Mode

Note : When changing to the level-sensitive release mode from the edge-sensitive release mode, the release mode is not switched until a rising edge of the $\overline{\text{STOP}}$ pin input is detected.

b. Edge-sensitive release mode (RELM = 0)

In this mode, STOP mode is released by a rising edge of the $\overline{STOP}$ pin input. This is used in applications where a relatively short program is executed repeatedly at periodic intervals. This periodic signal (for example, a clock from a low-power consumption oscillator) is input to the $\overline{STOP}$ pin.

In the edge-sensitive release mode, STOP mode is started even when the $\overline{STOP}$ pin input is high.

Example : Starting STOP mode operation in the edge-sensitive release mode

```
LD      (SYSCR1), 00000000B     ; OUTEN ← 0 (specifies high-impedance)
DI                              ; IMF ← 0 (disables interrupt service)
SET     (SYSCR1). STOP          ; STOP ← 1 (activates stop mode)
LDW     (IL), 11011101010111B)  ; IL11, 7, 5, 3 ← 1 (clears interrupt latches)
EI                              ; IMF ← 1 (enables interrupt service)
```

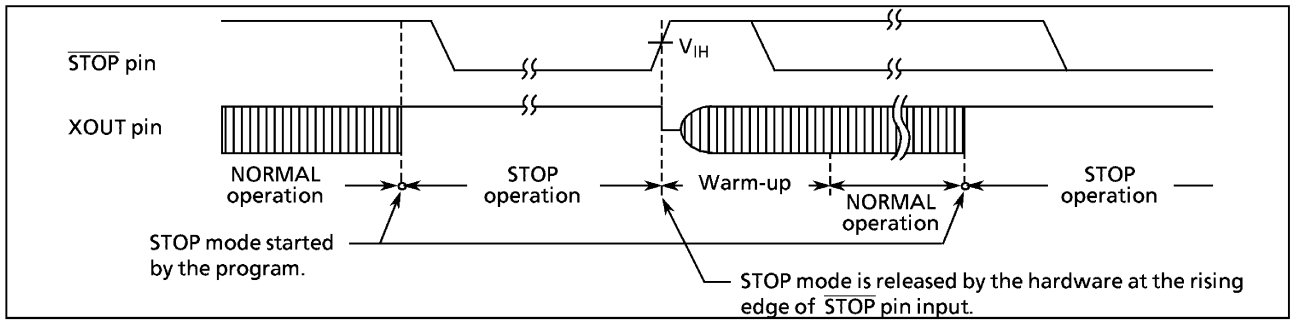


Figure 1-17. Edge-sensitive Release Mode

STOP mode is released by the following sequence:

① In the dual-clock mode. When returning to NORMAL2, both the high-frequency and low-frequency clock oscillators are turned on ; when returning to SLOW mode, only the low-frequency clock oscillator is turned on. When returning to Normal 1, only the high-frequency clock oscillator is turned on.

② A warming-up period is inserted to allow oscillation time to stabilize. During warm-up, all internal operations remain halted. Two different warming-up times can be selected with WUT (bits 2 and 3 in SYSCR1) as determined by the resonator characteristics.

③ When the warming-up time has elapsed, normal operation resumes with the instruction following the STOP mode start instruction (e.g. [SET (SYSCR1). 7]). The start is made after the divider of the timing generator is cleared to "0".

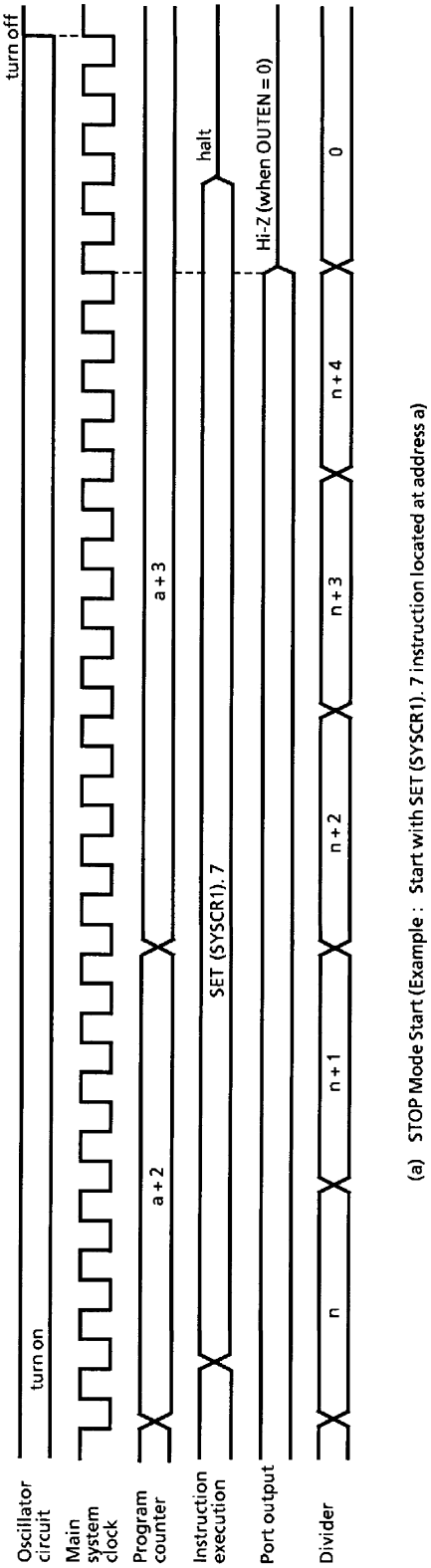
Table 1-1. Warming-up Time Example

Return to NORMAL1 mode			Return to SLOW mode	
WUT	At fc = 4.194304 MHz	At fc = 8 MHz	WUT	At fs = 32.768 kHz
3×2^{19} / fc [s]	375 [ms]	196.6 [ms]	3×2^{13} / fs [s]	750 [ms]
2^{19} / fc	125	65.5	2^{13} / fs	250

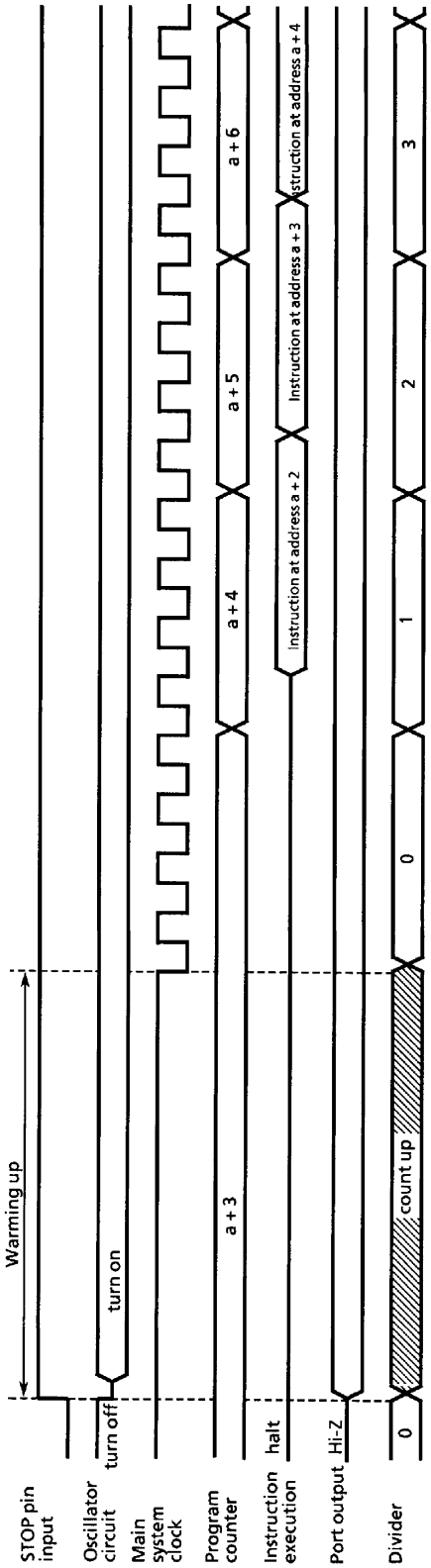
Note : The warming-up time is obtained by dividing the basic clock by the divider: therefore, the warming-up time may include a certain amount of error if there is any fluctuation of the oscillation frequency when STOP mode is released. Thus, the warming-up time must be considered an approximate value.

STOP mode can also be released by setting the $\overline{RESET}$ pin low, which immediately performs the normal reset operation.

In this case, even if the setting is to return to the SLOW mode, it starts from the NORMAL2 mode. (In case of 87PM14, starts from NORMAL1 mode after reset release)



(a) STOP Mode Start (Example : Start with SET (SYSCR1). 7 instruction located at address a)



(b) STOP Mode Release

Figure 1-18. STOP Mode Start / Release

Note : When STOP mode is released with a low hold voltage, the following cautions must be observed.
The power supply voltage must be at the operating voltage level before releasing the STOP mode. The $\overline{\text{RESET}}$ pin input must also be high, rising together with the power supply voltage. In this case, if an external time constant circuit has been connected, the $\overline{\text{RESET}}$ pin input voltage will increase at a slower rate than the power supply voltage. At this time, there is a danger that a reset may occur if input voltage level of the $\overline{\text{RESET}}$ pin drops below the non-inverting high-level input voltage (hysteresis input).

(2) **IDLE** mode (IDLE1, IDLE2, SLEEP)

IDLE mode is controlled by the system control register 2 and maskable interrupts. The following status is maintained during IDLE mode.

① Operation of the CPU and watchdog timer is halted. The on-chip peripherals continue to operate.
② The data memory, CPU registers and port output latches are all held in the status in effect before IDLE mode was entered.
③ The program counter holds the address of the instruction following the instruction which started IDLE mode.

Example : Starting IDLE mode.

```
SET        (SYSCR2) . 4        ;  IDLE←1
```

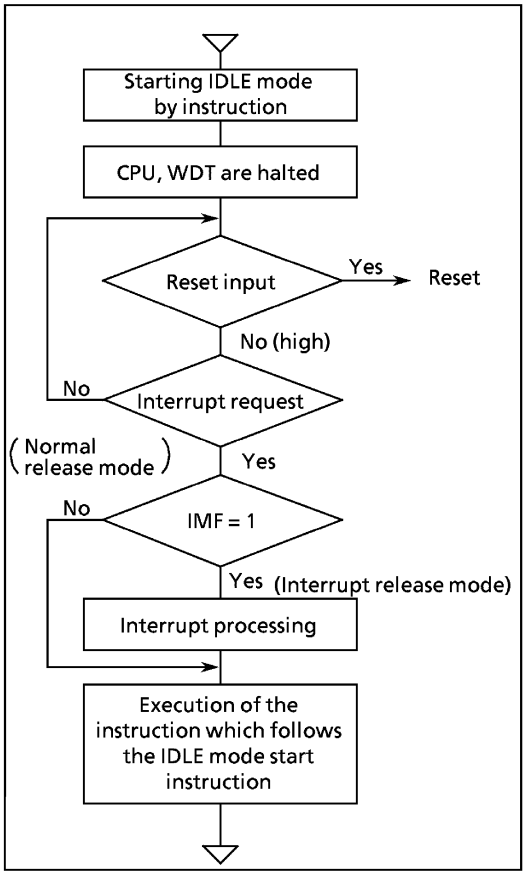


Figure 1-19. IDLE Mode

IDLE mode includes a normal release mode and an interrupt release mode. Selection is made with the interrupt master enable flag (IMF). Releasing the IDLE mode returns from IDLE1 to NORMAL1, from IDLE2 to NORMAL2, and from SLEEP to SLOW mode.

a. Normal release mode (IMF = "0")
IDLE mode is released by any interrupt source enabled by the individual interrupt enable flag (EF) or an external interrupt 0 ($\overline{\text{INT0}}$ pin) request. Execution resumes with the instruction following the IDLE mode start instruction (e.g. [SET (SYSCR2).4]). The interrupt latch (IL) of the interrupt source used for releasing the IDLE mode must be cleared to "0" by load instruction.

b. Interrupt release mode (IMF = "1")
IDLE mode is released and interrupt processing is started by any interrupt source enabled with the individual interrupt enable flag (EF) or an external interrupt 0 ($\overline{\text{INT0}}$ pin) request. After the interrupt is processed, the execution resumes from the instruction following the instruction which started IDLE mode.

IDLE mode can also be released by setting the $\overline{\text{RESET}}$ pin low, which immediately performs the reset operation. After reset, the 87C814/H14/K14/M14 are placed in NORMAL mode.
(The 87PM14 is placed in NORMAL1 mode after reset release)

Note : When a watchdog timer interrupt is generated immediately before IDLE mode is started, the watchdog timer interrupt will be processed but IDLE mode will not be started.

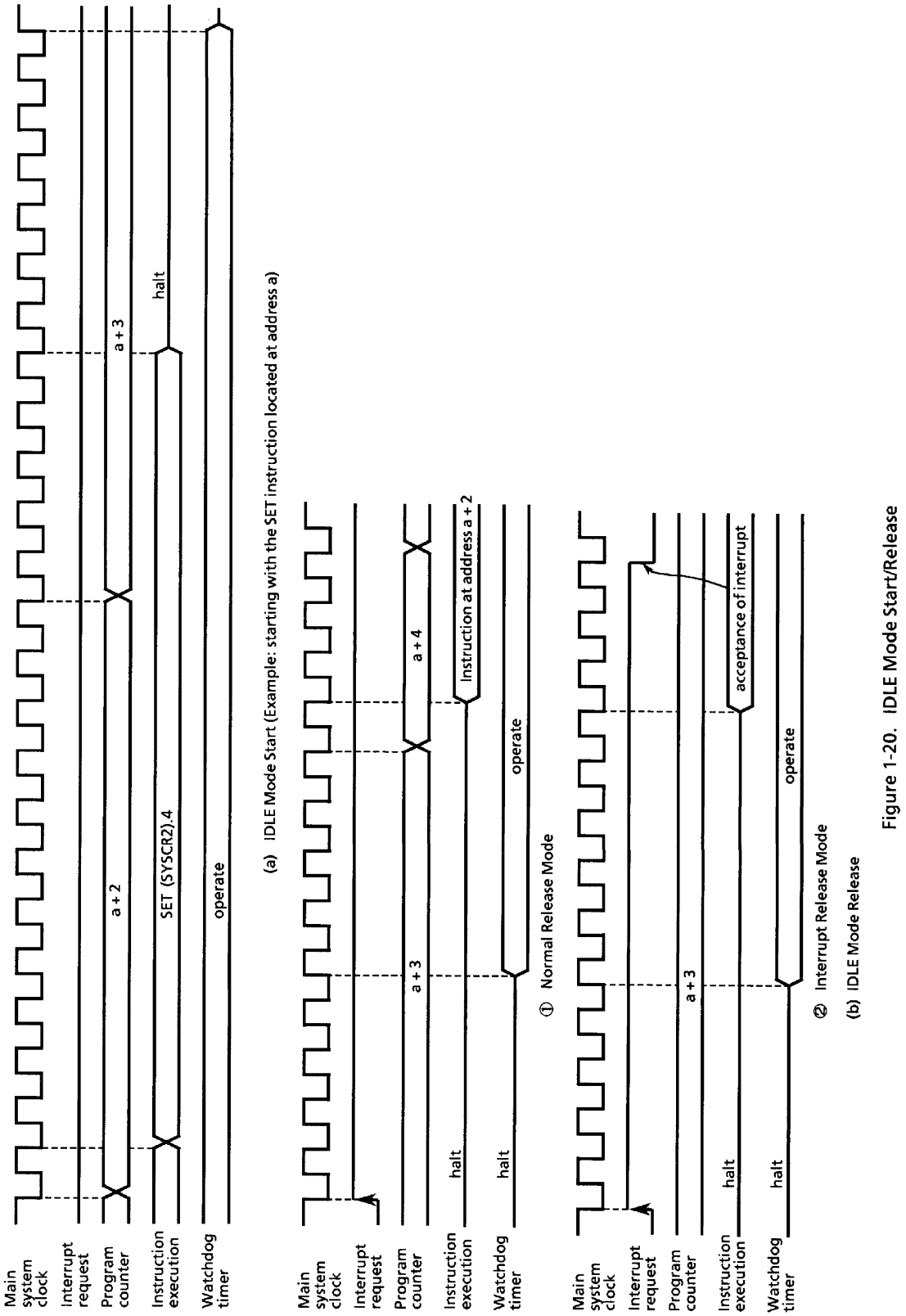


Figure 1-20. IDLE Mode Start/Release

(3) **SLOW** mode

SLOW mode is controlled by the system control register 2 and the timer/counter 2.

a. Switching from NORMAL2 mode to SLOW mode

First, set SYSCK (bit 5 in SYSCR2) to switch the main system clock to the low-frequency clock. Next, clear XEN (bit 7 in SYSCR2) to turn off high-frequency oscillation.

When the low-frequency clock oscillation is unstable, wait until oscillation stabilizes before performing the above operations. The timer/counter 2 (TC2) can conveniently be used to confirm that low-frequency clock oscillation has stabilized.

Note : The high frequency clock can be continued oscillation in order to return to NORMAL2 mode from SLOW mode quickly. Always turn off oscillation of high frequency clock when switching from SLOW mode to STOP mode.

Example1 : Switching from NORMAL2 mode to SLOW mode.

```
        SET     (SYSCR2) . 5        ; SYSCK←1   (Switches the main system clock to the
                                                low-frequency clock)
        CLR     (SYSCR2) . 7        ; XEN←0     (turns off high-frequency oscillation)
```

Example2 : Switching to SLOW mode after low-frequency clock oscillation has stabilized.

```
        LD      (TC2CR), 14H        ; Sets TC2 mode
                                      (timer mode, source clock : fs)
        LDW     (TREG2), 8000H      ; Sets warming-up time
                                      (according to Xtal characteristics)
        SET     (EIRH). EF14        ; Enable INTTC2
        LD      (TC2CR), 34H        ; Starts TC2
          :
PINTTC2 : LD    (TC2CR), 10H        ; Stops TC2
        SET     (SYSCR2) . 5        ; SYSCK←1
        CLR     (SYSCR2) . 7        ; XEN←0
        RETI
          :
VINTTC2 : DW    PINTTC2             ; INTTC2 vector table
```

b. Switching from SLOW mode to NORMAL2 mode

First, set XEN (bit 7 in SYSCR2) to turn on the high-frequency oscillation. When time for stabilization (warm-up) has been taken by the timer/counter 2 (TC2), clear SYSCK (bit 5 in SYSCR2) to switch the main system clock to the high-frequency clock.

Note1 : After the SYSCK is cleared to "0", the CPU core operate using low frequency clock when the main system clock is switching from low frequency clock to high frequency clock.

Note2 : SLOW mode can also be released by setting the $\overline{\text{RESET}}$ pin low, which immediately performs the reset operation. After reset, the 87C814/H14/K14/M14/PM14 are placed in NORMAL mode. (The PM14 is placed in NORMAL1 mode)

Example : Switching from SLOW mode to NORMAL2 mode (fc = 8 MHz, warming-up time is about 7.9 ms).

```
                SET     (SYSCR2) . 7          ; XEN←1      (turns on high-frequency oscillation)
                LD      (TC2CR), 10H          ; Sets TC2 mode
                                                (timer mode, source clock: fc)
                LD      (TREG2 + 1), 0F8H     ; Sets the warming-up time
                                                (according to frequency and resonator characteristics)
                LD      (TC2CR), 30H          ; Starts TC2
                 ⋮
PINTTC2 :       LD      (TC2CR), 10H          ; Stops TC2
                CLR     (SYSCR2) . 5          ; SYSCK←0    (Switches the main system clock to the
                                                            high-frequency clcok)
                RETI
                 ⋮
VINTTC2 :       DW      PINTTC2               ; INTTC2 vector table
```

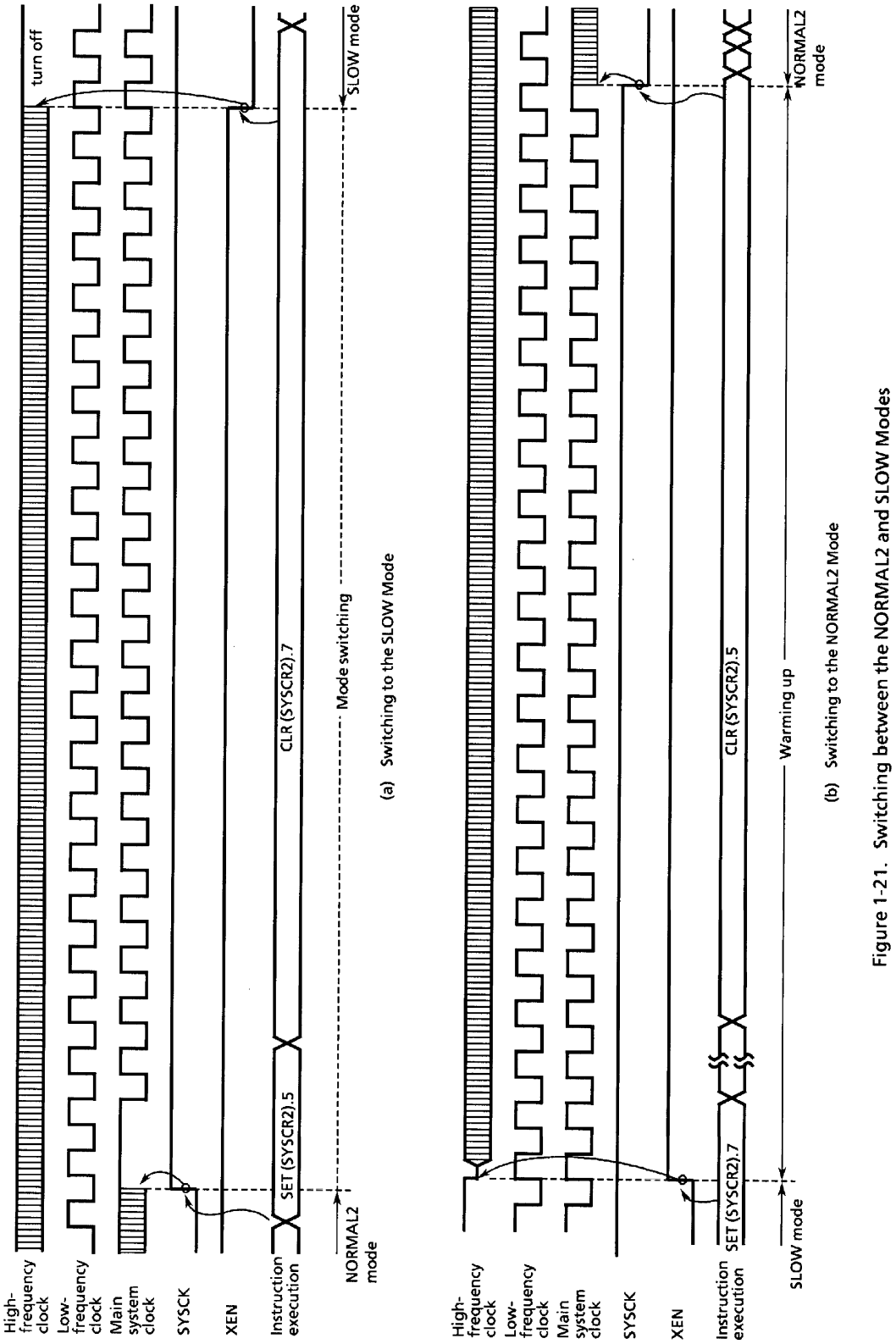


Figure 1-21. Switching between the NORMAL2 and SLOW Modes

1.9 Interrupt Controller

The 87C814/H14/K14/M14 each have a total of 13 interrupt sources: 5 externals and 8 internals. Nested interrupt control with priorities is also possible. Two of the internal sources are pseudo non-maskable interrupts; the remainder are all maskable interrupts.

Interrupt latches (IL) that hold the interrupt requests are provided for interrupt sources. Each interrupt vector is independent.

The interrupt latch is set to "1" when an interrupt request is generated and requests the CPU to accept the interrupt. The acceptance of maskable interrupts can be selectively enabled and disabled by the program using the interrupt master enable flag (IMF) and the individual interrupt enable flags (EF). When two or more interrupts are generated simultaneously, the interrupt is accepted in the highest priority order as determined by the hardware. Figure 1-22 shows the interrupt controller.

Table 1-2. Interrupt Sources

Interrupt Source		Enable Condition	Interrupt Latch	Vector Table Address	Priority
Internal/ External	(Reset)	Non-Maskable	—	$FFFE_H$	High 0
Internal	INTSW (Software interrupt)	Pseudo non-maskable	—	$FFFC_H$	1
Internal	INTWDT (Watchdog Timer interrupt)		IL_2	$FFFA_H$	2
External	INT0 (External interrupt 0)	IMF = 1, INT0EN = 1	IL_3	$FFF8_H$	3
Internal	INTTC1 (16-bit TC1 interrupt)	IMF · EF_4 = 1	IL_4	$FFF6_H$	4
External	INT1 (External interrupt 2)	IMF · EF_5 = 1	IL_5	$FFF4_H$	5
Internal	INTTBT (Time Base Timer interrupt)	IMF · EF_6 = 1	IL_6	$FFF2_H$	6
External	INT2 (External interrupt 2)	IMF · EF_7 = 1	IL_7	$FFF0_H$	7
Internal	INTTC3 (8-bit TC3 interrupt)	IMF · EF_8 = 1	IL_8	$FFEE_H$	8
Internal	INTSIO (Serial Interface1 interrupt)	IMF · EF_9 = 1	IL_9	$FFEC_H$	9
Internal	INTTC4 (8-bit TC4 interrupt)	IMF · EF_{10} = 1	IL_{10}	$FFEA_H$	10
External	INT3 (External interrupt 3)	IMF · EF_{11} = 1	IL_{11}	$FFE8_H$	11
reserved		IMF · EF_{12} = 1	IL_{12}	$FFE6_H$	12
reserved		IMF · EF_{13} = 1	IL_{13}	$FFE4_H$	13
Internal	INTTC2 (16-bit TC2 interrupt)	IMF · EF_{14} = 1	IL_{14}	$FFE2_H$	14
External	INT5 (External interrupt 5)	IMF · EF_{15} = 1	IL_{15}	$FFE0_H$	Low 15

(1) **Interrupt Latches** ($IL_{15\sim2}$)

Interrupt latches are provided for each source, except for a software interrupt. The latch is set to "1" when an interrupt request is generated, and requests the CPU to accept the interrupt. The latch is cleared to "0" just after the interrupt is accepted. All interrupt latches are initialized to "0" during reset.

Interrupt latches are assigned to addresses $003C_H$ and $003D_H$ in the SFR. Each latch can be cleared to "0" individually by an instruction; however, *the read-modify-write instruction such as bit manipulation or operation instructions cannot be used (Do not clear the IL_2 for a watchdog timer intlerrupt to "0").* Thus, interrupt requests can be canceled and initialized by the program. Note that interrupt latches cannot be set to "1" by any instruction.

The contents of interrupt latches can be read out by an instruction. Therefore, testing interrupt requests by software is possible.

Example 1 : Clears interrupt latches

```
LDW     (IL), 1011100000111111B     ; IL14, IL10~IL6←0
```

Example 2 : Reads interrupt latches

```
LD      WA, (IL)                    ; W←ILH, A←ILL
```

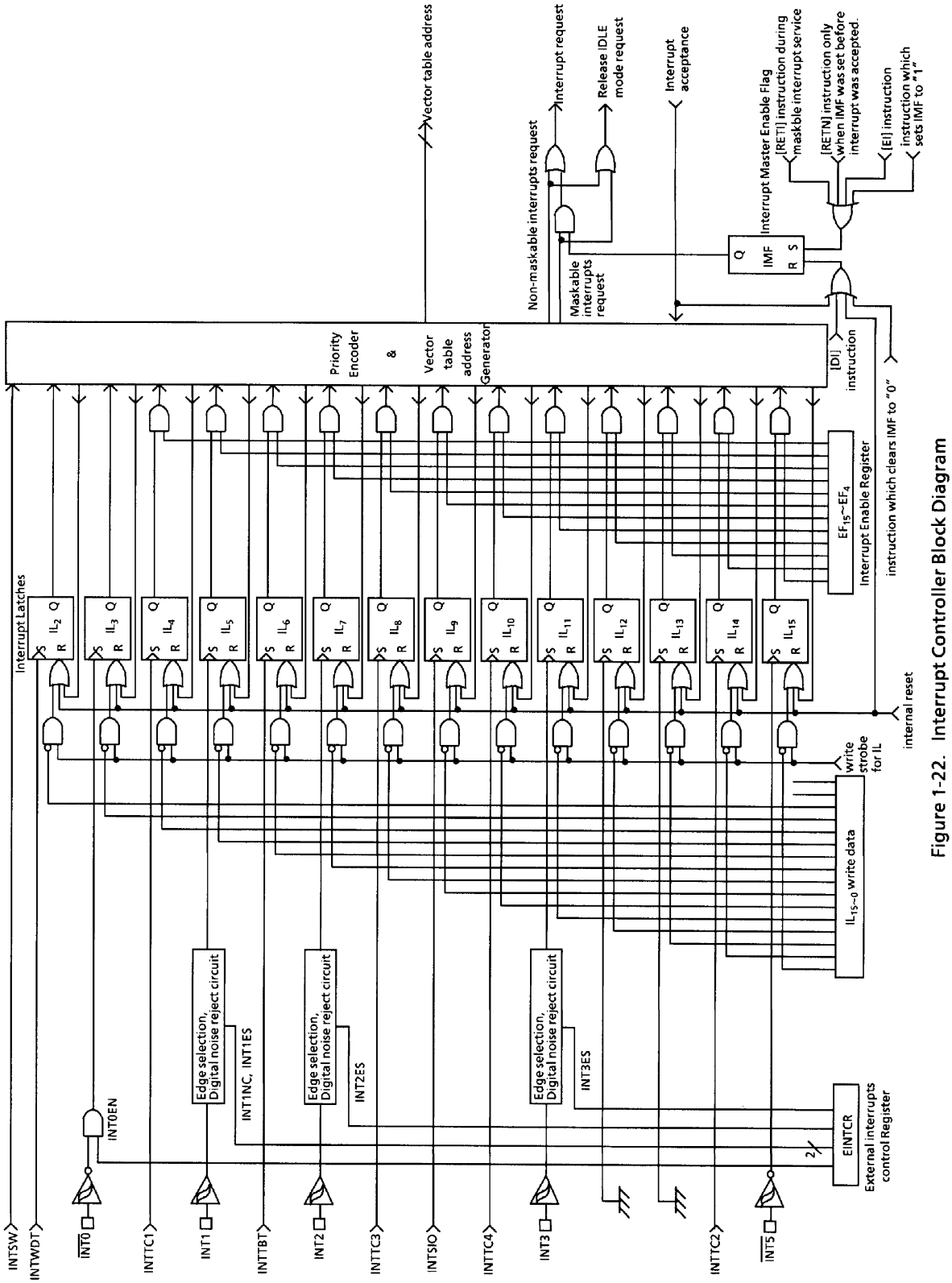


Figure 1-22. Interrupt Controller Block Diagram

Example 3: Tests an interrupt latch

```
TEST    (IL).7          ; if IL7 = 1 then jump
JR      F, SSET
```

(2) **Interrupt Enable Register** (EIR)

The interrupt enable register (EIR) enables and disables the acceptance of interrupts, except for the pseudo non-maskable interrupts (software and watchdog timer interrupts). Pseudo non-maskable interrupts are accepted regardless of the contents of the EIR; however, the pseudo non-maskable interrupts cannot be nested more than once at the same time. For example, the watchdog timer interrupt is not accepted during the software interrupt service.

The EIR consists of an interrupt master enable flag (IMF) and the individual interrupt enable flags (EF). This register is assigned to addresses $003A_H$ and $003B_H$ in the SFR, and can be read and written by an instruction (including read-modify-write instructions such as bit manipulation instructions).

① **Interrupt Master enable Flag** (IMF)

The interrupt master enable flag (IMF) enables and disables the acceptance of all interrupts, except for pseudo non-maskable interrupts. Clearing this flag to "0" disables the acceptance of all maskable interrupts. Setting to "1" enables the acceptance of interrupts.

When an interrupt is accepted, this flag is cleared to "0" to temporarily disable the acceptance of maskable interrupts. After execution of the interrupt service program, this flag is set to "1" by the maskable interrupt return instruction [RETI] to again enable the acceptance of interrupts. If an interrupt request has already been occurred, interrupt service starts immediately after execution of the [RETI] instruction.

Pseudo non-maskable interrupts are returned by the [RETN] instruction. In this case, the IMF is set to "1" only when pseudo non-maskable interrupt service is started with interrupt acceptance enabled (IMF = 1). Note that the IMF remains "0" when cleared in the interrupt service program.

The IMF is assigned to bit 0 at address $003A_H$ in the SFR, and can be read and written by an instruction. The IMF is normally set and cleared by the [EI] and [DI] instructions, and the IMF is initialized to "0" during reset.

Note : Do not set IMF to "1" during non-maskable interrupt service programs.

② **Individual interrupt Enable Flags** (EF_{15}~EF_4)

These flags enable and disable the acceptance of individual maskable interrupts, except for an external interrupt 0. Setting the corresponding bit of an individual interrupt enable flag to "1" enables acceptance of an interrupt, setting the bit to "0" disables acceptance.

Example 1 : Sets EF for individual interrupt enable, and sets IMF to "1".

```
LDW     (EIR), 1100100010100001B    ;   EF15, EF14, EF11, EF7, EF5, IMF←1
```

Example 2 : Sets an individual interrupt enable flag to "1".

```
SET     (EIRH).4                    ;   EF12←1
```

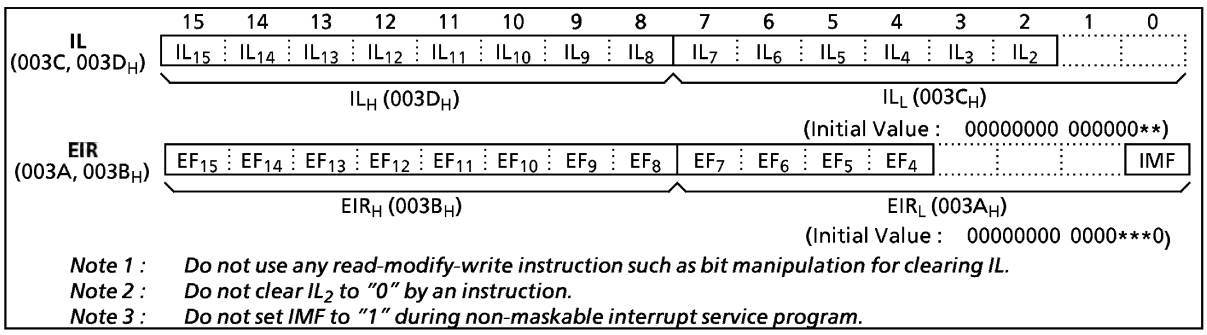


Figure 1-23. Interrupt Latch (IL) and Interrupt Enable Register (EIR)

1.9.1 Interrupt Sequence

An interrupt request is held until the interrupt is accepted or the interrupt latch is cleared to "0" by a reset or an instruction. Interrupt acceptance sequence requires 8 machine cycles (4 μs at fc = 8 MHz in the NORMAL mode) after the completion of the current instruction execution. The interrupt service task terminates upon execution of an interrupt return instruction [RETI] (for maskable interrupts) or [RETN] (for pseudo non-maskable interrupts).

(1) Interrupt acceptance processing

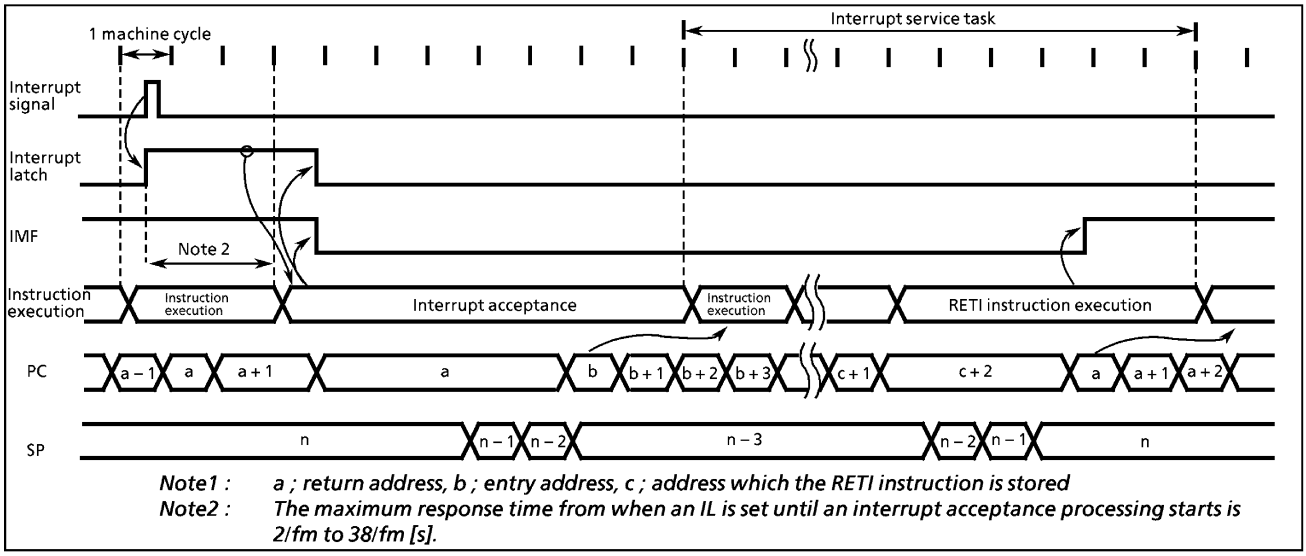
① The interrupt master enable flag (IMF) is cleared to "0" to temporarily disable the acceptance of any following maskable interrupts. When a non-maskable interrupt is accepted, the acceptance of any following interrupts is temporarily disabled.

② The interrupt latch (IL) for the interrupt source accepted is cleared to "0".

③ The contents of the program counter (return address) and the program status word are saved (pushed) on the stack. The stack pointer (SP) is three decrements.

④ The entry address of the interrupt service program is read from the vector table, and the entry address is loaded to the program counter.

⑤ The instruction stored at the entry address of the interrupt service program is executed.

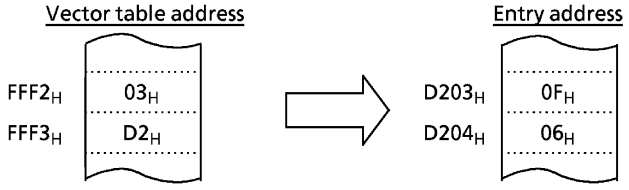


Note1 : a ; return address, b ; entry address, c ; address which the RETI instruction is stored

Note2 : The maximum response time from when an IL is set until an interrupt acceptance processing starts is 2/fm to 38/fm [s].

Figure 1-24. Timing Chart of Interrupt Acceptance and Interrupt Return Instruction

Example : Correspondence between vector table address for INTTBT and the entry address of the interrupt service program.



A maskable interrupt is not accepted until the IMF is set to "1" even if a maskable interrupt of higher priority than that of the current interrupt being serviced.

When nested interrupt service is necessary, the IMF is set to "1" in the interrupt service program. In this case, acceptable interrupt sources are selectively enabled by the individual interrupt enable flags. However, an acceptance of external interrupt 0 cannot be disabled by the EF; therefore, if disablement is necessary, either the external interrupt function of the $\overline{\text{INT0}}$ pin must be disabled with the INT0EN in the external interrupt control register (EINTCR) (When INT0EN = 0, the interrupt latch IL3 is not set, therefore, the falling edge of the INT0 pin input cannot be detected.) or interrupt processing must be avoided by the program.

Example 1 : Disables an external interrupt 0 using INT0EN

```
LD      (EINTCR), 00000000B    ; INT0EN←0
```

Example 2 : Disables the processing of external interrupt 0 under the software control (using bit 0 at address $00F0_H$ as the interrupt processing disable switch)

```
PINT0 :   TEST    (00F0H) . 0        ; Return without interrupt processing if (00F0H)0 = 1
          JRS     T, SINT0
          RETI
SINT0 :   Interrupt processing
          RETI
            :
VINT0 :   DW      PINT0
```

(2) General - purpose register save/restore processing

During interrupt acceptance processing, the program counter and the program status word are automatically saved on the stack, but not the accumulator and other registers. These registers are saved by the program if necessary. Also, when nesting multiple interrupt services, it is necessary to avoid using the same data memory area for saving registers.
The following method is used to save/restore the general-purpose registers:

① General-purpose register save/restore by register bank changeover:
The general-purpose registers can be saved at high-speed by switching to a register bank that is not in use. Normally, bank 0 is used for the main task and banks 1 to 15 are assigned to interrupt service tasks. To increase the efficiency of data memory utilization, the same bank is assigned for interrupt sources which are not nested.
The switched bank is automatically restored by executing an interrupt return instruction [RETI] or [RETN]. Therefore, it is not necessary for a program to save the RBS.

Example : Register Bank Changeover

```
PINTxx :  LD      RBS, n          ; Switches to bank n (1 μs at 8 MHz)
          interrupt processing
          RETI                    ; Restores bank and Returns
```

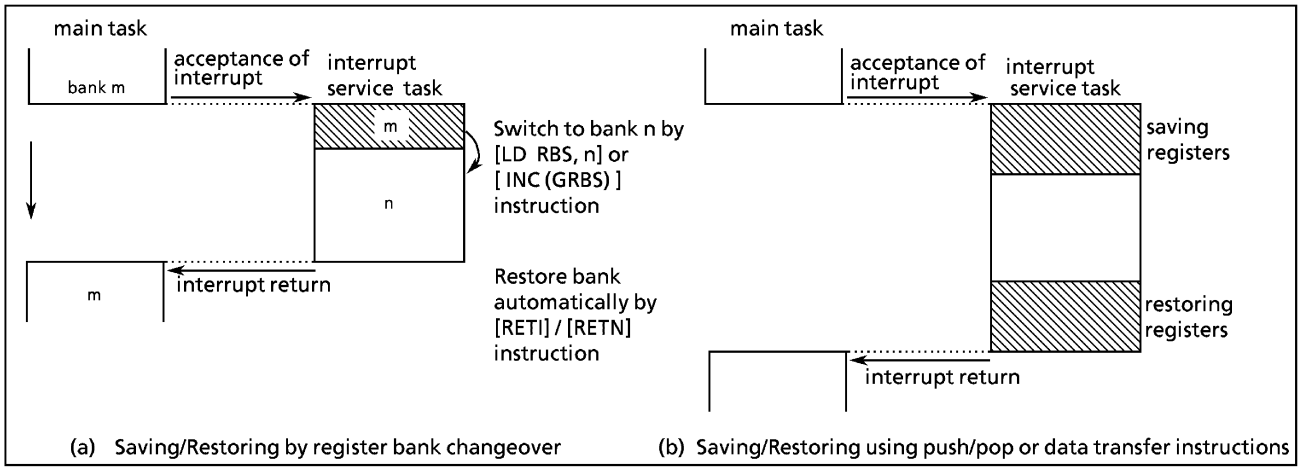
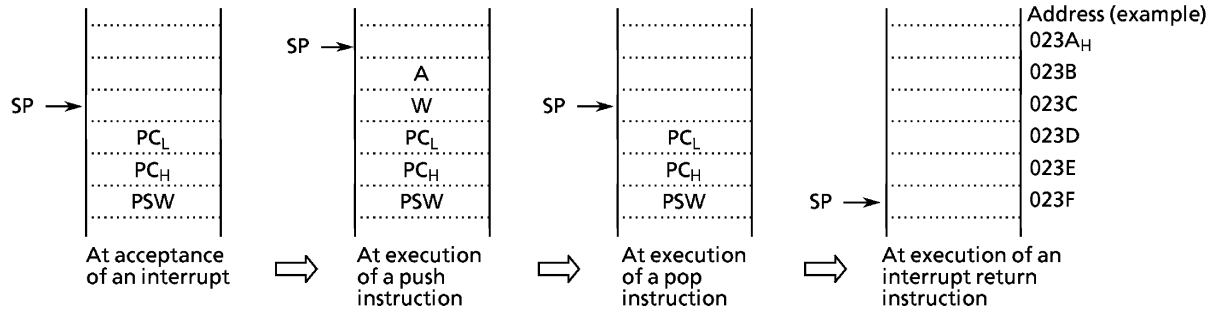


Figure 1-25. Saving/Restoring General-purpose Registers

② General-purpose register save/restore using push and pop instructions:
To save only a specific register, and when the same interrupt source occurs more than once, the general-purpose registers can be saved/restored using push/pop instructions.

Example : Register save using push and pop instructions

```
PINTxx :    PUSH    WA          ; Save WA register pair
            interrupt processing
            POP     WA          ; Restore WA register pair
            RETI                ; Return
```



③ General-purpose registers save/restore using data transfer instructions:
Data transfer instructions can be used to save only a specific general-purpose register during processing of a single interrupt.

Example : Saving/restoring a register using data transfer instructions

```
PINTxx :    LD      (GSAVA), A  ; Save A register
            interrupt processing
            LD      A, (GSAVA)  ; Restore A register
            RETI                ; Return from interrupt service
```

The interrupt return instructions [RETI] / [RETN] perform the following operations.

[RETI] Maskable interrupt return	[RETN] Non-maskable interrupt return
① The contents of the program counter and the program status word are restored from the stack.	① The contents of the program counter and program status word are restored from the stack.
② The stack pointer is incremented 3 times.	② The stack pointer is incremented 3 times.
③ The interrupt master enable flag is set to "1".	③ The interrupt master enable flag is set to "1" only when a non-maskable interrupt is accepted in interrupt enable status. However, the interrupt master enable flag remains at "0" when so clear by an interrupt service program.

Interrupt requests are sampled during the final cycle of the instruction being executed. Thus, the next interrupt can be accepted immediately after the interrupt return instruction is executed.

Note : When the interrupt processing time is longer than the interrupt request generation time, the interrupt service task is performed but not the main task.

1.9.2 External Interrupts

The 87C814/H14/K14/M14 each have five external interrupt inputs ($\overline{INT0}$, INT1, INT2, INT3, and $\overline{INT5}$). Three of these are equipped with digital noise rejection circuits (pulse inputs of less than a certain time are eliminated as noise). Edge selection is also possible with INT1, INT2 and INT3.

The $\overline{INT0}$/P10 pin can be configured as either an external interrupt input pin or an input/output port, and is configured as an input port during reset.

Edge selection, noise rejection control and $\overline{INT0}$/P10 pin function selection are performed by the external interrupt control register (EINTCR). When INT0EN = 0, the IL_3 will not be set even if the falling edge of $\overline{INT0}$ pin input is detected.

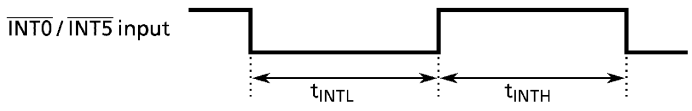
Table 1-3. External Interrupts

Source	Pin	Secondary function pin	Enable conditions	Edge	Digital noise rejection
INT0	$\overline{INT0}$	P10	IMF = 1, INT0EN = 1	falling edge	— (hysteresis input)
INT1	INT1	P11	IMF · EF_5 = 1	falling edge or rising edge	Pulses of less than 15/fc or 63/fc [s] are eliminated as noise. Pulses of 48/fc or 192/fc [s] or more are considered to be signals.
INT2	INT2	P12/TC1	IMF · EF_7 = 1		Pulses of less than 7/fc [s] are eliminated as noise. Pulses of 24/fc [s] or more are considered to be signals.
INT3	INT3	P86/S14	IMF · EF_{11} = 1		
INT5	$\overline{INT5}$	P20/$\overline{STOP}$	IMF · EF_{15} = 1	falling edge	— (hysteresis input)

Note 1 : The noise rejection function is turned off in SLOW and SLEEP modes. Also, the noise reject times are not constant for pulses input while transiting between operating modes (NORMAL2↔SLOW)

Note 2 : The noise rejection function is also affected for timer/counter input (TC1 and TC4 pins).

Note 3 : The pulse width (both "H" and "L" level) for input to the $\overline{INT0}$ and $\overline{INT5}$ pins must be over 1 machine cycle.



t_{INTL}, t_{INTH} > tcyc (*Note : tcyc = 4/fm [s]*)

Note 4 : If a noiseless signal is input to the external interrupt pin in the NORMAL 1/2 or IDLE 1/2 mode, the maximum time from the edge of input signal until the IL is set is as follows :

① INT1 pin 49/fc [s] (INT1NC = 1) , 193/fc [s] (INT1NC = 0)

② INT2,INT3 pins 25/fc [s]

Note 5 : When high-impedance is specified for port output in stop mode, port input is forcibly fixed to low level internally. Thus, interrupt latches of external interrupt inputs except INT5 (P20/STOP) which are also used as ports may be set to "1". To specify high-impedance for port output in stop mode, first disable interrupt service (IMF = 0) , activate stop mode. After releasing stop mode, clear interrupt latches using load instruction, then, enable interrupt service.

Example : Activating stop mode:

```
LD    (SYSCR1), 01000000B      ;  OUTEN←0 (specifies high-impedance)
DI                             ;  IMF←0 (disables interrupt service)
SET   (SYSCR1). STOP           ;  STOP←1 (activates stop mode)
LDW   (IL), 1111011101010111B  ;  IL11,7,5,3←0 (clears interrupt latches)
EI                             ;  IMF←1 (enables interrupt service)
```

EINTCR (0037_H)

7	6	5	4	3	2	1	0
INT1NC	INT0EN		(TC4ES)	INT3ES	INT2ES	INT1ES	

(Initial value : 00*0 000*)

INT1NC	Noise reject time select	0 : Pulses of less than 63/fc [s] are eliminated as noise 1 : Pulses of less than 15/fc [s] are eliminated as noise	R/W
INT0EN	P10/$\overline{INT0}$ pin configuration	0 : P10 input/output port 1 : $\overline{INT0}$ pin (Port P10 should be set to an input mode)	
INT3 ES INT2 ES INT1 ES	INT3 to INT1 edge select	0 : Rising edge 1 : Falling edge	

*Note 1 : fc ; High-frequency clock [Hz] * ; don't care*

Note 2 : Edge detection during switching edge selection is invalid.

Note 3 : Change EINTCR only when IMF = 0. After changing EINTCR, interrupt latches of external interrupt inputs must be cleared to "0" using load instruction.

Note 4 : In order to change of external interrupt input by rewriting the contents of INT2ES, INT3ES during NORMAL1/2 mode, clear interrupt latches of external interrupt inputs (INT2, INT3) after 8 machine cycles from the time of rewriting. During SLOW mode, 3 machine cycles are required.

*Note 5 : In order to change an edge of timer counter input by rewriting the contents of INT2ES, INT3ES during NORMAL1/2 mode, rewrite the contents after timer counter is stopped (TC*s = 0), that is, interrupt disable state. Then, clear interrupt laches of external interrupt inputs (INT2, INT3) after 8 machine cycles from the time of rewriting to change to interrupt enable state. Finally, state timer counter. During SLOW mode, 3 machine cycles are required.*

Example : When changing TC1 pin inputs edge in external trigger timer mode from rising edge to falling edge.

```
                LD  (TC1CR),01001000B     ;  TC1S ← 00 (stop TC1)
                DI                        ;  IMF ← 0 (disable interrupt service)
                LD  (EINTCR),00000100B    ;  INT2ES ← 1 (change edge selection)
                NOP
  8 machine
    cycles      NOP
                LD  (ILL),01111111B       ;  IL7 ← 0 (clear interrupt latch)
                EI                        ;  IMF ← 1 (enable interrupt service)
                LD  (TC1CR),01111000B     ;  TC1S ← 11 (start TC1)
```

Note 6 : If changing the contents of INT1ES during NORMAL1/2 mode, interrupt latch of external interrupt input INT1 must be cleared after 14 machine cycles (when INT1NC = 1) or 50 machine cycles (when INT1NC = 0) from the time of changing. During SLOW mode, 3 machine cycles are required.

Figure 1-26. External Interrupt Control Register

1.9.3 Software Interrupt (INTSW)

Executing the [SWI] instruction generates a software interrupt and immediately starts interrupt processing (INTSW is highest prioritized interrupt). However, if processing of a non-maskable interrupt is already underway, executing the SWI instruction will not generate a software interrupt but will result in the same operation as the [NOP] instruction. Thus, the [SWI] instruction behaves like the [NOP] instruction.

Note : Software interrupt generates during non-maskable interrupt processing to use SWI instruction for software break in a development tool.

Use the [SWI] instruction only for detection of the address error or for debugging.

① Address Error Detection

FF_H is read if for some cause such as noise the CPU attempts to fetch an instruction from a non-existent memory address. Code FF_H is the SWI instruction, so a software interrupt is generated and an address error is detected. The address error detection range can be further expanded by writing FF_H to unused areas of the program memory. the address trap reset is generated in case that an instruction is fetched from RAM or SFR areas.

Note : The fetch data from addresses $7F80_H$ to $7FFF_H$ (test ROM area) for 87C814/H14/K14/M14 is not "FF_H".

② Debugging

Debugging efficiency can be increased by placing the SWI instruction at the software break point setting address.

1.10 Watchdog Timer (WDT)

The watchdog timer rapidly detects the CPU malfunction such as endless looping caused by noise or the like, and resumes the CPU to the normal state.

The watchdog timer signal for detecting malfunction can be selected either as a reset output or a non-maskable interrupt request. However, selection is possible only once after reset. At first, the reset output is selected.

When the watchdog timer is not being used for malfunction detection, it can be used as a timer to generate an interrupt at fixed intervals.

1.10.1 Watchdog Timer Configuration

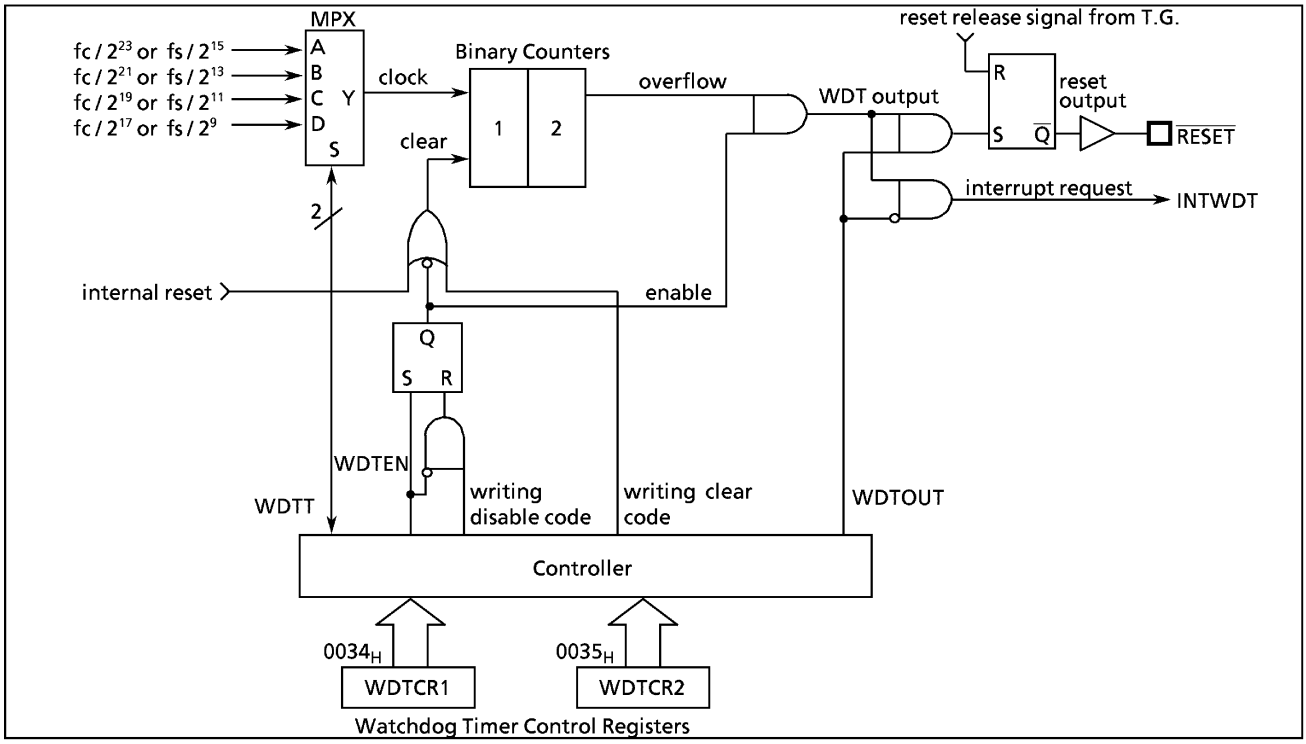


Figure 1-27. Watchdog Timer Configuration

1.10.2 Watchdog Timer Control

Figure 1-28 shows the watchdog timer control registers (WDTCR1, WDTCR2). The watchdog timer is automatically enabled after reset.

(1) Malfunction detection methods using the watchdog timer

The CPU malfunction is detected as follows:

① Setting the detection time, selecting output, and clearing the binary counter.

② Repeatedly clearing the binary counter within the setting detection time.

If a CPU malfunction occurs for any cause, the watchdog timer output will become active on the rise of an overflow from the binary counters unless the binary counters are cleared. At this time, when WDTOUT = 1 a reset is generated, which drives the $\overline{\text{RESET}}$ pin low to reset the internal hardware and the external circuits. When WDTOUT = 0, a watchdog timer interrupt (INTWDT) is generated.

The watchdog timer temporarily stops counting in STOP mode (including warm-up) or IDLE mode, and automatically restarts (continues counting) when STOP/IDLE mode is released.

Example : Sets the watchdog timer detection time to 2^{21}/fc [s] and resets the CPU malfunction.

```
                LD      (WDTCR2), 4EH          ; Clears the binary counters
                LD      (WDTCR1), 00001101B    ; WDTT←10, WDTOUT←1
Within 3/4 of   LD      (WDTCR2), 4EH          ; Clears the binary counters
WDT detection                                  (always clear immediately after changing WDTT)
time
                LD      (WDTCR2), 4EH          ; Clears the binary counters
Within 3/4 of
WDT detection
time
                LD      (WDTCR2), 4EH          ; Clears the binary counters
```

Watchdog Timer Control Register 1

WDTCR1 (0034_H)

7	6	5	4	3	2	1	0
				WDT EN	WDTT		WDT OUT

(Initial value : **** 1001)

WDTEN	Watchdog timer enable/disable	0 : Disable (It is necessary to write the disable code to WDTCR2) 1 : Enable	write only
WDTT	Watchdog timer detection time	00 : 2^{25} / fc or 2^{17} / fs [s] 01 : 2^{23} / fc or 2^{15} / fs 10 : 2^{21} / fc or 2^{13} / fs 11 : 2^{19} / fc or 2^{11} / fs	
WDTOUT	Watchdog timer output select	0 : Interrupt request 1 : Reset output	

Note 1 : WDTOUT cannot be set to "1" by program after clearing WDTOUT to "0".
*Note 2 : fc ; High-frequency clock [Hz] fs ; Low-frequency clock [Hz] * ; don't care*
Note 3 : WDTCR1 is a write-only register and must not be used with any of the read-modify-write instructions.
Note 4 : Disable the watchdog timer or clear the counter just before switching to STOP mode.
When the counter is cleared just before switching to STOP mode, clear the counter again subsequently to releasing STOP mode.

Watchdog Timer Control Register 2

WDTCR2 (0035_H)

7	6	5	4	3	2	1	0

(Initial value : **** ****)

WDTCR2	Watchdog timer control code write register	$4E_H$: Watchdog timer binary counter clear (clear code) $B1_H$: Watchdog timer disable (disable code) others : Invalid	write only

Note 1 : The disable code is invalid unless written when WDTEN = 0.
*Note 2 : * ; don't care*
Note 3 : Since WDTCR2 is a write-only register, read-modify-write instructions (e.g. ; bit manipulating instructions such as SET or CLR and arithmetic instructions such as AND or OR) cannot be used for read/write to this register.
Note 4 : To clear the binary counter doesn't initialize the source clock, therefore, it is recommended to clear the binary counter within 3/4 of the detection period.

Figure 1-28. Watchdog Timer Control Registers

Table 1-4. Watchdog Timer Detection Time

Operating mode			Detection time	
NORMAL1	NORMAL2	SLOW	At fc = 8 MHz	At fs = 32.768 kHz
2^{25} / fc [s]	2^{25} / fc, 2^{17} / fs	2^{17} / fs	4.194 s	4 s
2^{23} / fc	2^{23} / fc, 2^{15} / fs	2^{15} / fs	1.048 ms	1 s
2^{21} / fc	2^{21} / fc, 2^{13} / fs	——	262.1 ms	250 ms
2^{19} / fc	2^{19} / fc, 2^{11} / fs	——	65.5 ms	62.5 ms

(2) Watchdog Timer Enable

The watchdog timer is enabled by setting WDTEN (bit 3 in WDTCR1) to "1". WDTEN is initialized to "1" during reset, so the watchdog timer operates immediately after reset is released.

Example : Enables watchdog timer

```
LD      (WDTCR1), 00001000B     ; WDTEN←1
```

(3) Watchdog Timer Disable

The watchdog timer is disabled by writing the disable code ($B1_H$) to WDTCR2 after clearing WDTEN (bit 3 in WDTCR1) to "0". The watchdog timer is not disabled if this procedure is reversed and the disable code is written to WDTCR2 before WDTEN is cleared to "0". The watchdog timer is halted temporarily in STOP mode (including warm-up) and IDLE mode, and restarts automaticallyafter STOP or IDLE mode is released.

During disabling the watchdog timer, the binary counters are cleared.

Example : Disables watchdog timer

```
LDW     (WDTCR1) , 0B101H       ; WDTEN←0, WDTCR2←disable code
```

1.10.3 Watchdog Timer Interrupt (INTWDT)

This is a pseudo non-maskable interrupt which can be accepted regardless of the contents of the EIR. If a watchdog timer interrupt or a software interrupt is already accepted, however, the new watchdog timer interrupt waits until the previous interrupt processing is completed (the end of the [RETN] instruction execution).

The stack pointer (SP) should be initialized before using the watchdog timer output as an interrupt source with WDTOUT.

Example : Watchdog timer interrupt setting up.

```
LD      SP, 023FH               ; Sets the stack pointer
LD      (WDTCR1) , 00001000B    ; WDTOUT←0
```

1.10.4 Watchdog Timer Reset

If the watchdog timer output becomes active, a reset is generated, which drives the $\overline{\text{RESET}}$ pin (sink open drain output) low to reset the internal hardware and the external circuits. The reset output time is 2^{20}/fc [s] (131 ms at fc = 8 MHz). The high-fredquency clock oscillator also turns on when a watchdog timer reset is generated in SLOW mode.

Note : The high-frequency clock oscillator also turns on when a watchdog timer reset is generated in SLOW mode. Thus, the reset output time is 2^{20}/fc.
The reset output timer include a certain amount of error if there is any fluctuation of the oscillation frequency when the high-frequency clock oscillator turns on. Thus, the reset output time must be considered approximate value.

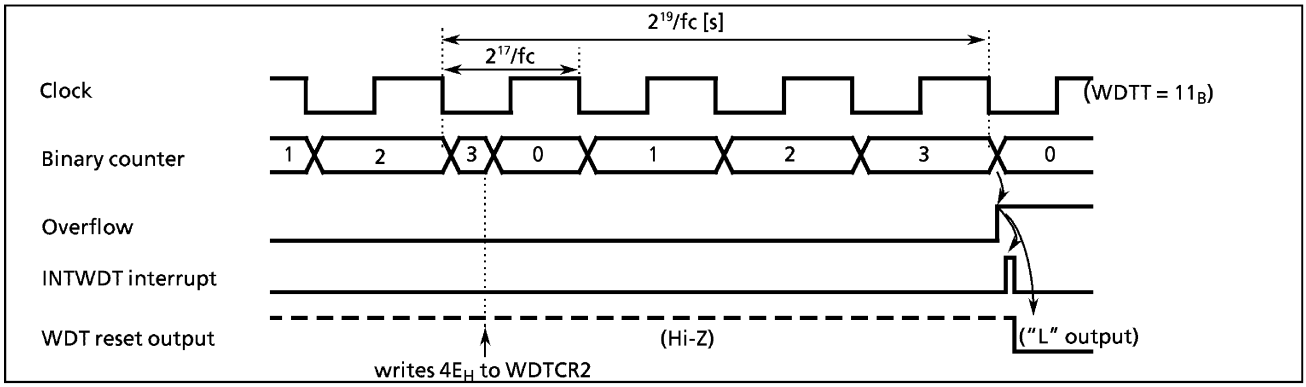


Figure 1-29. Watchdog Timer Interrupt / Reset

1.11 Reset Circuit

The 87C814/H14/K14/M14 each have four types of reset generation procedures: an external reset input, an address trap reset, a watchdog timer reset and a system clock reset. Table 1-5 shows on-chip hardware initialization by reset action. The internal source reset circuit (watchdog timer reset, address trap reset, and system clock reset) is not initialized when power is turned on. Thus, output from the $\overline{\text{RESET}}$ pin may go low (2^{20}/fc [s] (131 ms at 8 MHz) when power is turned on.

Table 1-5. Initializing Internal Status by Reset Action

On-chip Hardware		Initial Value	On-chip Hardware	Initial Value
Program counter	(PC)	$(FFFF_H) \cdot (FFFE_H)$	Divider of Timing generator	0
Register bank selector	(RBS)	0	Watchdog timer	Enable
Jump status flag	(JF)	1	Output latches of I/O ports	Refer to I/O port circuitry
Interrupt master enable flag	(IMF)	0	Control registers	Refer to each of control register
Interrupt individual enable flags	(EF)	0		
Interrupt latches	(IL)	0		

1.11.1 External Reset Input

When the $\overline{\text{RESET}}$ pin is held at low for at least 3 machine cycles (12/fc [s]) with the power supply voltage within the operating voltage range and oscillation stable, a reset is applied and the internal state is initialized.

When the $\overline{\text{RESET}}$ pin input goes high, the reset operation is released and the program execution starts at the vector address stored at addresses $FFFE_H$ - $FFFF_H$.

The $\overline{\text{RESET}}$ pin contains a Schmitt trigger (hysteresis) with an internal pull-up resistor. A simple power-on-reset can be applied by connecting an external capacitor and a diode.

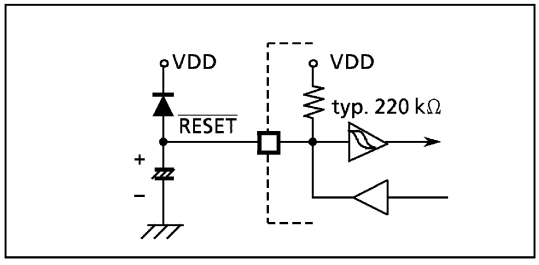


Figure 1-30. Simple Power-on-Reset Circuitry

1.11.2 Address Trap Reset

If a CPU malfunction occurs and an attempt is made to fetch an instruction from the RAM or the SFR area (addresses 0000_H - $023F_H$), an address-trap-reset will be generated. Then, the $\overline{RESET}$ pin output will go low. The reset time is 2^{20}/fc [s] (131 ms at 8 MHz).

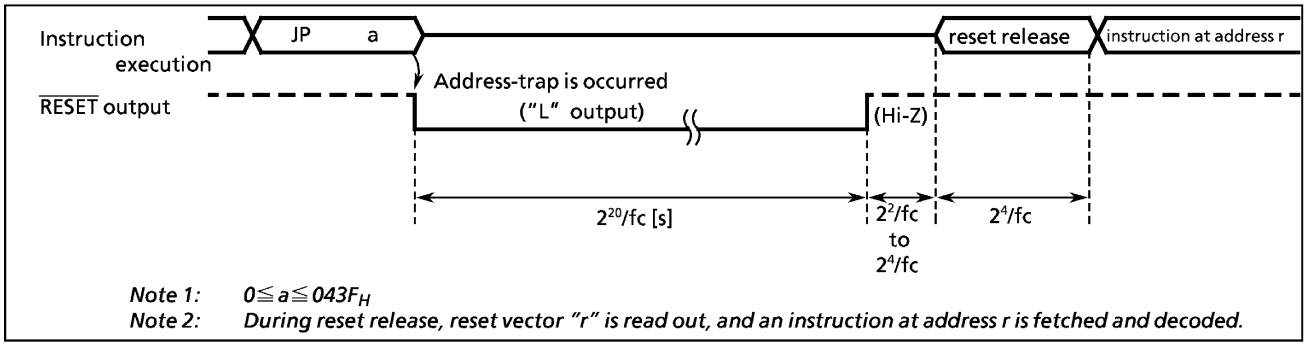


Figure 1-31. Address-Trap-Reset

1.11.3 Watchdog Timer Reset

Refer to Section "1.10 Watchdog Timer".

1.11.4 System-Clock-Reset

Clearing both XEN and XTEN (bits 7 and 6 in SYSCR2) to "0" stops both high-frequency and low-frequency oscillation, and causes the MCU to deadlock. This can be prevented by automatically generating a reset signal whenever XEN = XTEN = 0 is detected to continue the oscillation. Then, the $\overline{RESET}$ pin output goes low from high-impedance. The reset time is 2^{20}/fc [s] (131 ms at 8 MHz).

2. ON-CHIP PERIPHERALS FUNCTIONS

2.1 Special Function Registers (SFR) and Data Buffer Registers (DBR)

The TLCS-870 Series uses the memory mapped I/O system, and all peripheral control and data transfers are performed through the special function registers (SFR) and data buffer registers (DBR).

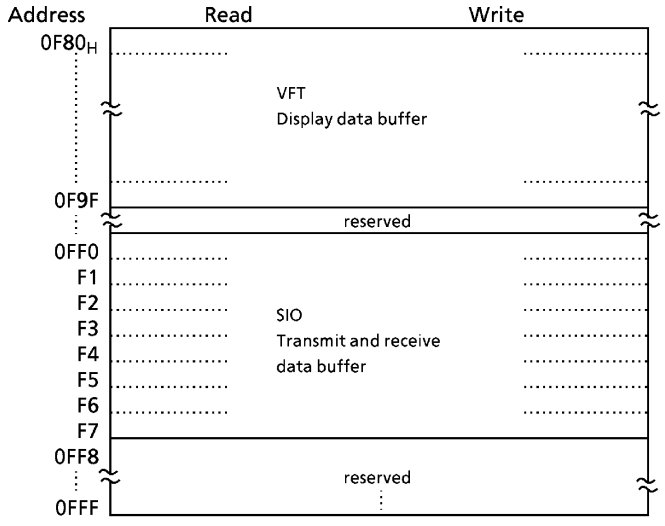
The SFR are mapped to addresses 0000_H – $003F_H$, and the DBR to addresses $0F80_H$ – $0FFF_H$.

Figure 2-1 shows the 87C814/H14/K14/M14 SFRs and DBRs.

Address	Read	Write
0000_H	P0 port	
01	P1 Port	
02	P2 Port	
03	P3 Port	
04	reserved	
05	P5 Port	
06	P6 Port	
07	P7 Port	
08	P8 Port	
09	reserved	
0A	–	P0CR (P0 I/O control)
0B	–	P1CR (P1 I/O control)
0C	–	P6CR (P6 I/O control)
0D	reserved	
0E		ADCCR (A/D converter control)
0F	ADCDR (A/D conv. Result)	–
10	–	$TREG1A_L$ (Timer register 1A)
11	–	$TREG1A_H$
12	$TREG1B_L$ (Timer register 1B)	
13	$TREG1B_H$	
14	–	TC1CR (TC1 control)
15	–	TC2CR (TC2 control)
16	–	$TREG2_L$ (Timer register 2)
17	–	$TREG2_H$
18	TREG3A (Timer register 3A)	
19	reserved	
1A	–	TC3CR (TC3 control)
1B	–	TREG4 (Timer register 4)
1C	–	TC4CR (TC4 control)
1D	reserved	
1E	reserved	
1F	reserved	

Address	Read	Write
0020_H	SIOSR (SIO status)	SIOCR1 (SIO control)
21	–	SIOCR2
22	reserved	
23	reserved	
24	reserved	
25	PWMSR (PWM Status)	PWMCR (PWM14 control)
26	–	PWMDBR
27	–	DVObCR (DVOb control)
28	reserved	
29	VFTSR (VFT status)	VFTCR1 (VFT control 1)
2A	–	VFTCR2 (VFT control 2)
2B		VFTCR3 (VFT control 3)
2C	reserved	
2D	reserved	
2E	reserved	
2F	reserved	
30	reserved	
31	reserved	
32	reserved	
33	reserved	
34	–	WDTCR1 (WDT control)
35	–	WDTCR2
36	TBTCR (TBT / TG / DVOa control)	
37	EINTCR (Interrupt control)	
38	SYSCR1 (System control)	
39	SYSCR2	
3A	EIR_L (Interrupt enable register)	
3B	EIR_H	
3C	IL_L (Interrupt latch)	
3D	IL_H	
3E	reserved	
3F	PSW (Program status word)	RBS (Register bank selector)

(a) Special Function Registers



(b) Data Buffer Registers

Note 1 : Do not access reserved areas by the program.

Note 2 : – : Cannot be accessed.

Note 3 : When defining address $003F_H$ with assembler symbols, use GPSW and GRBS.

Note 4 : Write-only registers and interrupt latches cannot use the read-modify-write instructions (bit manipulation instructions such as SET, CLR, etc. and logical operation instructions such as AND, OR, etc.)

Figure 2-1. SFR & DBR

2.2 I/O Ports

The 87C814/H14/K14/M14 each have 8 parallel input/output ports (55pins) as follows:

	Primary Function	Secondary Functions
Port P0	8-bit I/O port	———
Port P1	8-bit I/O port	External interrupt input, timer/counter input, and divider output
Port P2	3-bit I/O port	Low-frequency resonator connections, external interrupt input, and STOP mode release signal input
Port P3	4-bit I/O port	Serial interface, pulse width modulation output, PWM/PDO output, and timer/counter input
Port P5	8-bit I/O port	VFT digit driver output
Port P6	8-bit I/O port	Analog input
Port P7	8-bit I/O port	VFT digit/segment driver output
Port P8	8-bit I/O port	VFT segment driver/divider output, and external interrut input

Ports P1, P2, P3, P5, P6, P7 and P8 can also use secondary function.

Each output port contains a latch, which holds the output data. Input ports do not have latches, so the external input data should either be held externally until read or reading should be performed several times before processing. Figure 2-2 shows input/output timing examples.

External data is read from an I/O port in the S1 state of the read cycle during execution of the read instruction. This timing can not be recognized from outside, so that transient input such as chattering must be processed by the program.

Output data changes in the S2 state of the write cycle during execution of the instruction which writes to an I/O port.

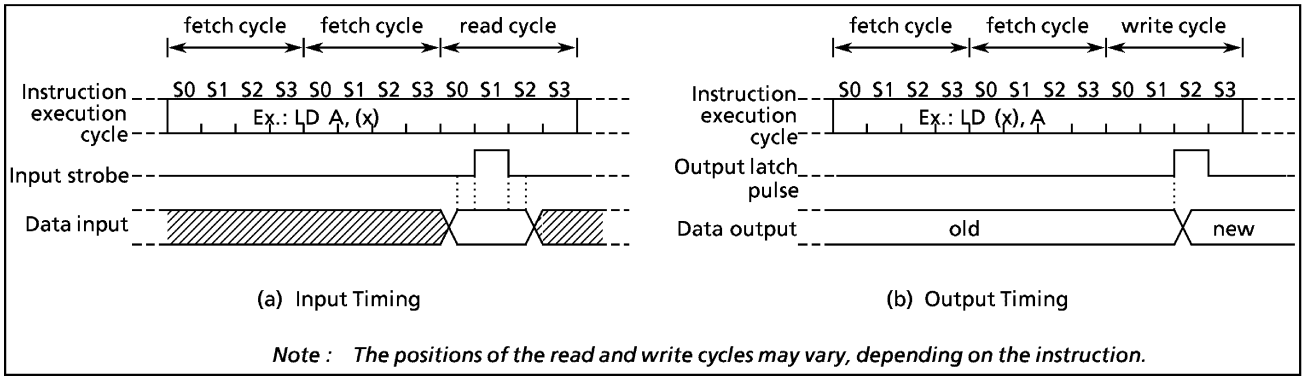


Figure 2-2. Input/Output Timing (Example)

When reading an I/O port except programmable I/O ports P0 and P1, whether the pin input data or the output latch contents are read depends on the instructions, as shown below:

(1) Instructions that read the output latch contents

① XCH r, (src)
② CLR/SET/CPL (src).b
③ CLR/SET/CPL (pp).g
④ LD (src).b, CF
⑤ LD (pp) . b, CF
⑥ ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), n
⑦ (src) side of ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), (HL)

(2) Instructions that read the pin input data

① Instructions other than the above (1)
② (HL) side of ADD/ADDC/SUB/SUBB/AND/OR/XOR (src), (HL)

2.2.1 Port P0 (P07 - P00)

Port P0 is an 8-bit general-purpose input/output port which can be configured as either an input or an output in one-bit unit under software control. Input/output mode is specified by the corresponding bit in the port P0 input/output control register (P0CR). Port P0 is configured as an input if its corresponding P0CR bit is cleared to "0", and as an output if its corresponding P0CR bit is set to "1".

During reset, P0CR is initialized to "0", which configures port P0 as input. The P0 output latches are also initialized to "0". Data is written into the output latch regardless of P0CR contents. Therefore initial output data should be written into the output latch before setting P0CR.

Note1 : Ports set to the input mode read the pin states. When input pin and output pin exist port P0 together, the contents of the output latch of ports set to the input mode may be rewritten by executing the bit manipulation instructions. Pins set to the output mode read a value of the output latch.

Note2 : The P0CR is a write-only register. It can not be operated by the read-modify instruction (Bit manipulation instructions of SET, CLR, etc. and Arithmetic instructions of AND, OR, etc.)

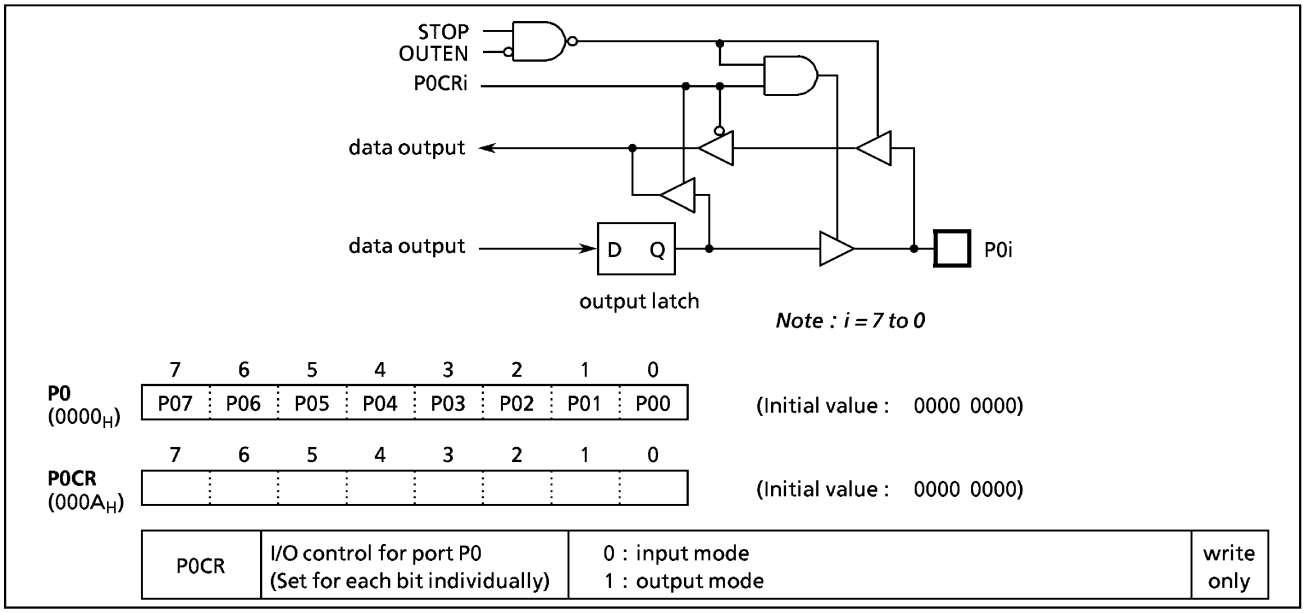


Figure 2-3. Port P0 and P0CR

Example : Setting the upper 4 bits of port P0 as an input port and the lower 4 bits as an output port (Initial output data are 1010_B).

```
LD      (P0), 00001010B     ; Sets initial data to P0 output latches
LD      (P0CR), 00001111B   ; Sets the port P0 input/output mode
```

2.2.2 Port P1 (P17 - P10)

Port P1 is an 8-bit input/output port which can be configured as an input or an output in one-bit unit under software control. Input/output mode is specified by the corresponding bit in the port P1 input/output control register (P1CR). Port P1 is configured as an input if its corresponding P1CR bit is cleared to "0", and as an output if its corresponding P1CR bit is set to "1". During reset, P1CR is initialized to "0", which configures port P1 as an input . The P1 output latches are also initialized to "0". Data is written into the output latch regardless of P1CR contents. Therfore initial output data should be written into the output latch before setting P1CR. Port P1 is also used as an external interrupt input, a timer/counter input, and a divider output. When used as a secondary function pin, the input pins should be set to the input mode, and the output pins should be set to the output mode and beforehand the output latch should be set to "1".

It is recommended that pins P11 and P12 should be used as external interrupt inputs, timer/counter input, or input ports. The interrupt latch is set on the rising or falling edge of the output when used as output ports.

Pin P10 ($\overline{INT0}$) can be configured as either an I/O port or an external interrupt input with INT0EN (bit 6 in EINTCR). During reset, the pin P10 ($\overline{INT0}$) is configured as an input port P10.

Note1 : Ports set to the input mode read the pin states. When input pin and output pin exist port P1 together, the contents of the output latch of ports set to the input mode may be rewritten by executing the bit manipulation instructions. Pins set to the output mode read a value of the output latch.

Note2 : The P1CR is a write-only register. It can not be operated by the read-modify instruction (Bit manipulation instructions of SET, CLR, etc. and Arithmetic instructions of AND, OR, etc.)

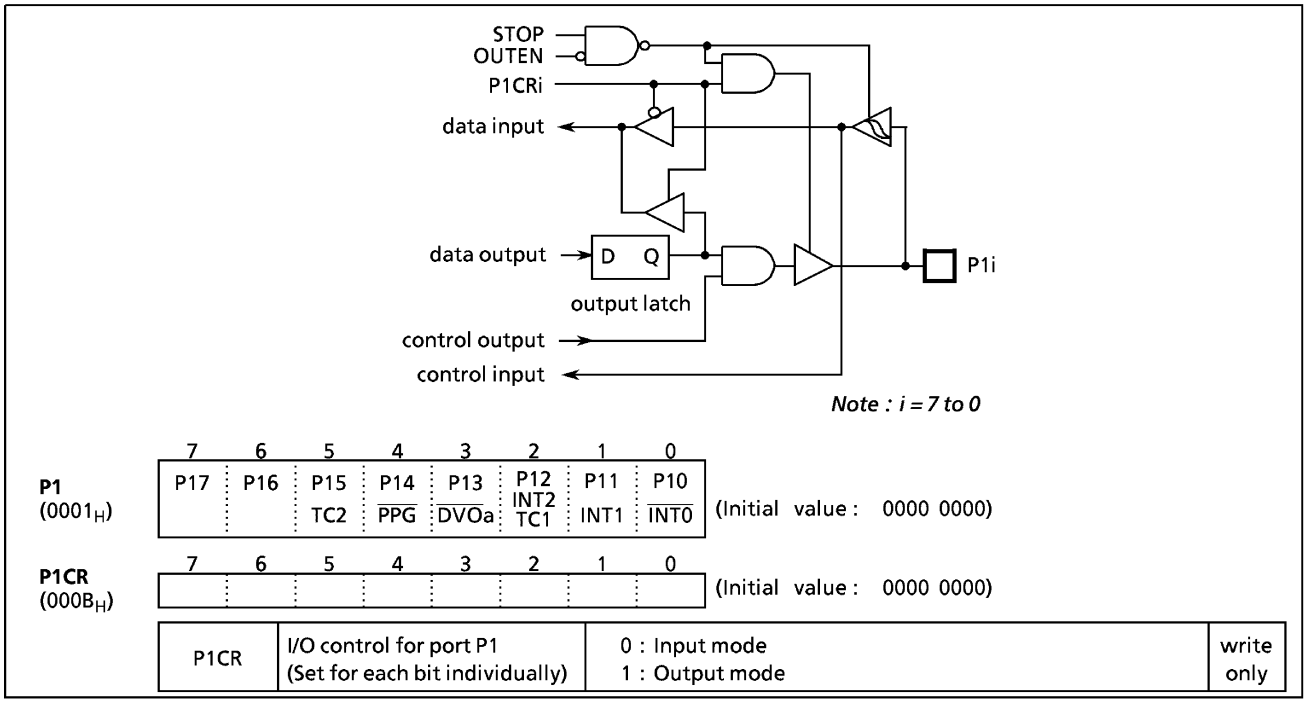


Figure 2-4. Port P1 and P1CR

Example : Sets P17, P16 and P14 as output ports, P13 and P11 as input ports, and the others as function pins. Internal output data is "1" for the P17 and P14 pins, and "0" for the P16 pin.

```
LD      (EINTCR), 01000000B     ; INT0EN←1
LD      (P1), 10111111B         ; P17←1, P14←1, P16←0
LD      (P1CR), 11010000B
```

2.2.3 Port P2 (P22 - P20)

Port P2 is a 3-bit input/output port. It is also used as an external interrupt input, and low-frequency crystal connection pins When used as an input port, or the secondary function pin, the output latch should be set to "1". During reset , the output latches are initialized to "1".

A low-frequency crystal (32.768 kHz) is connected to pins P21 (XTIN) and P22 (XTOUT) in the dual-clock mode. In the single-clock mode, pins P21 and P22 can be used as normal input/output ports.

It is recommended that the P20 pin should be used as an external interrupt input, a STOP mode release signal input, or an input port. If used as an output port, the interrupt latch is set on the falling edge of the output pulse.

When a read instruction for port P2 is executed, bits 7 to 3 in P2 read in as undefined data.

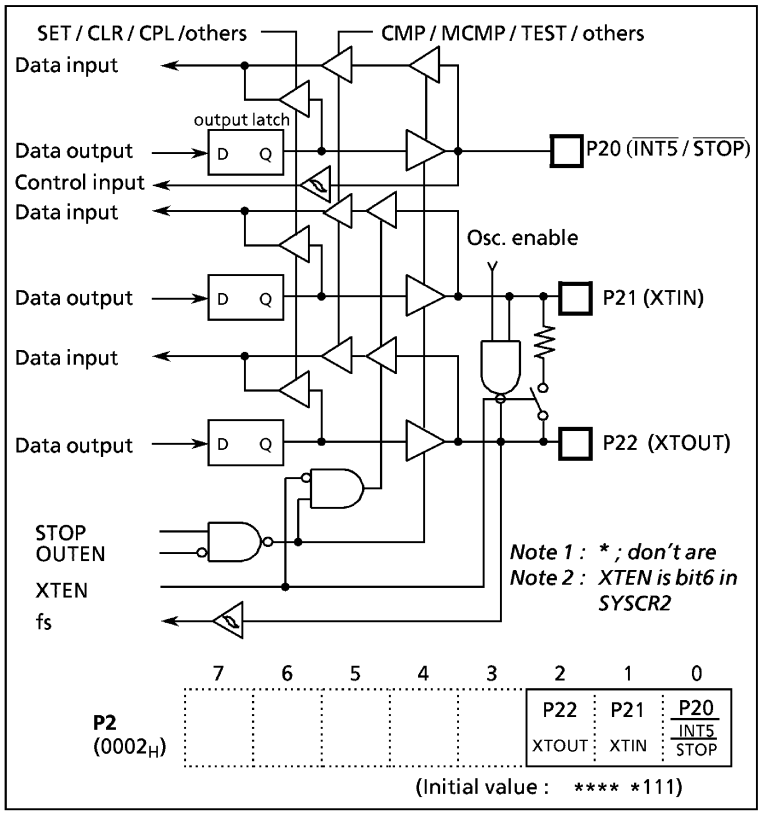


Figure 2-5. Port P2

2.2.4 Port P3 (P33 - P30)

Port P3 is an 4-bit input/output port, and is also used as serial interface (SIO) input/output, 14 built-in pulse width modulation (PWM) and a timer/counter input. When used as an input port or a secondary function pin, the output latch should be set to "1". The output latches are initialized to "1" during reset.

It is recommended that the P33 pin should be used as an external interrupt input, a timer/counter input, or an input port.

When a read instruction for port P3 is executed bit 7 to 4 in P3 read in as undefined data.

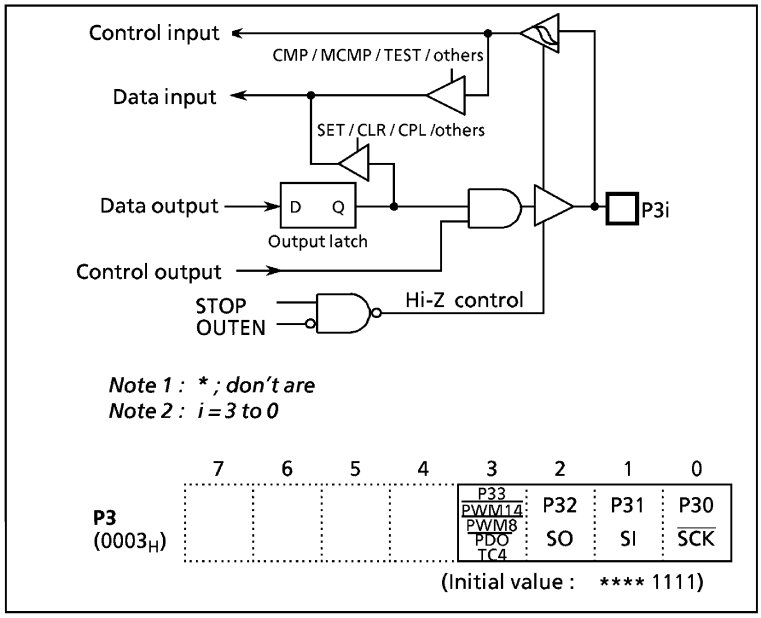


Figure 2-6. Port P3

2.2.5 Port P5 (P57 - P50)

Port P5 are 8-bit high-breakdown voltage input/output ports, and are also used as digit outputs, which can directly drive vacuum fluorescent tube (VFT). When used as an input port or a digit output, the output latch should be cleared to "0". The output latches are initialized to "0" during reset. Pins which are not set for digit output can be used as normal I/O port (refer to section "2.11.8 Port Function"). It is recommended that pins P57 to P50 should be used as digit output.

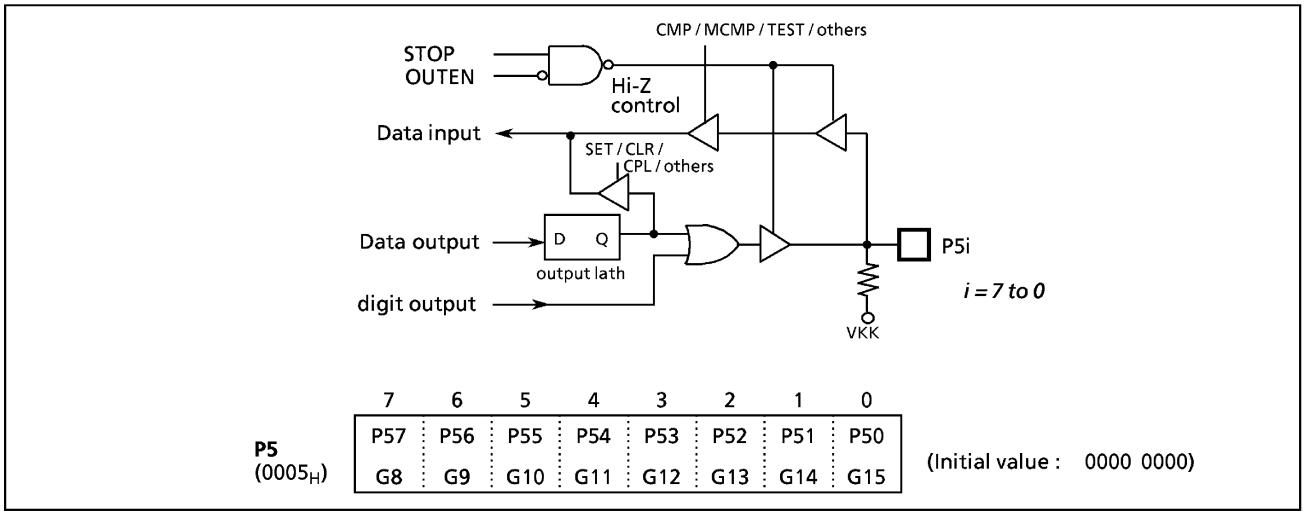


Figure 2-8. P5 Port

2.2.6 Port P6 (P67 - P60)

Ports P6 is an 8-bit input/output port which can be configured as an input or an output in one-bit unit under software control. Input/output mode is specified by the corresponding bit in the port P6 input/output control register (P6CR).

Port P6 is also used as an analog input for the A/D converter. When used as an analog input, AINDS (bit 4 in the ADCCR) must be cleared to "0" and its corresponding P6CR bit must be set to "1". In this case, unuse pin as analog input is configured as only input port.
During reset, AINDS is initialized to "0" and all bits of P6CR are initialized to "1", which configures port P6 as analog input. The P6 output latches are initialized to "0". Data is written into the output latch regardless of the P6CR contents. Therefore initial output data should be written into the output latch before setting P6CR.

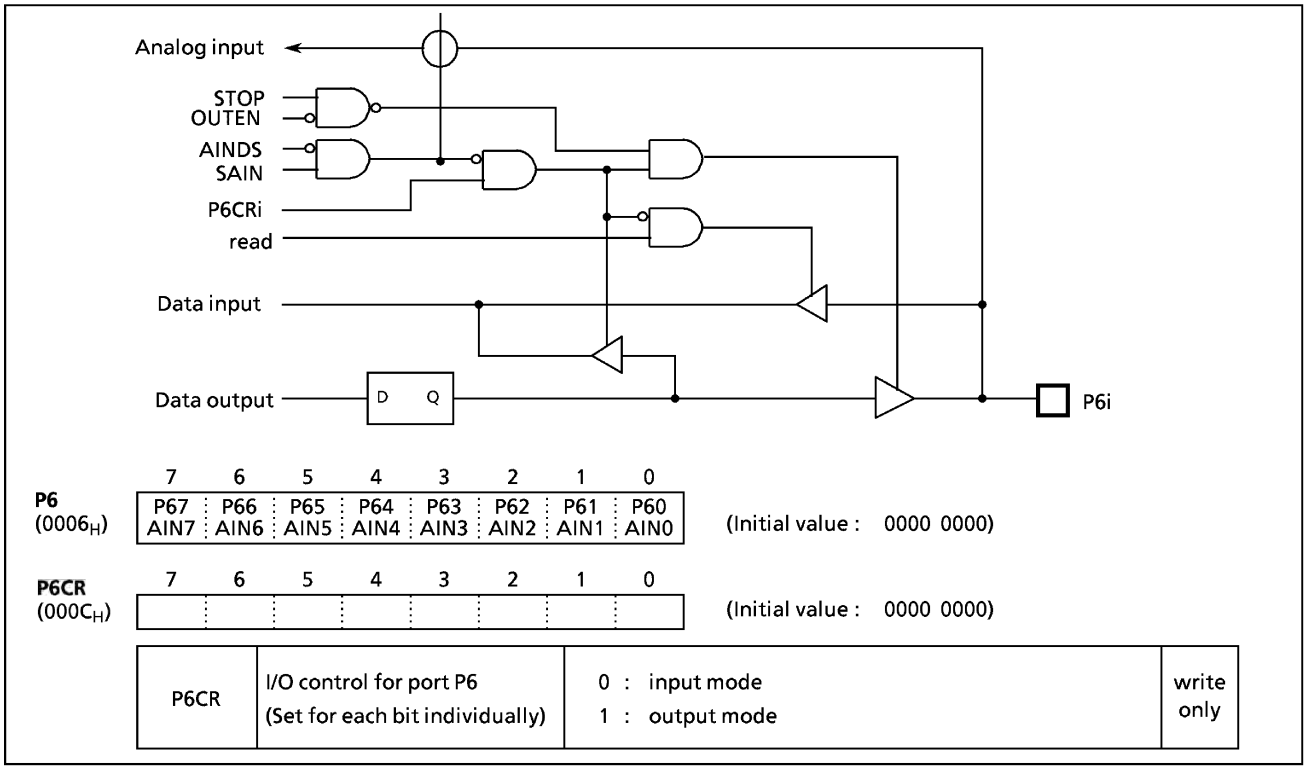
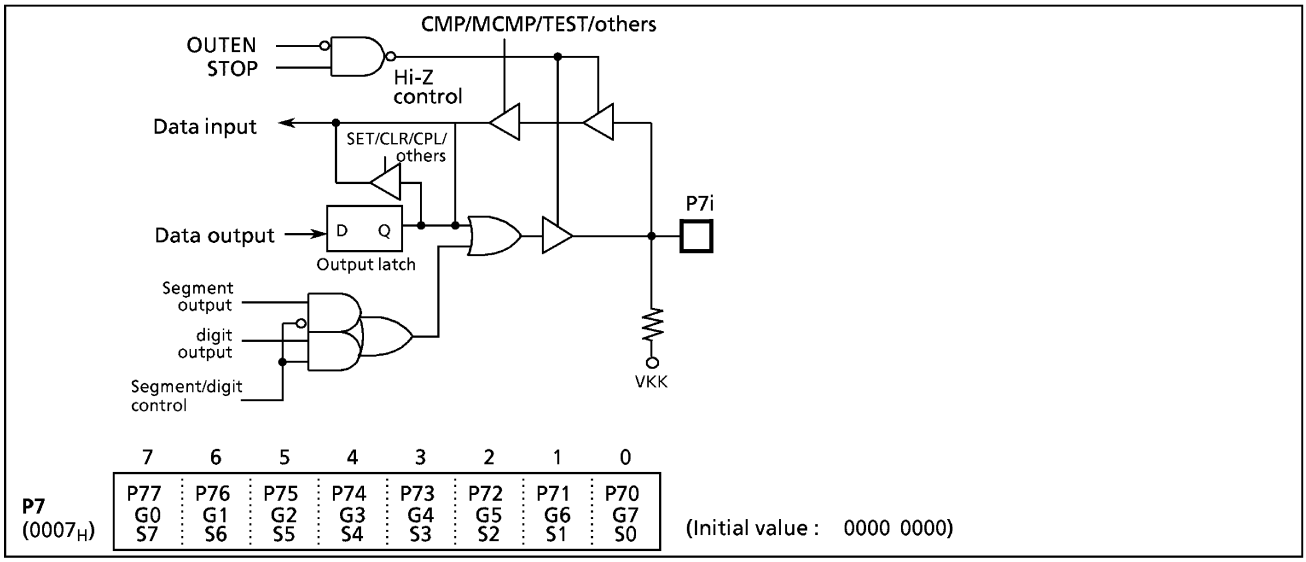


Figure 2-9. Port P6

2.2.7 Port P7 (P77 - P70)

Port P7 is an 8-bit high-breakdown voltage input / output port, and also used as a segment / digit output, which can directly drive vacuum fluorescent tube (VFT). When used as an input port or a segment / digit output, the output latch should be cleared to "0". The output latches are initialized to "0" during reset. It is recommended that pins P77 to P70 should be used as digit/segment output.



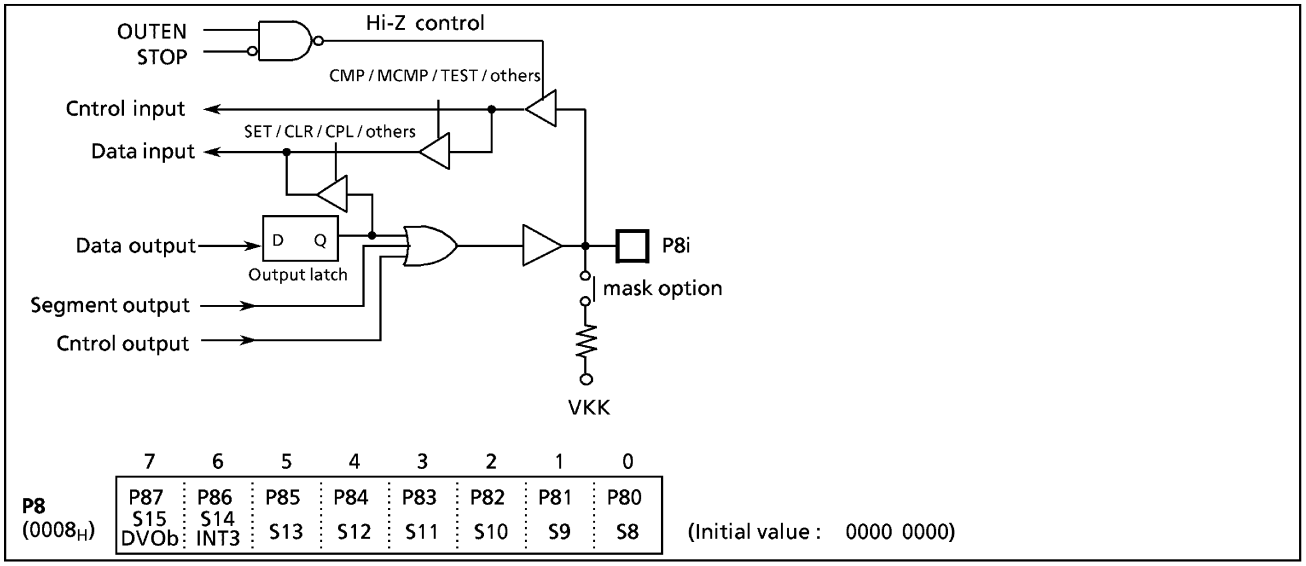
	7	6	5	4	3	2	1	0
P7 (0007_H)	P77 G0 S7	P76 G1 S6	P75 G2 S5	P74 G3 S4	P73 G4 S3	P72 G5 S2	P71 G6 S1	P70 G7 S0

(Initial value : 0000 0000)

Figure 2-10. Port P7

2.2.8 Port P8 (P87 - P80)

Port P8 is an 8-bit high-breakdown voltage input/output port, a divider output, an external interrupt input, and also used as a segment output, which can directly drive vacuum fluorescent tube (VFT). When used as an input port, a segment output or secondary function pin, the output latch should be cleared to "0". The output latches are initialized to "0" during reset. Pins P87 to P84 can be connected built-in pull-down resistors by mask option. It is recommended that pins witch have a built-in pull-down resistors should be used segment output.



	7	6	5	4	3	2	1	0
P8 (0008_H)	P87 S15 DVOb	P86 S14 INT3	P85 S13	P84 S12	P83 S11	P82 S10	P81 S9	P80 S8

(Initial value : 0000 0000)

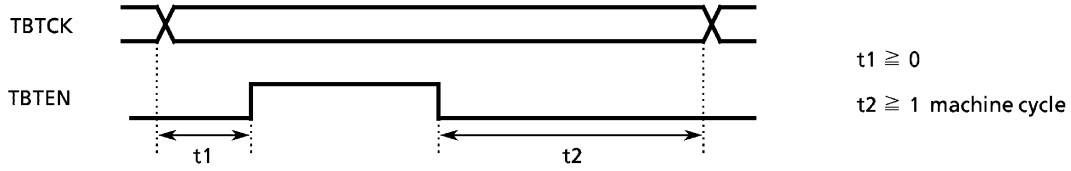
Figure 2-11. Port P8

2.3 Time Base Timer (TBT)

The time-base timer is used to generate the base time for key scan and dynamic display processing. For this purpose, it generates a time-base timer interrupt (INTTBT) at predetermined intervals.

This interrupt is generated beginning with the first rising edge of the source clock (the timing generator's divider output selected by TBTCK) after the time-base timer is enabled. Note that since the divider cannot be cleared by a program, the first interrupt only may occur earlier than the set interrupt period. (See Figure 2-10 (b).)

When selecting the interrupt frequency, make sure the time-base timer is disabled. (Do not change the selected interrupt frequency when disabling the active timer either.) However, you can select the interrupt frequency simultaneously when enabling the timer.



Example : Sets the time base timer frequency to fc/2^{16} [Hz] and enables an INTTBT interrupt.

```
LD      (TBTCR) , 00001010B
SET     (EIRL) . 6
```

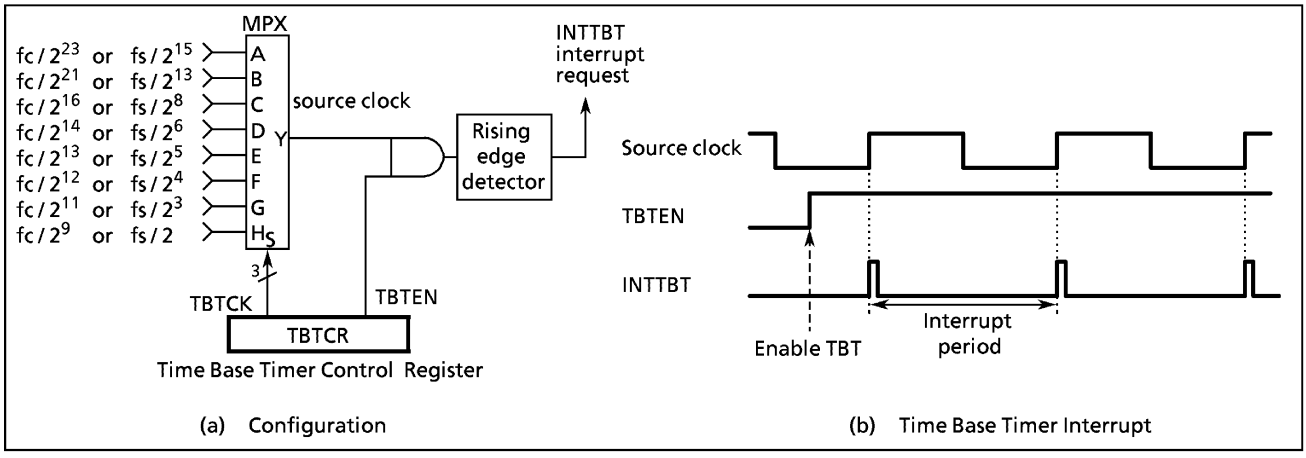


Figure 2-14. Time Base Timer

TBTCR (0036_H)

7	6	5	4	3	2	1	0
(DVOEN)	(DVOCK)		(DV7CK)	TBTEN	TBTCK		

(Initial value : 0**0 0***)

TBTEN	Time base timer enable/disable	0 : Disable 1 : Enable	R/W
TBTCK	Time base timer interrupt frequency select	000 : $fc/2^{23}$ or $fs/2^{15}$ [Hz] 001 : $fc/2^{21}$ or $fs/2^{13}$ 010 : $fc/2^{16}$ or $fs/2^{8}$ 011 : $fc/2^{14}$ or $fs/2^{6}$ 100 : $fc/2^{13}$ or $fs/2^{5}$ 101 : $fc/2^{12}$ or $fs/2^{4}$ 110 : $fc/2^{11}$ or $fs/2^{3}$ 111 : $fc/2^{9}$ or $fs/2$	

*Note : fc ; High-frequency clock [Hz], fs ; Low-frequency clock [Hz], * ; don't care*

Figure 2-15. Time Base Timer and Divider Output Control Register

Table 2-1. Time Base Timer Interrupt Frequency

TBTCK	NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode	Interrupt Frequency	
	DV7CK = 0	DV7CK = 1		At fc = 8 MHz	At fs = 32.768 kHz
000	$fc/2^{23}$	$fs/2^{15}$	$fs/2^{15}$	0.95 Hz	1 Hz
001	$fc/2^{21}$	$fs/2^{13}$	$fs/2^{13}$	3.81	4
010	$fc/2^{16}$	$fs/2^{8}$	–	122.07	128
011	$fc/2^{14}$	$fs/2^{6}$	–	488.28	512
100	$fc/2^{13}$	$fs/2^{5}$	–	976.56	1024
101	$fc/2^{12}$	$fs/2^{4}$	–	1953.12	2048
110	$fc/2^{11}$	$fs/2^{3}$	–	3906.25	4096
111	$fc/2^{9}$	$fs/2$	–	15625	16384

2.4 Divider Output ($\overline{DVO}$a, DVOb)

A 50% duty pulse can be output using the divider output circuit, which is useful for piezo-electric buzzer drive. Divider output is from pin P13 ($\overline{DVO}$a) and P87 (DVOb). The P13 output latch should be set to "1" and then the P13 should be configured as an output mode, and the P87 output latch clear to "0". Divider output circuit is controlled by the control register (TBTCR) and DVObCR shown in Figure 2-12.

$\overline{DVO}$a control register

TBTCR (0036_H)

7	6	5	4	3	2	1	0
DVOEN	DVOCK		(DV7CK)	(TBTEN)	(TBTCK)		

(Initial value : 0**0 0***)

DVOEN	Divider output enable/disable	0 : Disable 1 : Enable	R/W
DVOCK	Divider output ($\overline{DVO}$a) frequency selection	00 : $fc/2^{13}$ or $fs/2^5$ [Hz] 01 : $fc/2^{12}$ or $fs/2^4$ 10 : $fc/2^{11}$ or $fs/2^3$ 11 : $fc/2^{10}$ or $fs/2^2$	

DVOb control register

DVObCR (0027_H)

7	6	5	4	3	2	1	0
DVObEN	DVObCK						

(Initial value 000* ***)

DVObEN	Divider output enable/disable	0 : Disable 1 : Enable	Write only
DVObCK	Divider output (DVOb) frequency selection	00 : $fc/2^{13}$ or $fs/2^5$ [Hz] 01 : $fc/2^{12}$ or $fs/2^4$ 10 : $fc/2^{11}$ or $fs/2^3$ 11 : $fc/2^{10}$ or $fs/2^2$	

*Note : fc ; High-frequency clock [Hz], fs ; Low-frequency clock [Hz], * ; don't care*

Figure 2-16. Divider Output Control Register

Example : 1 kHz pulse output to P13 (at fc = 8 MHz)

```
SET     (P1).3                  ; P13 output latch ←1
LD      (P1CR), 00001000B       ; Configures P13 as an output mode
LD      (TBTCR), 10000000B      ; DVOEN←1, DVOCK←00
```

Table 2-2. Frequency of Divider Output

DVOCK (DVObCK)	Frequency of Divider Output	At fc = 8 MHz	At fs = 32.768 kHz
00	$fc/2^{13}$ or $fs/2^5$	0.976 [kHz]	1.024 [kHz]
01	$fc/2^{12}$ $fs/2^4$	1.953	2.048
10	$fc/2^{11}$ $fs/2^3$	3.906	4.096
11	$fc/2^{10}$ $fs/2^2$	7.812	8.192

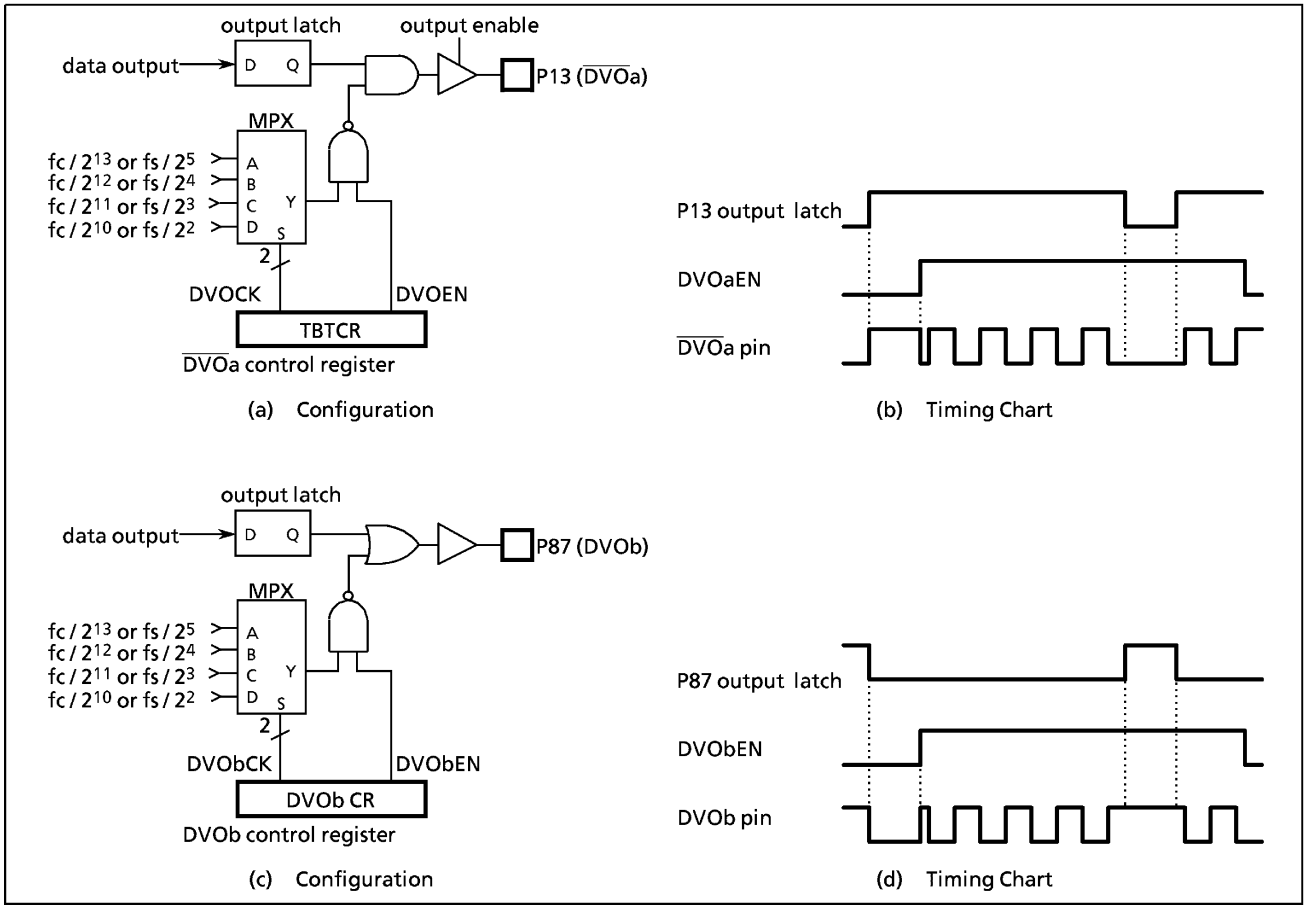


Figure 2-17. Divider Output

2.5 16-bit Timer/Counter 1 (TC1)

2.5.1 Configuration

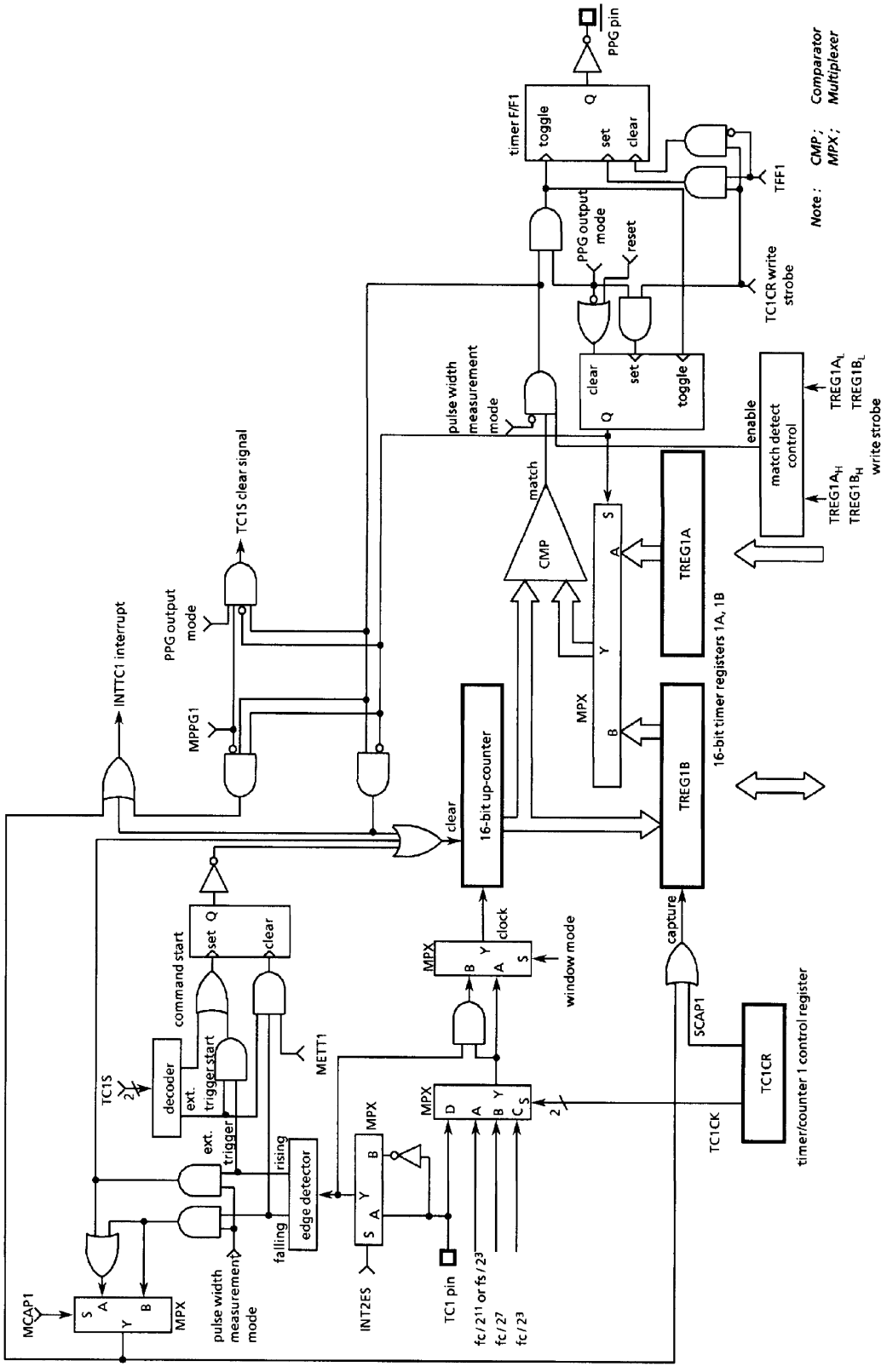


Figure 2-18. Timer/Counter 1

2.5.2 Control

The timer/counter 1 is controlled by a timer/counter 1 control register (TC1CR) and two 16-bit timer registers (TREG1A and TREG1B). Reset does not affect TREG1A and TREG1B.

TREG1A (0010, 0011_H)

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
$TREG1A_H$ (0011_H)								$TREG1A_L$ (0010_H)							

Write only

TREG1B (0012, 0013_H)

$TREG1B_H$ (0013_H)	$TREG1B_L$ (0012_H)

Read / Write (Write available in only PPG output mode)

TC1CR (0014_H)

7	6	5	4	3	2	1	0
TFF1	SCAP1 MCAP1 METT1 MPPG1	TC1S		TC1CK		TC1M	

(Initial value : 0000 0000)

TC1M	TC1 mode select	00 : timer / external trigger timer / event counter mode 01 : window mode 10 : pulse width measurement mode 11 : PPG output mode	Write only
TC1CK	TC1 source clock select	00 : internal clock fc/2^{11} or fs/2^3 [Hz] 01 : internal clock fc/2^7 10 : internal clock fc/2^3 11 : external clock (TC1 pin input)	
TC1S	TC1 start control	00 : stop & counter clear 01 : command start 10 : reserved 11 : external trigger start	
SCAP1	software capture control	0 : – 1 : software capture trigger (Note 3)	
MCAP1	pulse width measurement control	1 : double edge capture 1 : single edge capture	
METT1	external trigger timer control	0 : trigger start 1 : trigger start & stop	
MPPG1	PPG output control	0 : continuous pulse 1 : single pulse	
TFF1	timer F/F1 control for PPG output mode	0 : clear 1 : set	

Note 1 : fc ; High-frequency clock [Hz], fs ; Low-frequency clock [Hz]

Note 2 : Writing to the low-byte of the timer registers ($TREG1A_L$, $TREG1B_L$), the comparison is inhibited until the high-byte ($TREG1A_H$, $TREG1B_H$) is written.
After writing to the high-byte, the comparison of 1 cycle (during instruction execution) is ignored.

Note 3 : Set the mode, source clock, edge (INT2ES), PPG control and timer F/F control when TC1 stops (TC1S = 00).

Note 4 : Software capture can be used in only timer and event counter modes.

Note 5 : Values to be loaded to timer registers must satisfy the following condition.
TREG1A > TREG1B > 0 (PPG output mode) ; TREG1A > 0 (others)

Note 6 : Always write "0" to TFF1 except the PPG output mode.

Note 7 : TC1CR is a write-only register, which cannot be accessed by any read-modify-write instruction such as bit operate, etc.

Note 8 : TREG1B cannot be written after setting to PPG output mode.

Figure 2-19. Timer Registers and TC1 Control Register

2.5.3 Function

Timer/counter 1 has six operating modes: timer, external trigger timer, event counter, window, pulse width measurement, programmable pulse generator output mode.

(1) **Timer** Mode

In this mode, counting up is performed using the internal clock. The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared to"0". Counting up resumes after the counteriscleared. The current contents of up-counter can be transfered to TREG1B by setting SCAP1 (bit 6 in TC1CR) to "1" (software capture function). SCAP1 is automatically cleared to "0" after capaturing.

Table 2-3. Timer/Counter 1 Source Clock (Internal Clock)

Source clock			Resolution		Maximum time setting	
NORMAL1/2, IDLE1/2 modes		SLOW, SLEEP modes				
DV7CK = 0	DV7CK = 1		At fc = 8MHz	At fs = 32.768 kHz	At fc = 8 MHz	At fs = 32.768 kHz
$fc/2^3$ [Hz]	$fc/2^3$ [Hz]	–	1 μs	–	65.5 ms	–
$fc/2^7$	$fc/2^7$	–	16 μs	–	1.0 s	–
$fc/2^{11}$	$fs/2^3$	$fs/2^3$ [Hz]	256 μs	244.14 μs	16.8 s	16.0 s

Example 1 : Sets the timer mode with source clock $fs/2^3$[Hz] and generates an interrupt 1 s. later (at fs = 32.768 kHz).

```
LD      (TC1CR), 00000000B      ; Sets the TC1 mode and source clock
LDW     (TREG1A), 1000H         ; Sets the timer register (1s ÷ 2^3 / fs = 1000H)
LD      (TC1CR), 00010000B      ; Starts TC1
```

Example 2 : Software capture

```
LD      (TC1CR), 01010000B      ; SCAP1←1 (Captures)
LD      WA, (TREG1B)            ; Reads captured value
```

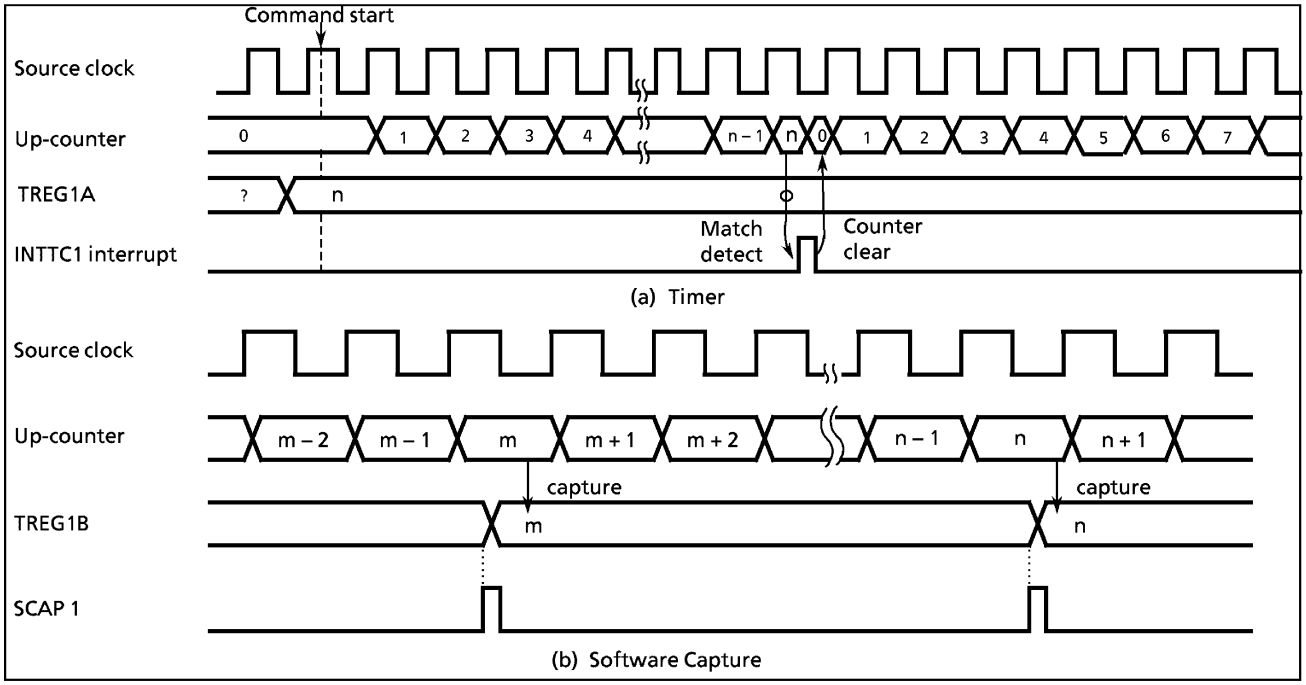


Figure 2-20. Timer Mode Timing Chart

(2) **External Trigger Timer** mode

In this mode, counting up is started by an external trigger. This trigger is the edge of the TC1 pin input. Either the rising or falling edge can be selected with INT2ES. Edge selection is the same as for the external interrupt input INT2 pin. Source clock is used an internal clock selected with TC1CK. The contents of TREG1A is compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared to"0" and halted. The counter is restarted by the selected edge of the TC1 pin input.

The TC1 pin input has the same noise rejection as the INT2 pin; therefore, pulses of 7/fc [s] or less are rejected as noise. A pulse width of 24/fc [s] or more is required for edge detection in NORMAL1/2 or IDLE1/2 mode. The noise rejection circuit is turned off in SLOW and SLEEP modes. But, a pulse width of 4/fs [S] or more is required.

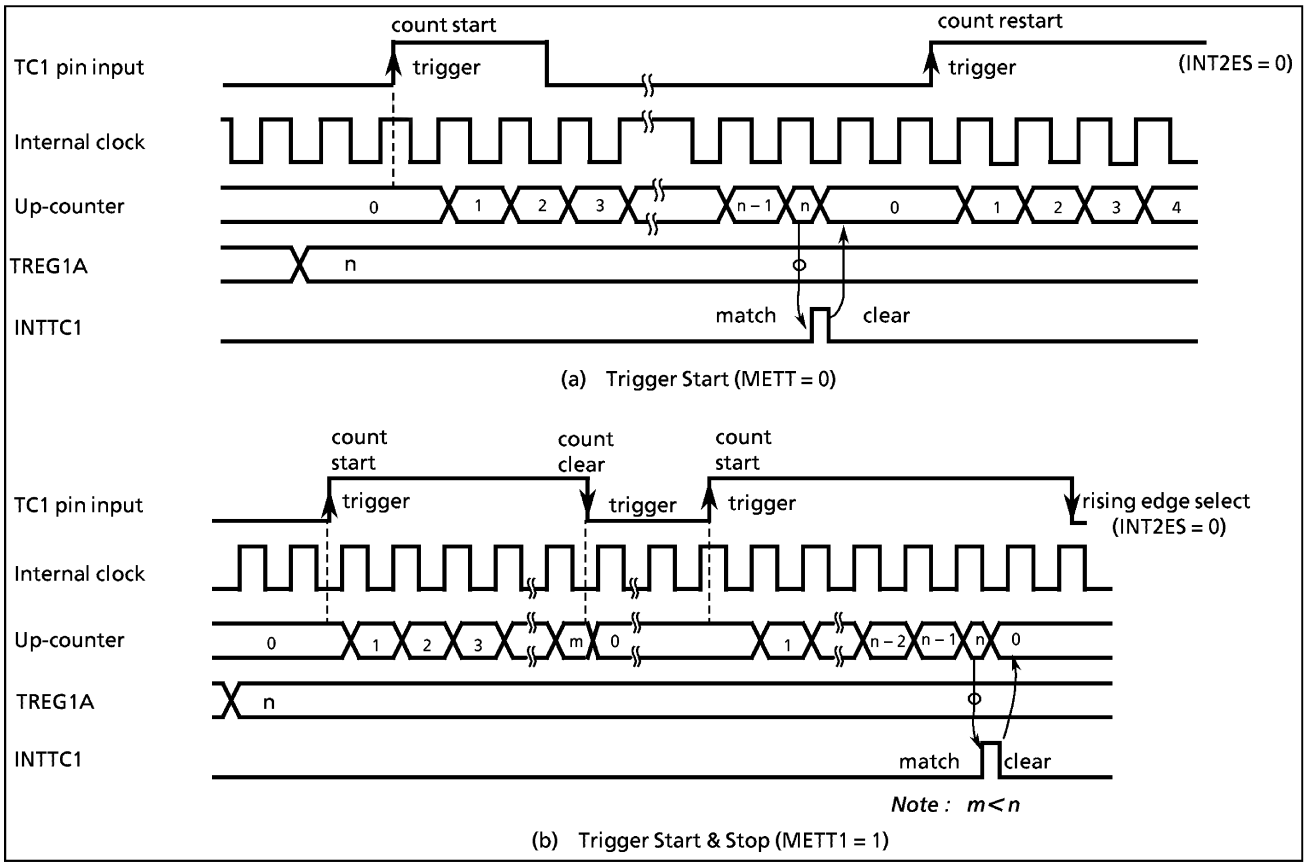


Figure 2-21. External Trigger Timer Mode Timing Chart

(3) **Event Counter** Mode

In this mode, events are counted on the edge of the TC1 pin input. Either the rising or falling edge can be selected with INT2ES in EINTCR. The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. The maximum applied frequency is fc/2^4 [Hz] in NORMAL1/2 or IDLE1/2 mode and fs/2^4 [Hz] in SLOW or SLEEP mode.

Setting SCAP1 to "1" transferres the current contents of up-counter to TREG1B (software capture function). SCAP is automatically cleared after capturing.

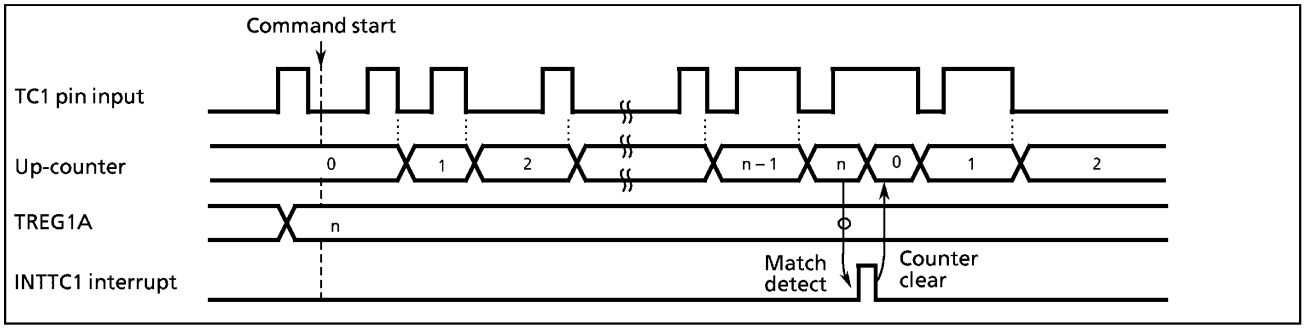


Figure 2-22. Event Counter Mode Timing Chart (INT2ES = 1)

(4) **Window** mode

Counting up is performed on the rising edge of the pulse that is the logical AND-ed product of the TC1 pin input (window pulse) and an internal clock. The contents of TREG1A are compared with the contents of up-counter. If a match is found, an INTTC1 interrupt is generated, and the counter is cleared. Positive or negative logic for the TC1 pin input can be selected with INT2ES. Setting SCAP1 to "1" transferes the current contents of up-counter to TREG1B. It is necessary that the maximum applied frequency (TC1 input) be such that the counter value can be analyzed by the program. That is, the frequency must be considerably slower than the selected internal clock.

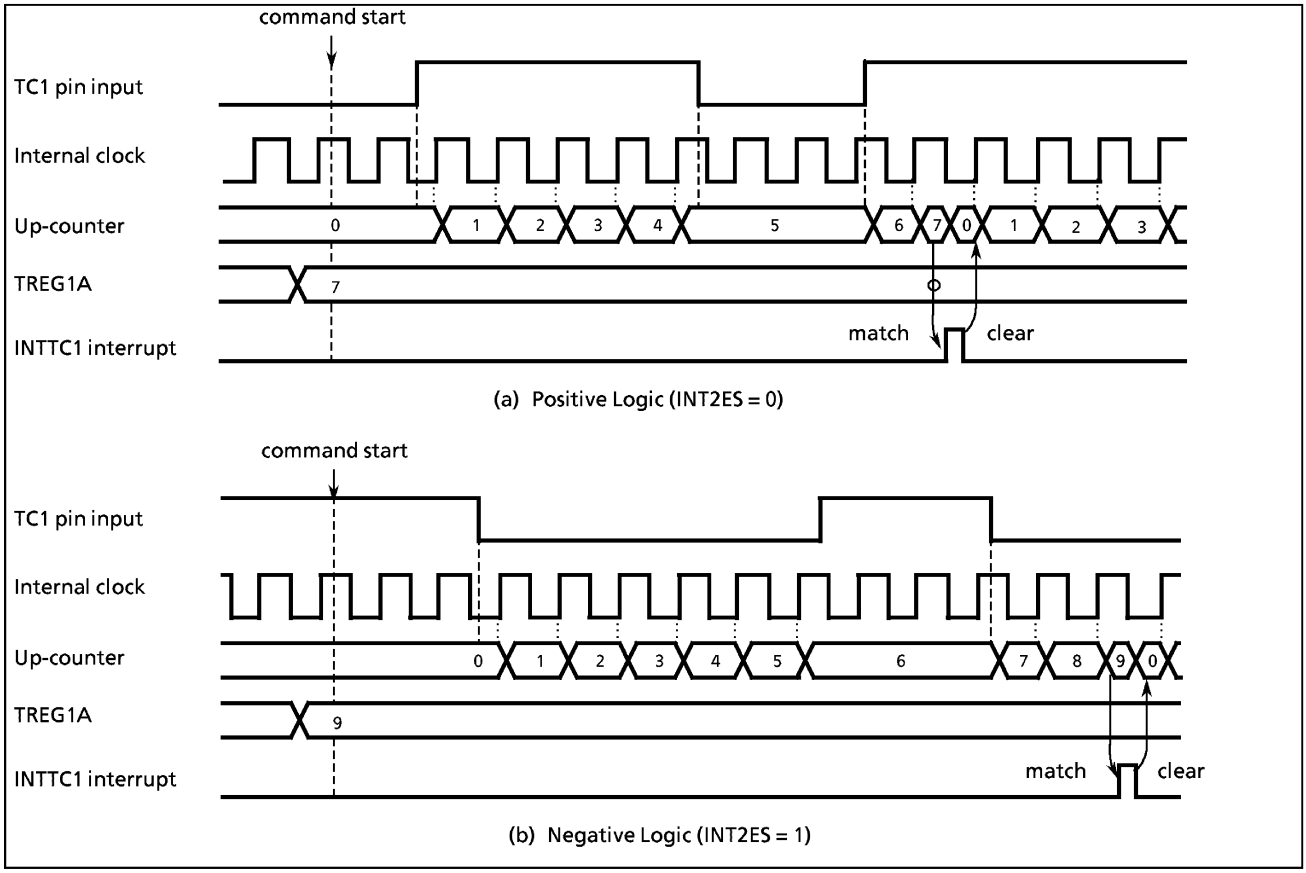


Figure 2-23. Window Mode Timing Chart

(5) **Pulse width measurement** mode

Counting is started by the external trigger (set to external trigger start by TC1S). The trigger can be selected either the rising or falling edge of the TC1 pin input. The source clock is used an internal clock. On the next falling (rising) edge, the counter contents are transferred to TREG1B and an INTTC1 interrupt is generated. The counter is cleared when the single edge capture mode is set. When double edge capture is set, the counter continues and, at the next rising (falling) edge, the counter contents are again transferred to TREG1B. If a falling (rising) edge capture value is required, it is necessary to read out TREG1B contents until a rising (falling) edge is detected. Falling or rising edge is selected with INT2ES, and single edge or double edge is selected with MCAP1 (bit 6 in TC1CR).

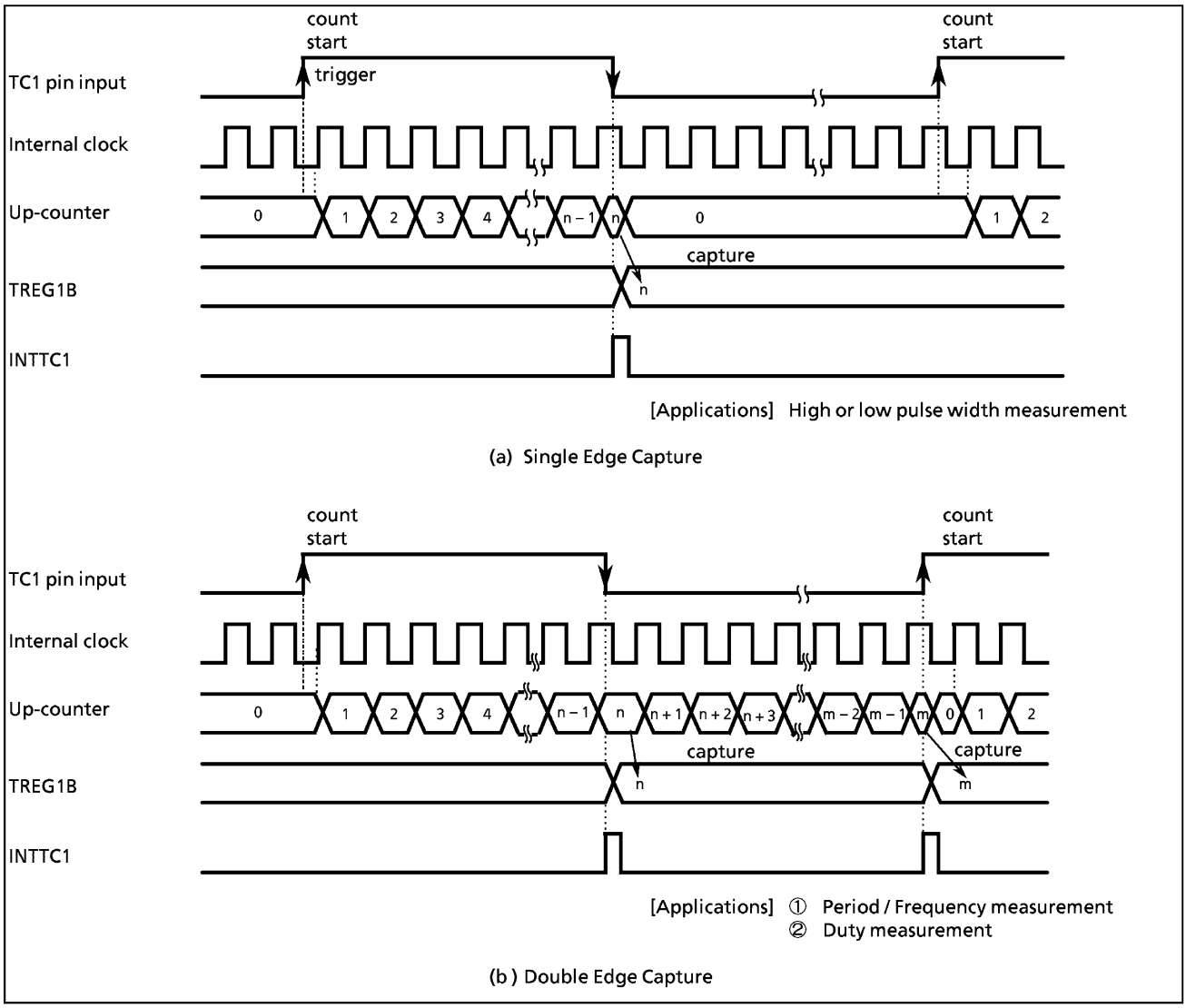


Figure 2-24. Pulse Width Measurement Mode Timing Chart

Example : Duty measurement (Resolution $fc/2^7$ [Hz])

```
        CLR     (INTTC1C). 0            ;   INTTC1 service switch initial setting
        LD      (EINTCR), 00000000B     ;   Sets the rise edge at the INT2 edge
        LD      (TC1CR), 00000110B      ;   Sets the TC1 mode and source clock
        SET     (EIRL). 4               ;   Enables INTTC1
        LD      (TC1CR), 00110110B      ;   Starts TC1 with an external trigger
         :
PINTTC1 : CPL   (INTTC1C). 0            ;   Complements INTTC1 service switch
        JRS     F, SINTTC1
        LD      (HPULSE), (TREG1BL)     ;   Reads TREG1B
        LD      (HPULSE + 1), (TREG1BH)
        RETI
SINTTC1 : LD    (WIDTH), (TREG1BL)      ;   Reads TREG1B (Period)
        LD      (WIDTH + 1), (TREG1BH)
         :
```

(6) **Programmable Pulse Generate** (PPG) **output** mode

Counting is started by an edge of the TC1 pin input (either the rising or falling edge can be selected) or by a command. The source clock is used an internal clock. First, the contents of TREG1B are compared with the contents of the up-counter. If a match is found, timer F/F1 output is toggled. Next, timer F/F1 is again toggled and the counter is cleared by matching with TREG1A. An INTTC1 interrupt is generated at this time. Timer F/F output is connected to the P14 ($\overline{PPG}$) pin. In the case of $\overline{PPG}$ output, set the P14 output latch to "1" and configure as an output with $P1CR_4$. Timer F/F1 is cleared to "0" during reset. The timer F/F1 value can also be set by program and either a positive or negative logic pulse output is available. Also, writing to the TREG1B is not possible unless the timer / counter 1 is set to the PPG output mode with TC1M.

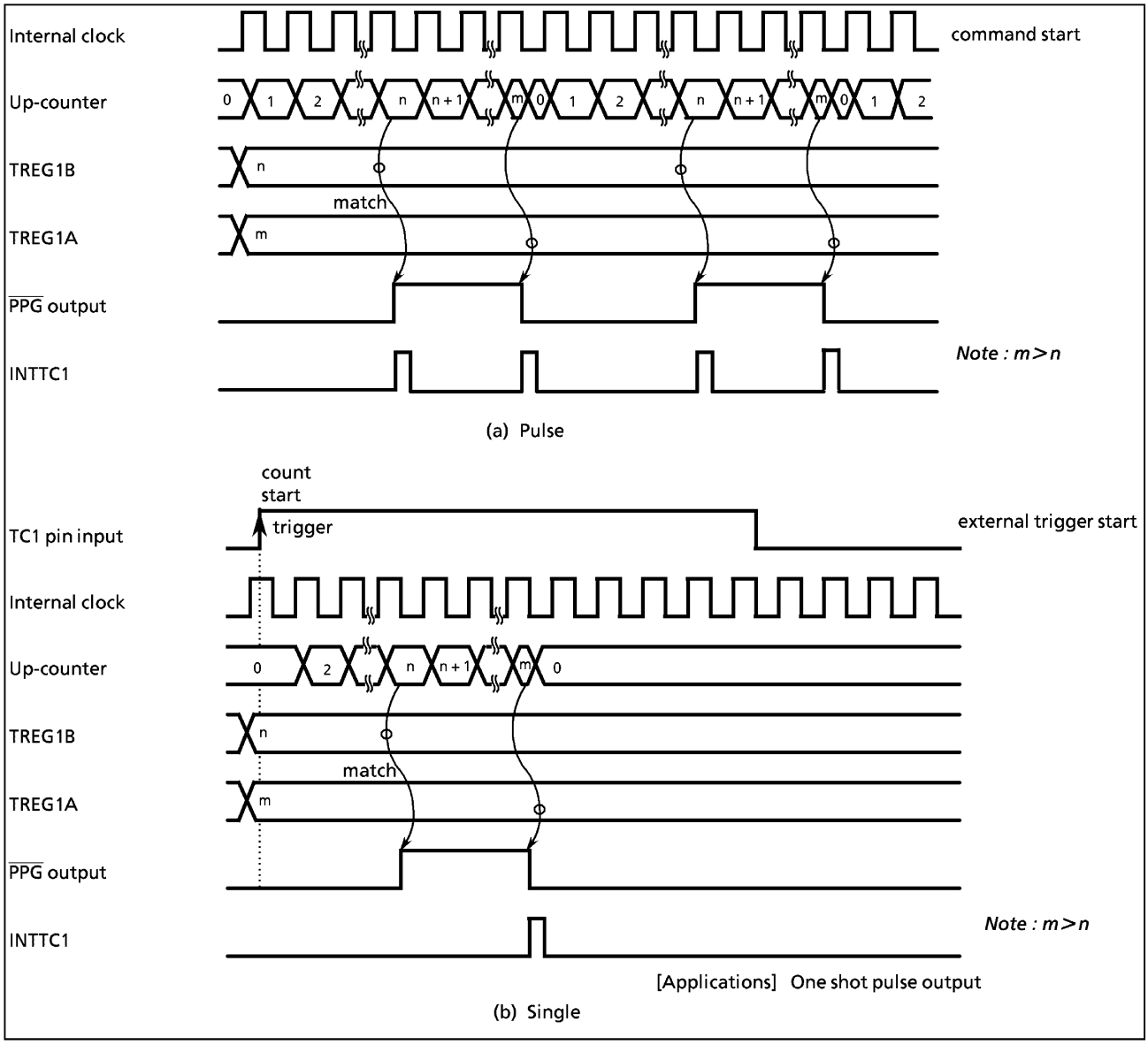


Figure 2-25. PPG Output Mode Timing Chart

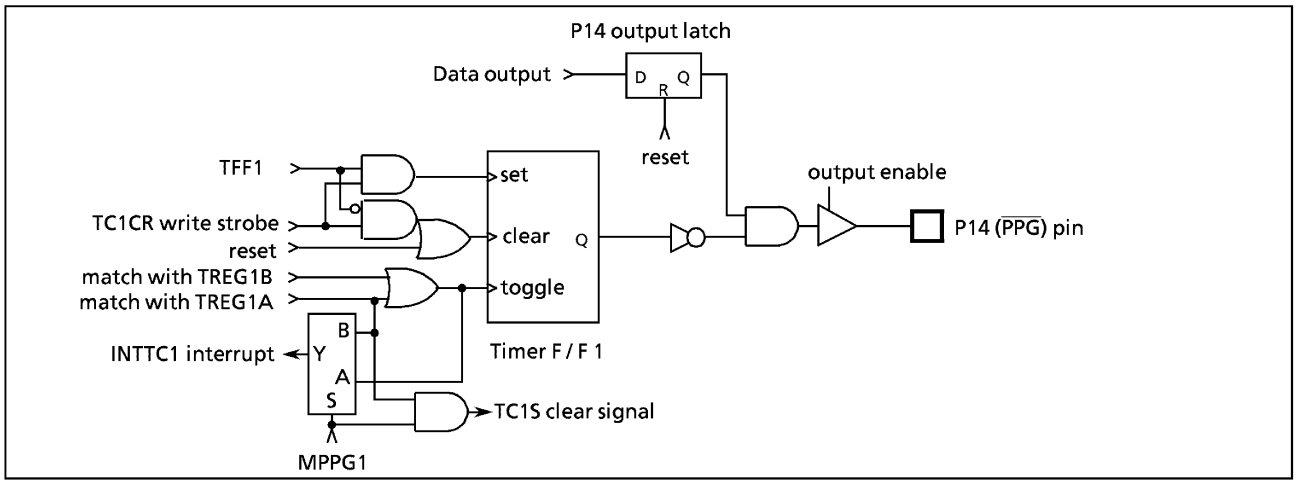


Figure 2-26. PPG Output

2.6 16-bit Timer/Counter 2 (TC2)

2.6.1 Configuration

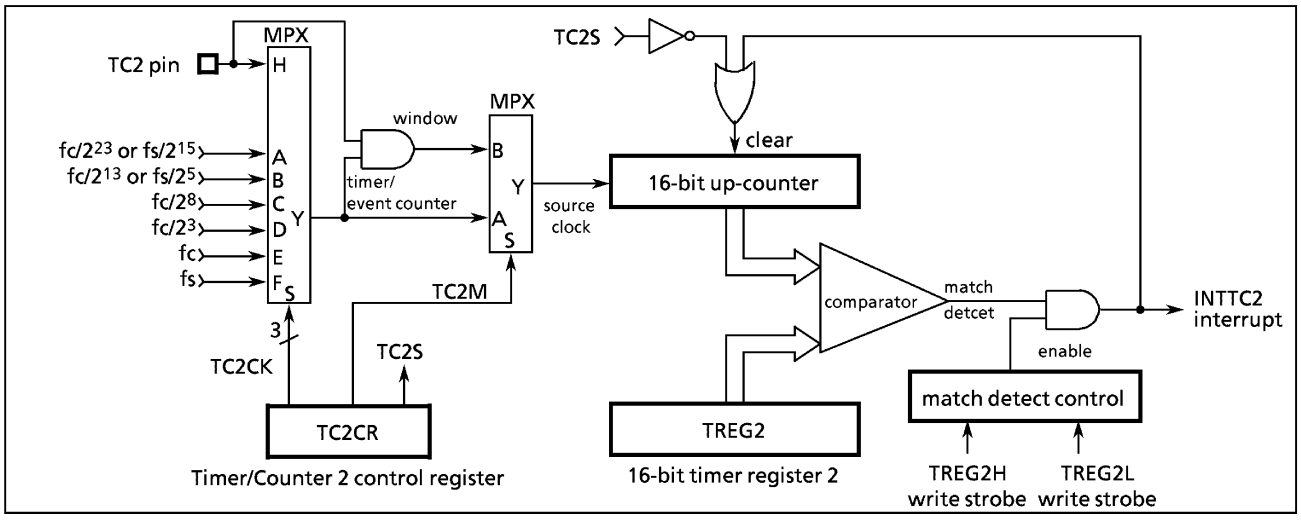


Figure 2-27. Timer/Counter 2 (TC2)

2.6.2 Control

The timer/counter 2 is controlled by a timer/counter 2 control register (TC2CR) and a 16-bit timer register 2 (TREG2). Reset does not affect TREG2.

TREG2 (0016, 0017_H)

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
$TREG2_H$ (0017_H)								$TREG2_L$ (0016_H)							

write only

TC2CR (0015_H)

7	6	5	4	3	2	1	0
"0"	"0"	TC2S	TC2CK			"0"	TC2M

(Initial value : **00 00*0)

TC2M	Timer/counter 2 operating mode select	0 : Timer/Event counter mode 1 : Window mode	write only
TC2CK	Timer/counter 2 source clock select	000 : Internal clock fc / 2^{23} or fs / 2^{15} [Hz] 001 : 〃 fc / 2^{13} or fs / 2^5 010 : 〃 fc / 2^8 011 : 〃 fc / 2^3 100 : 〃 fc (Note 5) 101 : 〃 fs 110 : Reserved 111 : External clock (TC2 pin input)	
TC2S	Timer/counter 2 start control	0 : Stop and counter clear 1 : Start	

*Note 1 : fc; High-frequency clock [Hz], fs; Low-frequency clock [Hz], *; don't care*

Note 2 : When writing to the low-byte of timer register 2 ($TREG2_L$), the comparison is inhibited until the high-byte ($TREG2_H$) is written.
After writing to the high-byte, any match during 1 machine cycle (instruction execution cycle) is ignored.

Note 3 : Set the mode and source clock when timer/counter stops (TC2S = 0).

Note 4 : Values to be loaded to the timer register must satisfy the following condition.
TREG2 > 0 ($TREG2_{15\sim11}$>0 when warm-up).

Note 5 : "fc" can be selected as the source clock only in the timer mode during the SLOW mode.

Note 6 : Always write "0" to bit 0 in TC2CR.

Note 7: TC2CR and TREG2 are write-only registers and must not be used with any of the read-modify-write instructions.

Figure 2-28. Timer Register 2 and TC2 Control Register

2.6.3 Function

The timer/counter 2 has three operating modes: timer, event counter and window modes. Also timer/counter 2 is used for warm-up when switching from SLOW mode to NORMAL2 mode.

(1) **Timer** Mode

In this mode, the internal clock is used for counting up. The contents of TREG2 are compared with the contents of up-counter. If a match is found, a timer/ counter 2 interrupt (INTTC2) is generated, and the counter is cleared. Counting up is resumed after the counter is cleared.

Also, when fc is selected as the source clock during SLOW mode, the lower 11 bits of TREG2 are ignored and an INTTC2 interrupt is generated by matching the upper 5 bits. Thus, in this case, only the $TREG2_H$ setting is necessary.

Table 2-4. Source Clock (Internal Clock) for Timer/Counter 2

Source clock				Resolution		Maximum time setting	
NORMAL1/2, IDLE1/2 mode		SLOW mode	SLEEP mode				
DV7CK = 0	DV7CK = 1			At fc = 8 MHz	At fs = 32.768 kHz	At fc = 8 MHz	At fs = 32.768 kHz
$fc/2^{23}$[Hz]	$fs/2^{15}$[Hz]	$fs/2^{15}$ [Hz]	$fs/2^{15}$ [Hz]	1.05 s	1 s	19.1 h	18.2 h
$fc/2^{13}$	$fs/2^{5}$	$fs/2^{5}$	$fs/2^{5}$	1.02 ms	1 ms	1.1 min	1 min
$fc/2^{8}$	$fc/2^{8}$	–	–	32 μs		2.1 s	
$fc/2^{3}$	$fc/2^{3}$	–	–	1 μs		65.5 ms	
–	–	fc (Note)	–	125 ns		8.2 ms	
fs	fs	–	–		30.5 μs		2 s

Note : "fc" can be used only in the timer mode.

Example : Sets the timer mode with source clock $fc/2^3$ [Hz] and generates an interrupt every 25 ms (at fc = 8 MHz).

```
LD      (TC2CR), 00001100B      ; Sets the TC2 mode and source clock
LDW     (TREG2), 61A8H          ; Sets TREG2 (25 ms ÷ 2^3/fc = 61A8H)
LD      (TC2CR), 00101100B      ; Starts TC2
```

(2) **Event Counter** Mode

In this mode, events are counted on the rising edge of the TC2 pin input. The contents of TREG2 are compared with the contents of the up-counter. If a match is found, an INTTC2 interrupt is generated, and the counter is cleared. The maximum frequency applied to the TC2 pin is $fc/2^4$ [Hz] in NORMAL1/2 or IDLE1/2 mode, and $fs/2^4$ [Hz] in SLOW or SLEEP mode.

Example : Sets the event counter mode and generates an INTT2 interrupt 640 counts later.

```
LD      (TC2CR), 00011100B      ; Sets the TC2 mode
LDW     (TREG2), 0280H          ; Sets TREG2
LD      (TC2CR), 00111100B      ; Starts TC2
```

(3) **Window** Mode

In this mode, counting up is performed on the rising edge of the pulse that is the logical AND-ed product of the TC2 pin input (window pulse) and an internal clock. The internal clock is selected with TC2CK. The contents of TREG2 are compared with the contents of up-counter. If a match is found, an INTTC2 interrupt is generated, and the up-counter is cleared to "0". It is necessary that the maximum applied frequency (TC2 input) be such that the counter value can be analyzed by the program. That is, the frequency must be considerably slower than the selected internal clock.

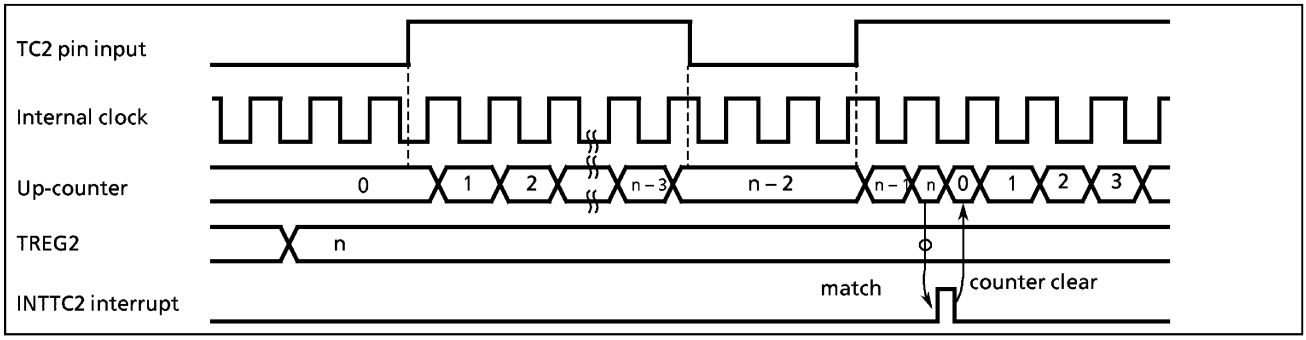


Figure 2-29. Window Mode Timing Chart

2.7 8-Bit Timer/Counter 3 (TC3)

2.7.1 Configuration

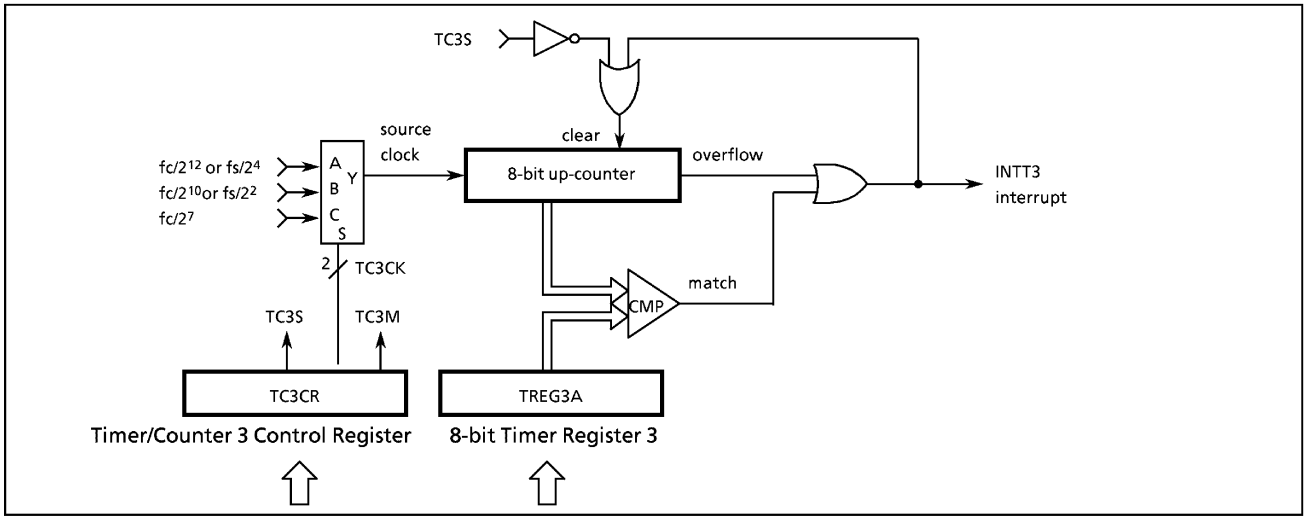


Figure 2-30. Timer/Counter 3

2.7.2 Control

TREG3A (0018_H)

7	6	5	4	3	2	1	0

Read/Write

TC3CR ($001A_H$)

7	6	5	4	3	2	1	0
"0"	"0"	"0"	TC3S	TC3CK		"0"	TC3M

(Initial value : ***0 00*0)

TC3M	Timer/counter 3 operation mode set	0 : Timer mode 1 : reserved	Write only
TC3CK	Timer/counter 3 source clock select	00 : Internal clock fc / 2^{12} or fs / 2^4 [Hz] 01 : Internal clock fc / 2^{10} or fs / 2^2 10 : Internal clock fc / 2^7 11 : reserved	
TC3S	Timer/counter 3 start select	0 : Stop & clear 1 : Start	

*Note 1 : fc ; High-frequency clock [Hz] fs ; Low-frequency clock [Hz] * ; don't care*
Note 2 : Set the mode and the source clock when the TC3 stops (TC3S = 0).
Note 3 : Values to be loaded into timer register 3A must satisfy the following condition.
TREG3A > 0 (in the timer mode)
Note 4 : TC3CR is a write-only register and must not be used with any of read-modify-write instructions.

Figure 2-31. Timer Register 3A and TC3 Control Register

The timer/counter 3 is controlled by a timer/counter 3 control register (TC3CR) and two 8-bit timer registers (TREG3A). Reset does not affect these timer registers.

2.7.3 Function

The timer/counter 3 has timer mode.

(1) **Timer** Mode

In this mode, the internal clock is used for counting up. The contents of TREG3A are compared with the contents of up-counter. If a match is found, a timer/counter 3 interrupt (INTTC3) is generated, and the up-counter is cleared. Counting up resumes after the up-counter is cleared.

Table 2-5. Source Clock (Internal Clock) for Timer Counter 3

Source clock			Resolution		Maximum setting time	
NORMAL1 / 2, IDLE1 / 2 mode		SLOW, SLEEP mode				
DV7CK = 0	DV7CK = 1		fc = 8 MHz	fs = 32.768 kHz	fc = 8 MHz	fs = 32.768 kHz
$fc/2^{12}$ [Hz]	$fs/2^{4}$ [Hz]	$fs/2^{4}$ [Hz]	512 μs	488.28 μs	131.1 ms	124.5 ms
$fc/2^{10}$	$fs/2^{2}$	–	128 μs	122.07 μs	32.6 ms	31.1 ms
$fc/2^{7}$	–	–	16 μs	–	4.1 ms	–

2.8 8-bit Timer/Counter (TC4)

2.8.1 Configuration

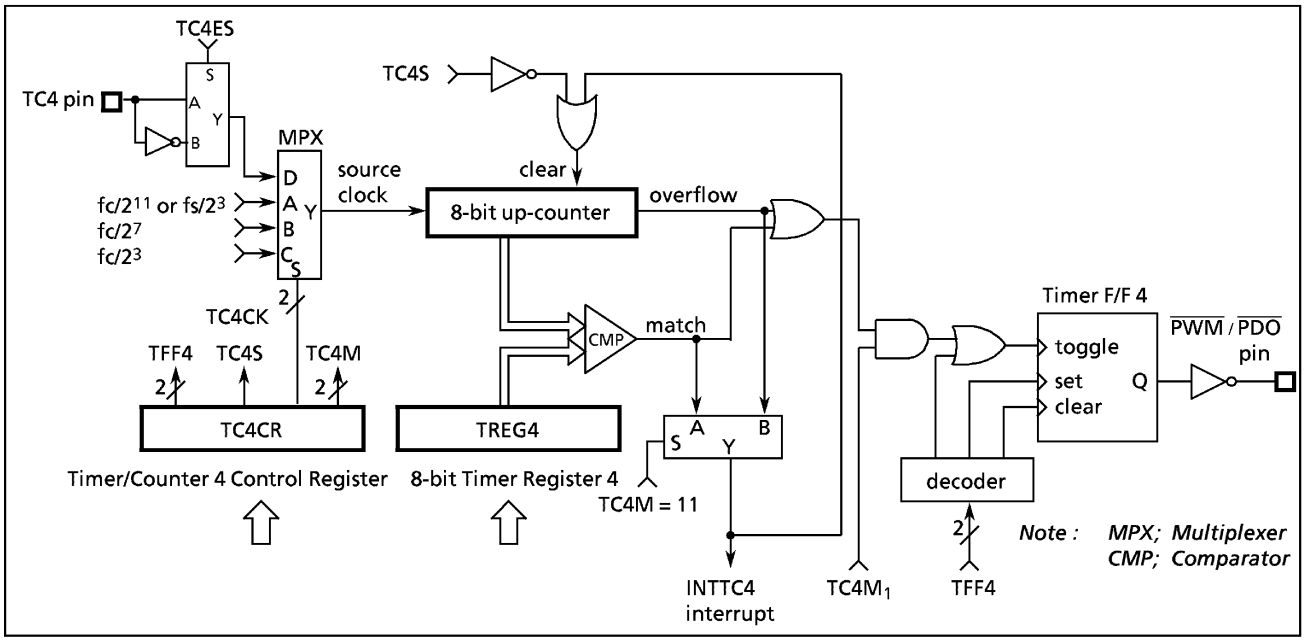


Figure 2-32. Timer/Counter 4

2.8.2 Control

The timer/counter 4 is controlled by a timer/counter 4 control register (TC4CR) and an 8-bit timer register 4 (TREG4). Reset does not affect TREG4.

TREG4 ($001B_H$)

7	6	5	4	3	2	1	0

Write only

TC4CR ($001C_H$)

7	6	5	4	3	2	1	0
TFF4		"0"	TC4S	TC4CK		TC4M	

(Initial value : 00*0 0000)

TC4M	TC4 operating mode select	00 : Timer/event counter mode 01 : Reserved 10 : Programmable divider output (PDO) mode 11 : Pulse width modulation (PWM) output mode	Write only
TC4CK	TC4 source clock select	00 : Internal clock fc / 2^{11} or fs / 2^3[Hz] 01 : Internal clock fc / 2^7 10 : Internal clock fc / 2^3 11 : External clock (TC4 pin input)	
TC4S	TC4 start control	0 : Stop & clear 1 : Start	
TFF4	Timer F/F 4 control	00 : Clear 01 : Toggle 10 : Set 11 : – (Note 3)	

*Note 1 : fc; High-frequency clock [Hz], fs; Low-frequency clock [Hz], *; don't care*

Note 2 : Set the operating mode, the source clock selection, the timer F/F 4 control and the edge selection (INT4ES) when the TC4 stops (TC4S = 0).

Note 3 : TFF4 must be set to "11" in the timer and event counter modes.

Note 4 : Values to be loaded to the timer register must satisfy the following condition.

$TREG4 > 0$

Note 5 : TC4CR and TREG4 are write-only registers and must not be used with any of read-modify-write instructions.

Figure 2-33a. Timer Register 4 and TC4 Control Register

EINTCR (0037_H)

7	6	5	4	3	2	1	0
(INT1 NC)	(INT0 EN)		TC4 ES	(INT3 ES)	(INT2 ES)	(INT1 ES)	

(Initial value : 00*0 000*)

TC4ES	Edge selection of TC 4pin input	0 : Rising edge select 1 : Falling edge select	R/W

Figure 2-33b. External interrupt Control Register

2.8.3 Function

The timer/counter 4 has four operating modes : timer, event counter, programmable divider output, and PWM output mode.

(1) **Timer** Mode

In this mode, the internal clock is used for counting up. The contents of TREG4 are compared with the contents of up-counter. If a match is found, a timer/counter 4 interrupt (INTTC4) is generated and the up-counter is cleared to "0". Counting up resumes after the up- counter is cleared.

Table 2-6. Source Clock (Internal Clock) for Timer/Counter 4

Source clock			Resolution		Maximum setting time	
NORMAL1 / 2, IDLE1 / 2 mode		SLOW, SLEEP mode				
DV7CK = 0	DV7CK = 1		fc = 8 MHz	fs = 32.768 kHz	fc = 8 MHz	fs = 32.768 kHz
$fc/2^{11}$ [Hz]	$fs/2^3$ [Hz]	$fs/2^3$ [Hz]	256 μs	244.14 μs	65.3 ms	62.2 ms
$fc/2^7$	–	–	16 μs	–	4.1 ms	–
$fc/2^3$	–	–	1 μs	–	255 μs	–

(2) **Event Counter** Mode

In this mode, the TC4 pin input (external clock) pulse is used for counting up. Either the rising or falling edge can be selected with TC4ES (bit 4 in EINTCR). The contents of the TREG4 are compared with the contents of the up-counter. If a match is found, an INTTC4 interrupt is generated and the counter is cleared. The maximum applied frequency is $fc/2^4$ [Hz] in NORMAL1/2 or IDLE1/2 mode, and $fs/2^4$ [Hz] in SLOW or SLEEP mode. Two or more machine cycles are required for both the high and low levels of the pulse width.

(3) **Programmable Divider Output** (PDO) Mode

The internal clock is used for counting up. The contents of TREG4 are compared with the contents of the up-counter. Timer F/F 4 output is toggled and the counter is cleared each time a match is found. Timer F/F 4 output is inverted and output to the $\overline{PDO}$ (P33) pin. This mode can be used for 50 % duty pulse output. Timer F/F 4 can be initialized by program, and it is initialized to "0" during reset. An INTTC4 interrupt is generated each time the $\overline{PDO}$ output is toggled.

Example : Output a 1024 Hz pulse (at fc = 4.194304 MHz)

```
SET     (P3). 3              ;  P33 output latch ← 1
LD      (TREG4), 10H         ;  1/2048 ÷ 2^7/fc = 10H
LD      (TC4CR), 00011010B   ;  Starts TC4
```

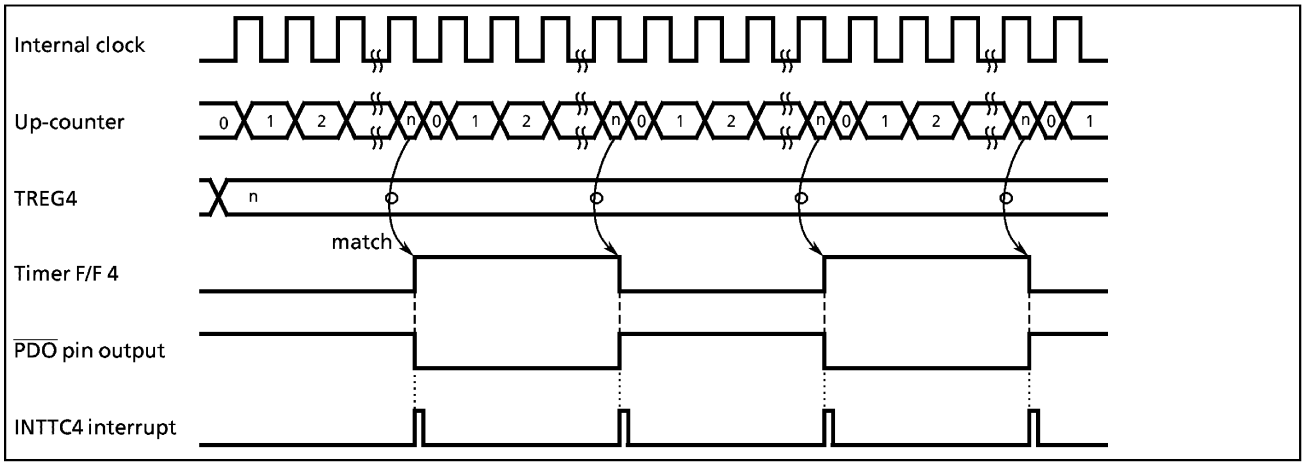


Figure 2-34. Timing Chart for PDO Mode

(4) **Pulse Width Modulation** (PWM) **Output** Mode

PWM output with a resolution of 8 bits is possible. The internal clock is used for counting up. The contents of TREG4 are compared with the contents of up-counter. If a match is found, the timer F/F 4 output is toggled. The counter continues counting. And, when an overflow occurs, the timer F/F 4 output is again toggled and the counter is cleared. Timer F/F 4 output is inverted and output to the $\overline{PWM}$ (P33) pin. An INTTC4 interrupt is generated when an overflow occurs.

TREG4 is configured a 2-stage shift register and, during output, will not switch until one output cycle is completed even if TREG4 is overwritten; therefore, output can be altered continuously. Also, the first time, TREG4 is shifted by setting TC4S (bit 4 in TC4CR) to "1" after data are loaded to TREG4.

Note 1 : Do not overwrite TREG4 only when an INTTC4 interrupt is generated. Usually, TREG4 is overwritten in the routine of INTTC4 interrupt service.
Note 2 : PWM output mode can be used only in the NORMAL 1, 2 and IDLE 1, 2 mode.

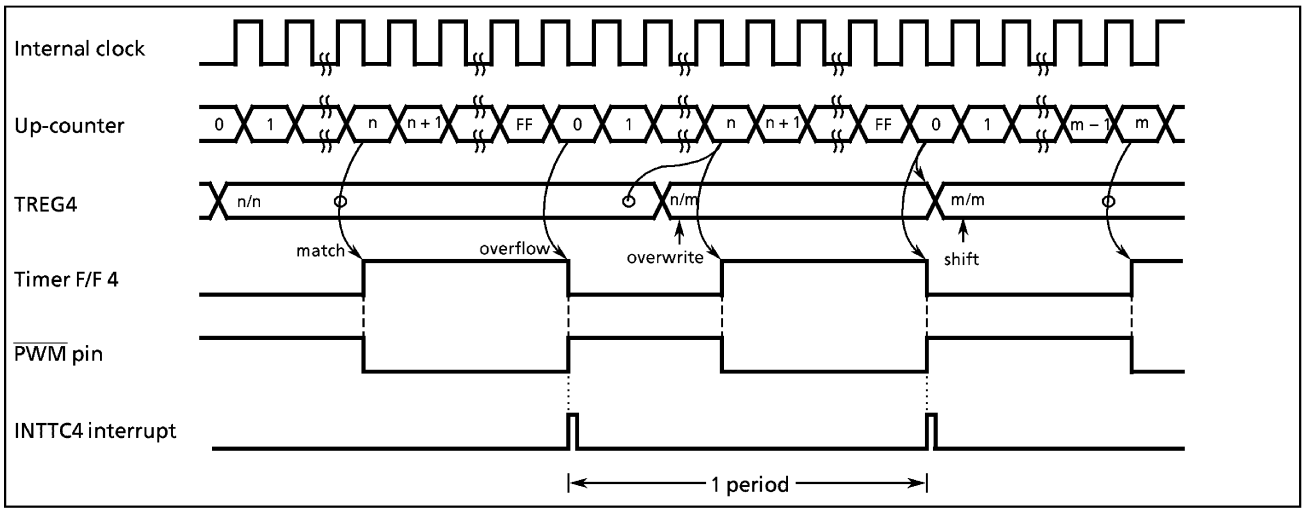


Figure 2-35. Timing Chart for PWM Mode

Table 2-7. PWM Output Mode

Source clock			Resolution		Maximum setting time	
NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode				
DV7CK = 0	DV7CK = 1		fc = 8 MHz	fs = 32.768 kHz	fc = 8 MHz	fs = 32.768 kHz
$fc / 2^{11}$ [Hz]	$fs / 2^3$ [Hz]	$fs / 2^3$ [Hz]	256 μs	244.14 μs	65.5 ms	62.5 ms
$fc / 2^7$	$fc / 2^7$	–	16 μs		4.1 ms	
$fc / 2^3$	$fc / 2^3$	–	1 μs		256 μs	

2.9 Serial Interface (SIO)

The 87C814/H14/K14/M14 each have two clocked-synchronous 8-bit serial interfaces (SIO). Each serial interface has an 8-byte transmit and receive data buffer that can automatically and continuously transfer up to 64 bits of data.

The serial interfaces are connected to external devices via pins P32 (SO), P31 (SI), P30 ($\overline{SCK}$). The serial interface pins are also used as port P3. When used as serial interface pins, the output latches of these pins should be set to "1". In the transmit mode, pins P31 can be used as normal I/O ports, and in the receive mode, the pins P32 can be used as normal I/O ports.

2.9.1 Configuration

The SIO have the same configuration, except for the addresses/bit positions of the control/ status registers and buffer registers.

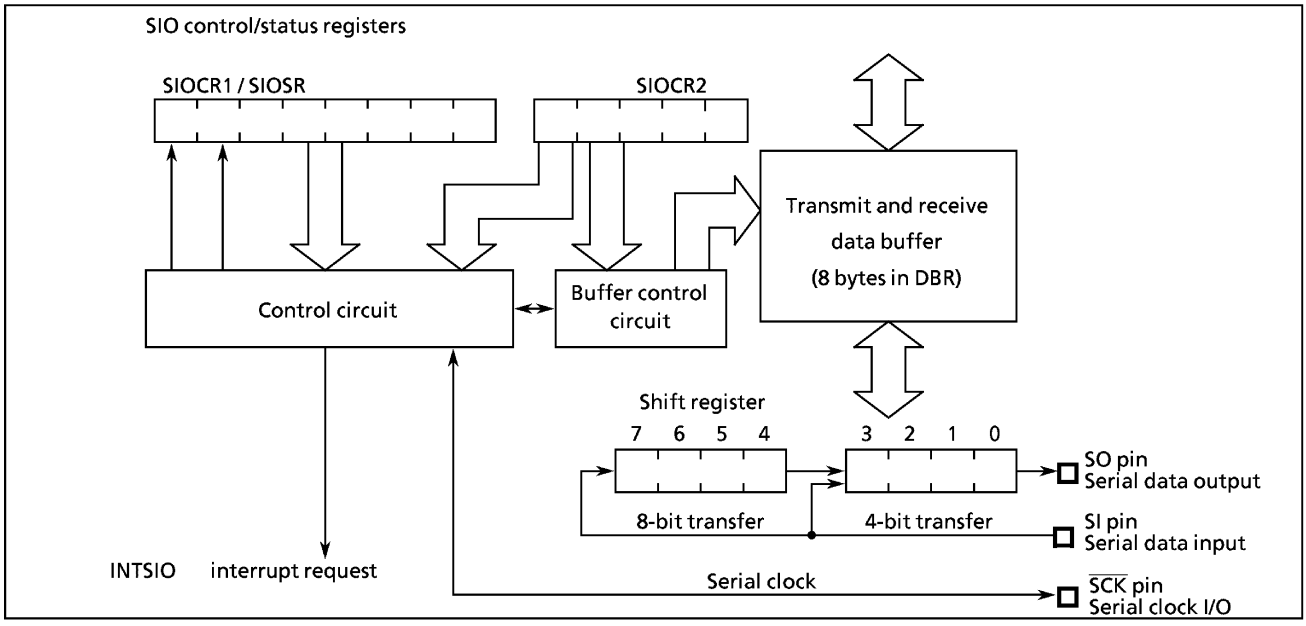


Figure 2-36. Serial Interfaces

2.9.2 Control

The serial interfaces are controlled by SIO control registers (SIOCR1/SIOCR2). The serial interface status can be determined by reading SIO status registers (SIOSR).

The transmit and receive data buffer is controlled by the BUF (bits 2-0 in SIOCR2). The data buffer is assigned to addresses $0FF0_H$ - $0FF7_H$ in the DBR area, and can continuously transfer up to 8 words (bytes or nibbles) at one time. When the specified number of words has been transferred, a buffer empty (in the transmit mode) or a buffer full (in the receive mode or transmit/receive mode) interrupt (INTSIO) is generated.

When the internal clock is used as the serial clock in the 8-bit receive mode and the 8-bit transmit/receive mode, a fixed interval wait can be applied to the serial clock for each word transferred. Four different wait times can be selected with WAIT (bits 4 and 3 in SIOCR2).

SIO Control Registers 1

SIOCR1 (0020_H)

7	6	5	4	3	2	1	0
SIOS	SIO INH	SIOM			SCK		

(Initial value : 0000 0000)

SIOS	Indicate transfer start/stop	0 : Stop 1 : Start	write only
SIOINH	Continue/abort transfer	0 : Continue transfer 1 : Abort transfer (automatically cleared after abort)	
SIOM	Transfer mode select	000 : 8-bit transmit mode 010 : 4-bit transmit mode 100 : 8-bit transmit/receive mode 101 : 8-bit receive mode 110 : 4-bit receive mode	
SCK	Serial clock select	000 : Internal clock fc / 2^{13} or fs / 2^5 [Hz] 001 : Internal clock fc / 2^8 010 : Internal clock fc / 2^6 011 : Internal clock fc / 2^5 111 : External clock (input from $\overline{SCK}$ pin) (Output on $\overline{SCK}$ pin)	

Note 1 : fc ; High-frequency clock [Hz], fs ; Low-frequency clock [Hz]

Note 2 : Set SIOS to "0" and SIOINH to "1" when setting the transfer mode or serial clock.

Note 3 : SIOCR1 is write-only register, which cannot access any of in read-modify-write instruction such as bit operate, etc.

SIO Status Registers

SIOSR (0020_H)

7	6	5	4	3	2	1	0
SIOF	SEF	"1"	"1"	"1"	"1"	"1"	"1"

SIOF	Serial transfer operating status monitor	0 : Transfer terminated (Subsequently to SIOS cleared to "0", when a transfer is terminated to SIOINH is set, SIOF is cleared to "0".) 1 : Transfer in process	read only
SEF	Shift operating status monitor	0 : Shift operation terminated 1 : Shift operation in process	

SIO Control Registers 2

SIOCR2 (0021_H)

7	6	5	4	3	2	1	0
			WAIT		BUF		

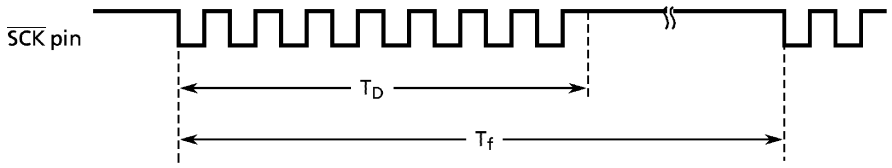
(Initial value: ***0 0000)

WAIT	Wait control	00 : $T_f = T_D$ 01 : $T_f = 2T_D$ 10 : $T_f = 4T_D$ 11 : $T_f = 8T_D$	Write only
BUF	Number of transfer words	Buffer address used SIO 000 : 1 word transfer — $0FF0_H$ 001 : 2 words transfer — 0FF0 - $0FF1_H$ 010 : 3 words transfer — 0FF0 - $0FF2_H$ 011 : 4 words transfer — 0FF0 - $0FF3_H$ 100 : 5 words transfer — 0FF0 - $0FF4_H$ 101 : 6 words transfer — 0FF0 - $0FF5_H$ 110 : 7 words transfer — 0FF0 - $0FF6_H$ 110 : 8 words transfer — 0FF0 - $0FF7_H$	

*Note 1 : *; don't care*

Note 2 : WAIT is valid only in the 8-bit transmit / receive and 8-bit receive modes.

Note 3 : T_f; frame time, T_D ; data transfer time



Note 4 : The lower 4 bits of each buffer are used during 4-bit transfers. Zeros (0) are stored to the upper 4bits when receiving.

Note 5 : Transmitting starts at the lowest address. Received data are also stored starting from the lowest address to the highest address. For example, in the case of SIO, the first buffer address transmitted is $0FF0_H$.

Note 6 : The value to be loaded to BUF is held after transfer is completed.

Note 7 : SIOCR2 is write-only registers, which cannot access any of in read-modify-write instruction such as bit operate, etc.

Figure 2-37. SIO Control Registers and Status Registers

(1) Serial Clock

a. Clock Source

SCK (bits 2 - 0 in SIOCR) is able to select the following:

① Internal Clock

Any of four frequencies can be selected. The serial clock is output to the outside on the $\overline{SCK}$ pin. The $\overline{SCK}$ pin goes high when transfer starts.

When data writing (in the transmit mode) or reading (in the receive mode or the transmit/receive mode) cannot keep up with the serial clock rate, there is a wait function that automatically stops the serial clock and holds the next shift operation until the read/write processing is completed.

Table 2-8. Serial Clock Rate

Serial clock			Maximum transfer rate	
NORMAL1/2, IDLE1/2 mode		SLOW, SLEEP mode		
DV7CK = 0	DV7CK = 1		At fc = 8 MHz	At fs = 32.768 kHz
$fc / 2^{13}$ [Hz]	$fs / 2^5$ [Hz]	$fs / 2^5$ [Hz]	0.95 Kbit/s	1 Kbit/s
$fc / 2^8$	$fc / 2^8$	–	30.5	
$fc / 2^6$	$fc / 2^6$	–	122	
$fc / 2^5$	$fc / 2^5$	–	244	

Note : 1Kbit = 1024 bit

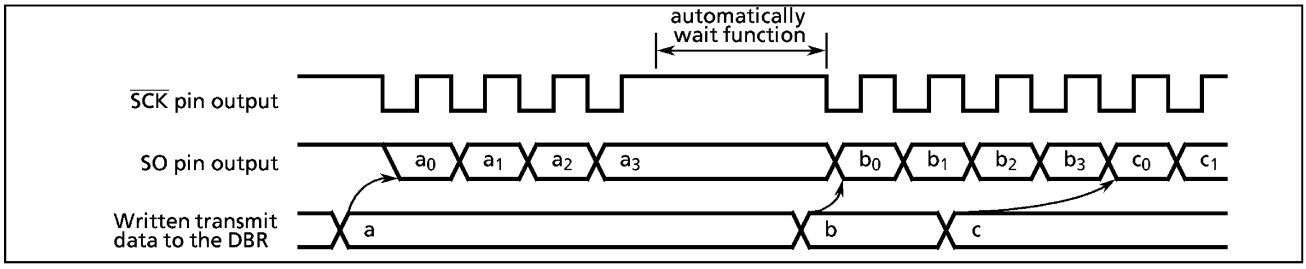
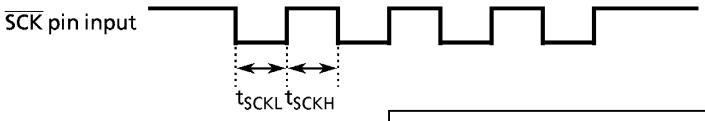


Figure 2-38. Clock Source (Internal Clock)

② External Clock

An external clock connected to the SCK pin is used as the serial clock. In this case, the P30 (SCK) output latch must be set to "1". To ensure shifting, a pulse width of at least 4 machine cycles is required. Thus, the maximum transfer speed is 244K-bit/s. (at fc = 8 MHz).



Note : tcyc = 4/fc (In NORAML1/2, IDLE1/2 modes)
4/fs (In SLOW, SLEEP modes)

b. Shift edge

The leading edge is used to transmit, and the trailing edge is used to receive.

① Leading Edge

Transmitted data are shifted on the leading edge of the serial clock (falling edge of the SCK pin input/output).

② Trailing Edge

Received data are shifted on the trailing edge of the serial clock (rising edge of the SCK pin input/output).

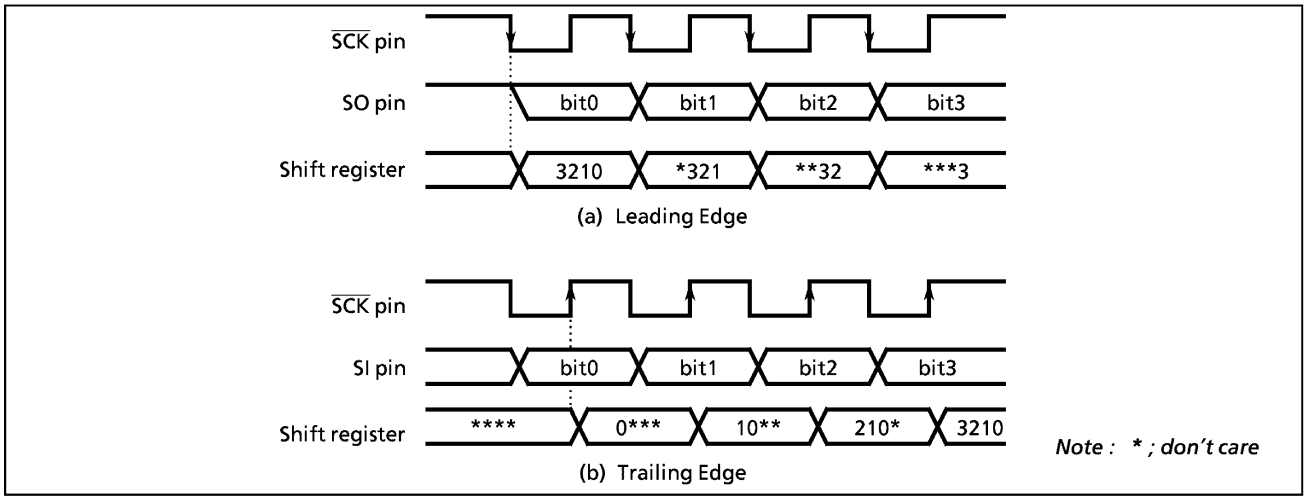


Figure 2-39. Shift Edge

(2) Number of Bits to Transfer

Either 4-bit or 8-bit serial transfer can be selected. When 4-bit serial transfer is selected, only the lower 4 bits of the transmit/receive data buffer register are used. The upper 4 bits are cleared to "0" when receiving.

The data is transferred in sequence starting at the least significant bit (LSB).

(3) Number of Words to Transfer

Up to 8 words consisting of 4 bits of data (4-bit serial transfer) or 8 bits (8-bit serial transfer) of data can be transferred continuously. The number of words to be transferred is loaded to BUF in SIOBCR. An INTSIO interrupt is generated when the specified number of words has been transferred. If the number of words is to be changed during transfer, the serial interface must be stopped before making the change.

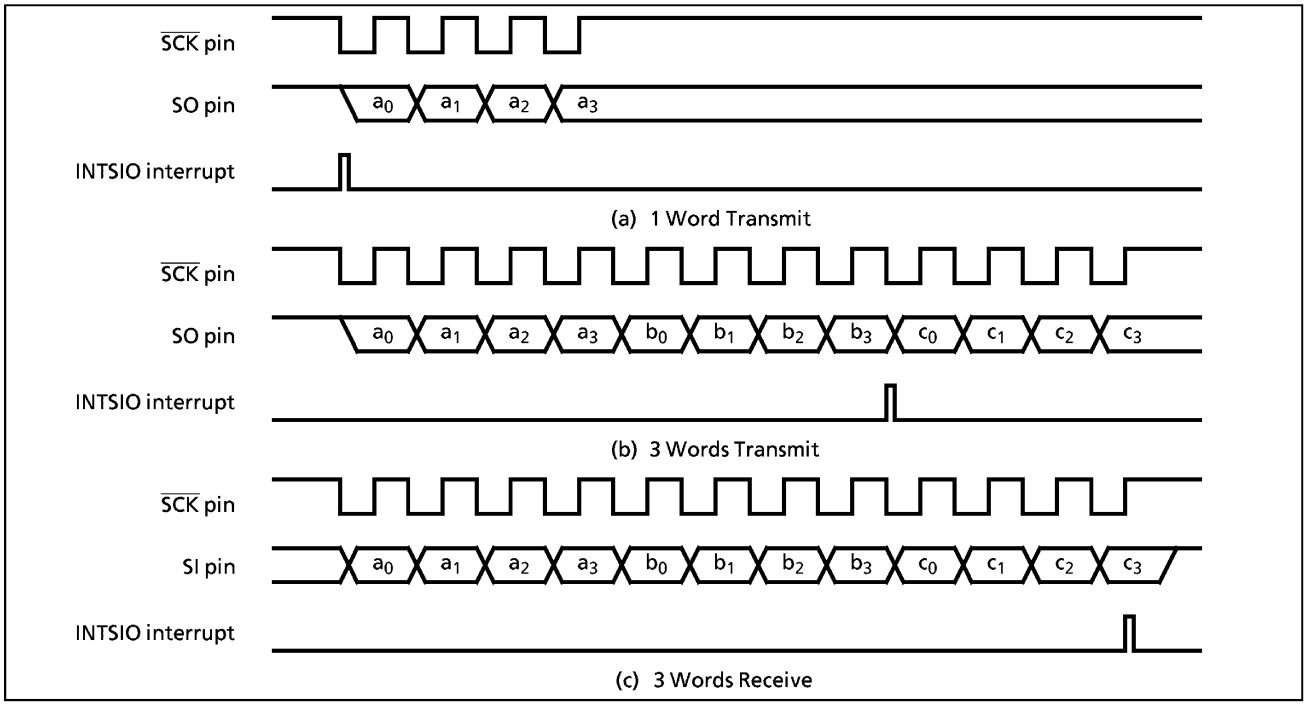


Figure 2-40. Number of Bits to Transfer (Example : 4-bit serial transfer)

2.9.3 Transfer Mode

SIOM (bits 5 - 3 in SIOCR1) is used to select the transmit, receive, or transmit/receive mode.

(1) **4-bit and 8-bit Transmit** Modes

In these modes, the SIOCR1 is set to the transmit mode and then the data to be transmitted first are written to the data buffer registers (DBR). After the data are written, the transmission is started by setting SIOS to "1". The data are then output sequentially to the SO pin in synchronous with the serial clock, starting with the least significant bit (LSB). As soon as the LSB has been output, the data are transferred from the data buffer register to the shift register. When the final data bit has been transferred and the data buffer register is empty, an INTSIO (buffer empty) interrupt is generated to request the next transmitted data.

When the internal clock is used, the serial clock will stop and an automatic-wait will be initiated if the next transmitted data are not loaded to the data buffer register by the time the number of data words specified with the BUF has been transmitted. Writing even one word of data cancels the automatic-wait; therefore, when transmitting two or more words, always write the next word before transmission of the previous word is completed.

Note :	*Waits are also canceled by writing to a DBR not being used as a transmit data buffer register; therefore, during SIO do not use such DBR for other applications.*

When an external clock is used, the data must be written to the data buffer register before shifting next data. Thus, the transfer speed is determined by the maximum delay time from the generation of the interrupt request to writing of the data to the data buffer register by the interrupt service program.

The transmission is ended by clearing SIOS to "0" or set SIOINH to "1" in buffer empty interrupt service program to end transmitting. That the transmission has ended can be determined from the status of SIOF (bit 7 in SIOSR) because SIOF is cleared to "0" when a transfer is completed. When SIOINH is set, the transmission is immediately ended and SIOF is cleared to "0".
When an external clock is used, it is also necessary to clear SIOS to "0" before shifting the next data; otherwise, dummy data will be transmitted and the operation will end.
If the number of words is to be changed, SIOS is cleared to "0". After confirmed that SIOF has been cleared to "0", BUF must be rewritten.

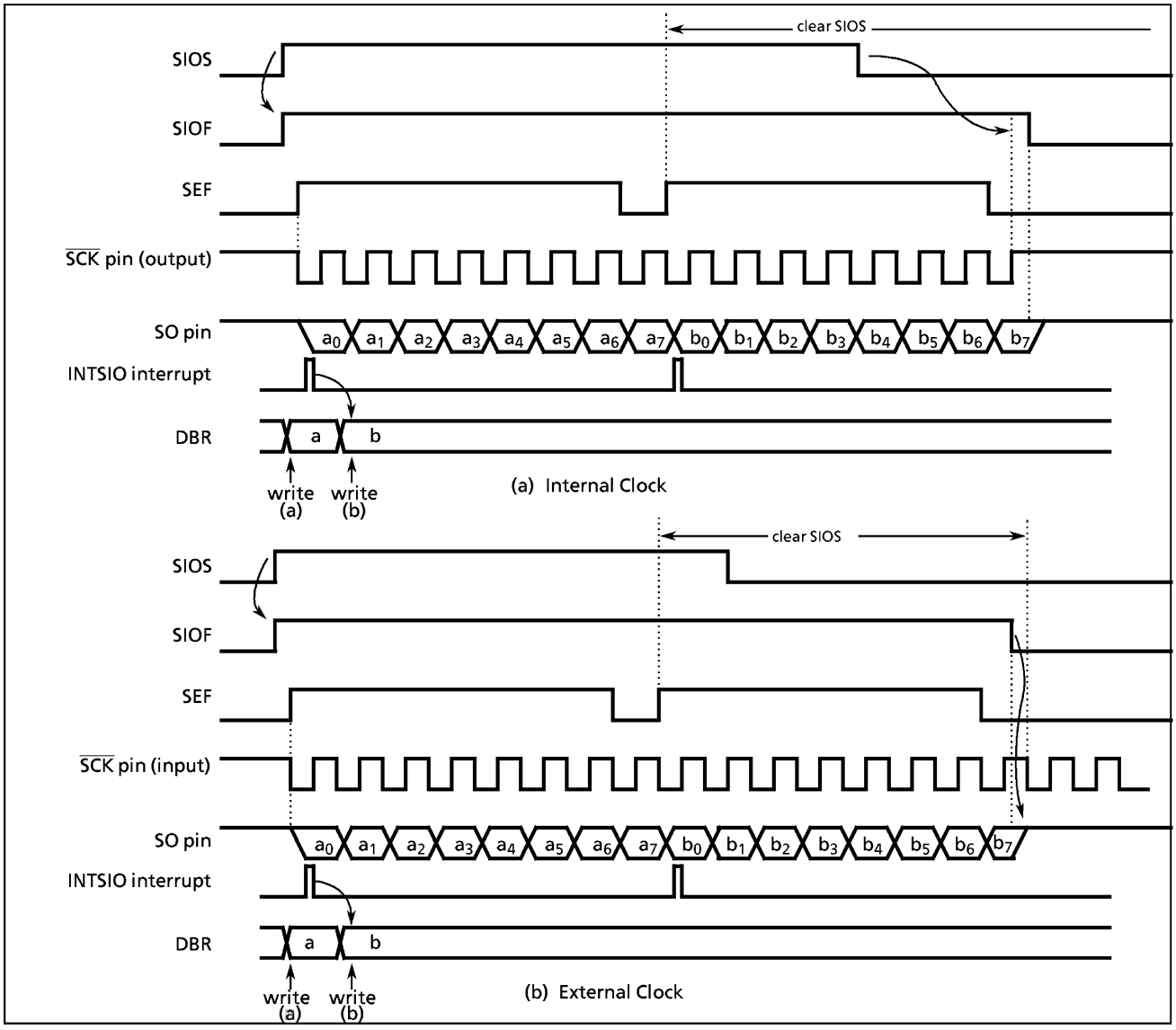


Figure 2-41. Transfer Mode (Example: 8-bit, 1 Word Transfer)

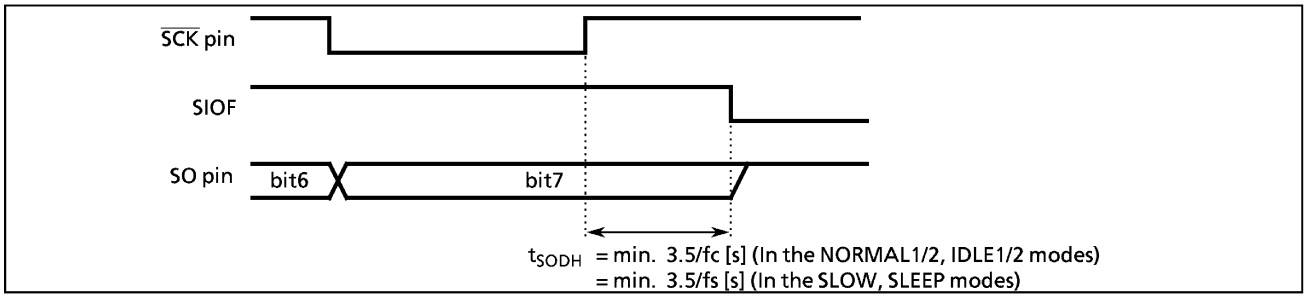


Figure 2-42. Transmitted Data Hold Time at End of Transmit

(2) **4-bit and 8-bit Receive** Modes

After setting the control registers to the receive mode, set SIOS to "1" to enable receiving. The data are then transferred to the shift register via the SI pin in synchronous with the serial clock. When one word of data has been received, it is transferred from the shift register to the data buffer register (DBR). When the number of words specified with the BUF has been received, an INTSIO (buffer full) interrupt is generated to request that these data be read out. The data are then read from the data buffer registers by the interrupt service program.

When the internal clock is used, and the previous data are not read from the data buffer register before the next data are received, the serial clock will stop and an automatic-wait will be initiated until the data are read. A wait will not be initiated if even one data word has been read.

Note : Waits are also canceled by reading a DBR not being used as a received data buffer register is read; therefore, during SIO do not use such DBR for other applications.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, the previous data are read before the next data are transferred to the data buffer register. If the previous data have not been read, the next data will not be transferred to the data buffer register and the receiving of any more data will be canceled. When an external clock is used, the maximum transfer speed is determined by the delay between the time when the interrupt request is generated and when the data received have been read. When the transmit is started, after the SIOF goes "1" output from the SO pin holds final bit of the last data until falling edge of the SCK.

The transmission is ended by clearing SIOS to "0" or set SIOINH to "1" in buffer full interrupt service program to end receiving.

When SIOS is cleared, the current data are transferred to the buffer in 4-bit or 8-bit blocks. The receiving mode ends when the transfer is completed. SIOF is cleared to "0" when receiving is ended and thus can be sensed by program to confirm that receiving has ended. When SIOINH is set, the receiving is immediately ended and SIOF is cleared to "0".

If it is necessary to change the number of words in external clock operation, SIOS should be cleared to "0" then BUF must be rewritten after confirming that SIOF has been cleared to "0".

If it is necessary to change the number of words in internal clock operation, during automatic wait which occurs after completion of data receiving. BUF must be rewritten before the received data is read out.

Note : The buffer contents are lost when the transfer mode is switched. If it should become necessary to switch the transfer mode, end receiving by clearing SIOS to "0", read the last data and then switch the transfer mode.

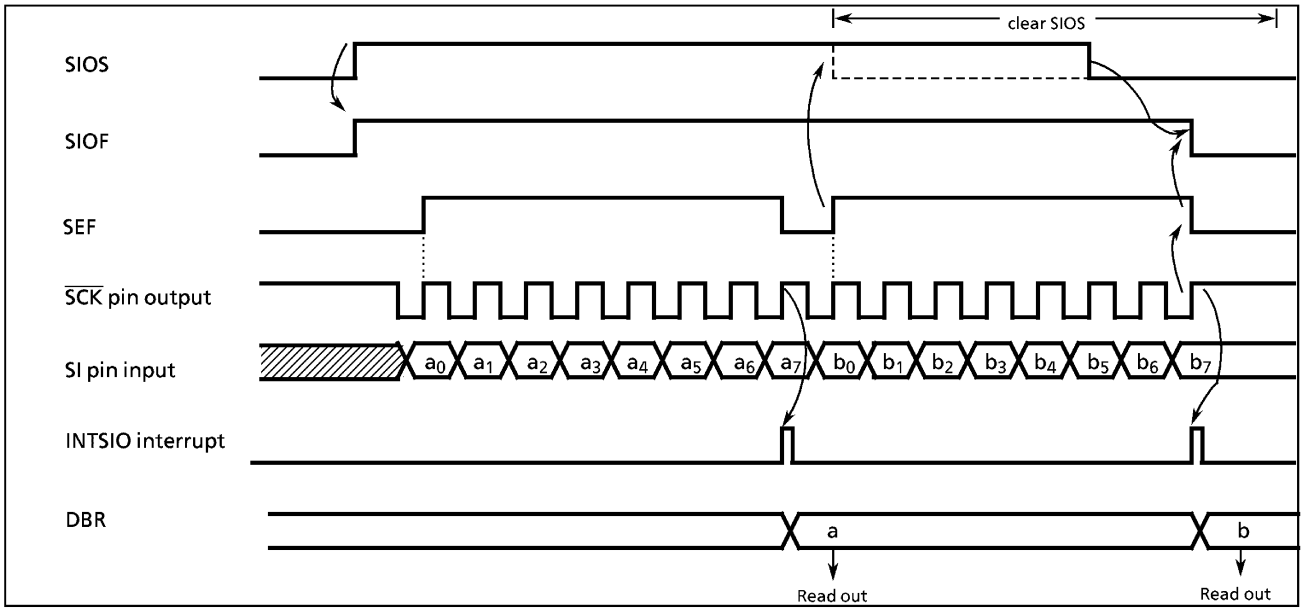


Figure 2-43. Receive Mode (Example : 8-bit, 1 Word, Internal Clock)

(3) **8-bit Transmit/Receive** Mode

After setting the control registers to the 8-bit transmit/receive mode, write the data to be transmitted first to the data buffer registers (DBR). After that, enable transceiving by setting SIOS to "1". When transmitting, the data are output from the SO pin at leading edges of the serial clock. When receiving, the data are input to the SI pin at the trailing edges of the serial clock. 8-bit data are transferred from the shift register to the data buffer register. An INTSIO interrupt is generated when the number of data words specified with the BUF has been transferred. The interrupt service program reads the received data from the data buffer register and then writes the data to be transmitted. The data buffer register is used for both transmitting and receiving; therefore, always write the data to be transmitted after reading the received data.

When the internal clock is used, a wait is initiated until the received data are read and the next data are written. A wait will not be initiated if even one data word has been written.

When an external clock is used, the shift operation is synchronized with the external clock; therefore, it is necessary to read the received data and write the data to be transmitted next before starting the next shift operation. When an external clock is used, the transfer speed is determined by the maximum delay between generation of an interrupt request and the received data are read and the data to be transmitted next are written.

The transmit is ended by clearing SIOS to "0" or set SIOINH to "1" in interrupt service program. When SIOS is cleared, the current data are transferred to the data buffer register in 8-bit blocks. The transmit mode ends when the transfer is completed. SIOF is cleared to "0" when receiving is ended and thus can be sensed by program to confirm that receiving has ended. When SIOINH is set, the transmission is immediately ended and SIOF is cleared to "0".

If the number of words is to be changed during transfer, SIOS must be cleared to "0" and BUF is rewritten after SIOF is determined to be cleared to "0" during automatic-wait operation of an external clock operation. The number of words can be changed in an internal clock. In this case BUF must be rewritten before the received data is read out.

Note : Waits are also canceled by writing to a DBR not being used as the transmit/received data buffer registers; therefore, do not use such DBR for other applications.

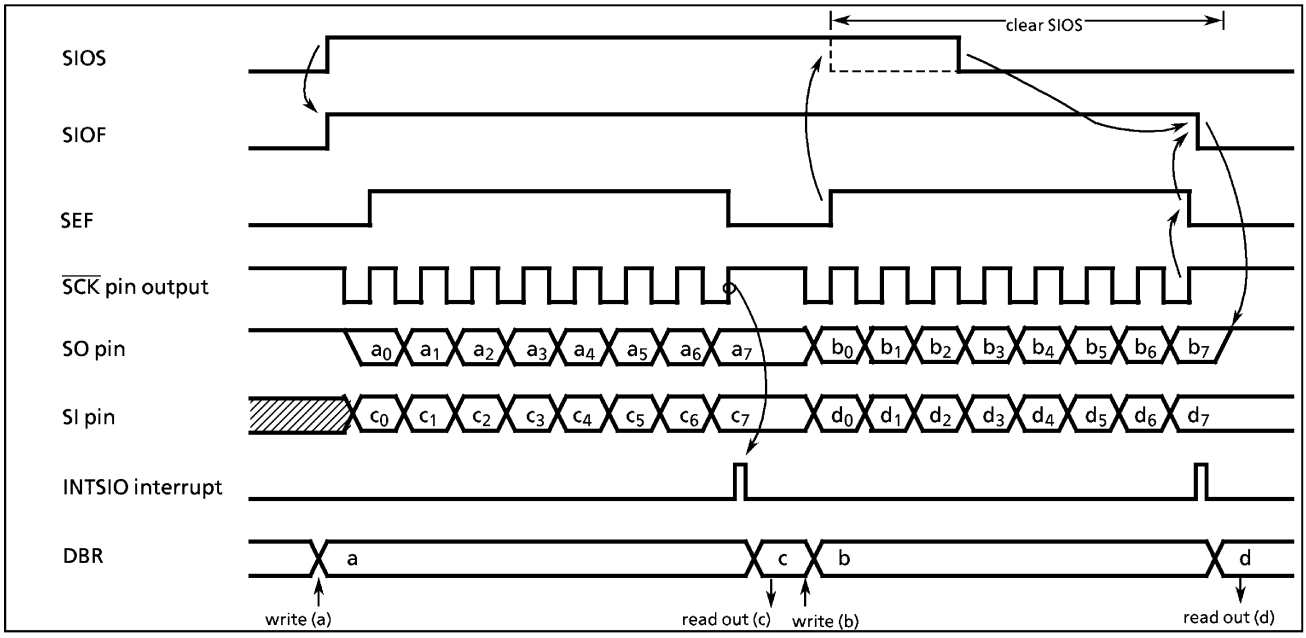


Figure 2-44. Transmit/Receive Mode (Example : 8-bit, 1word, internal clock)

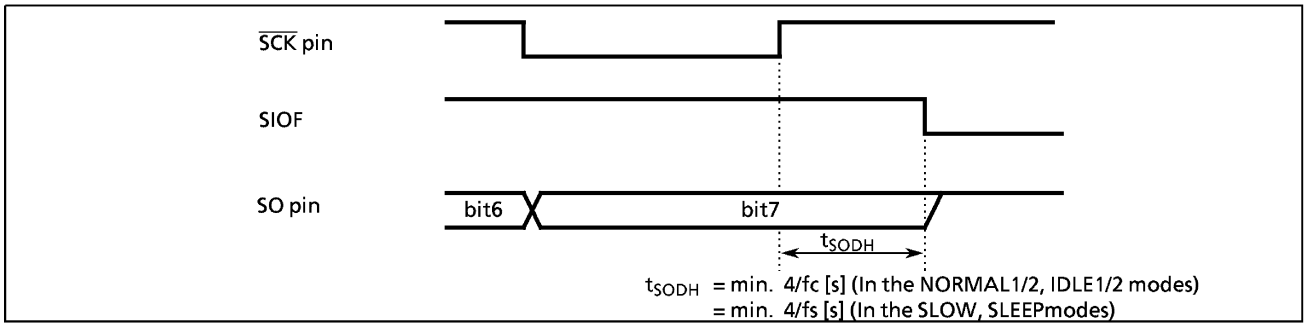


Figure 2-45. Transmitted Data Hold Time at End of Transmit/receive

2.10 8-bit A/D Converter (ADC)

The 87C814/H14/K14/M14 each have an 8-channel multiplexed-input 8-bit successive approximate type A/D converter with sample and hold.

2.10.1 Configuration

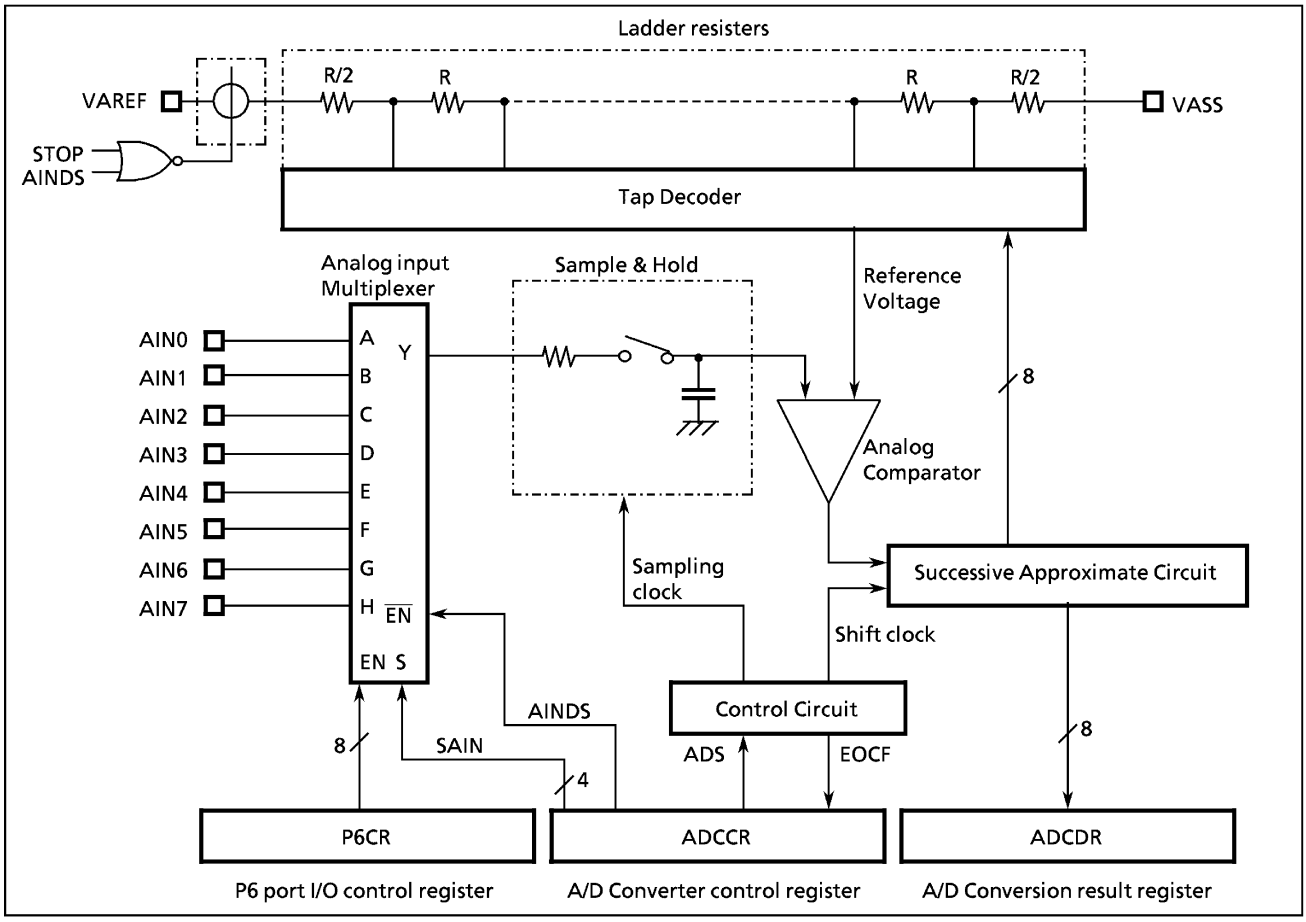


Figure 2-46. A/D Converter

2.10.2 Control

The A/D converter is controlled by an A/D converter control register (ADCCR).

A/D Conversion Result Register

	7	6	5	4	3	2	1	0	
ADCDR ($000F_H$)									Read only

A/D Converter Control Register

ADCCR ($000E_H$)

7	6	5	4	3	2	1	0
EOCF	ADS		AINDS	SAIN			

(Initial value : 00*0 0000)

SAIN	Analog input selection	0000 : AIN0 0001 : AIN1 0010 : AIN2 0011 : AIN3 0100 : AIN4 0101 : AIN5 0110 : AIN6 0111 : AIN7 1 *** : reserved	R/W
AINDS	Analog input control	0 : Enable 1 : Disable	
ADS	A/D conversion start	0 : – 1 : A/D conversion start	
EOCF	End of A/D conversion flag	0 : Under conversion or Before conversion 1 : End of conversion	R

*Note 1 : * ; don't care*

Note 2 : Select analog input when A/D converter stops.

Note 3 : The ADS is automatically cleared to "0" after starting conversion.

Note 4 : The EOCF is cleared to "0" when reading the ADCDR.

Note 5 : The EOCF is read-only.

Figure 2-47. A/D Converter Control Register and A/D Conversion Result Register

2.10.3 Operation

Apply analog reference voltage to pins VAREF and VASS.

(1) Start of A/D conversion

First, set the corresponding P6CR bit to "0" for analog input.
Clear the AINDS (bit 4 in ADCCR) to "0" and select one of eight analog input AIN7-AIN0 with the SAIN (bits 3-0 in ADCCR).
A/D conversion is started by clearing the ADS (bit 6 in ADCCR) to "0".
Conversion is accomplished in 46 machine cycles (184/fc [s]).
The EOCF (bit 7 in ADCCR) is set to "1" at end of conversion.

Note : The pin that is not used as an analog input can be used as regular input/output pins. During conversion, do not perform output instruction to maintain a precision for all of the pins.

(2) Reading of A/D conversion result

After the end of conversion, read the conversion result from the ADCDR.
The EOCF is automatically cleared to "0" when reading the ADCDR.

(3) A/D conversion in STOP mode

When the MCU places in the STOP mode during the A/D conversion, the conversion is terminated and the ADCDR contents become indefinite.

However, if the STOP mode is started after the end of conversion (EOCF = 1), the ADCDR contents are held.

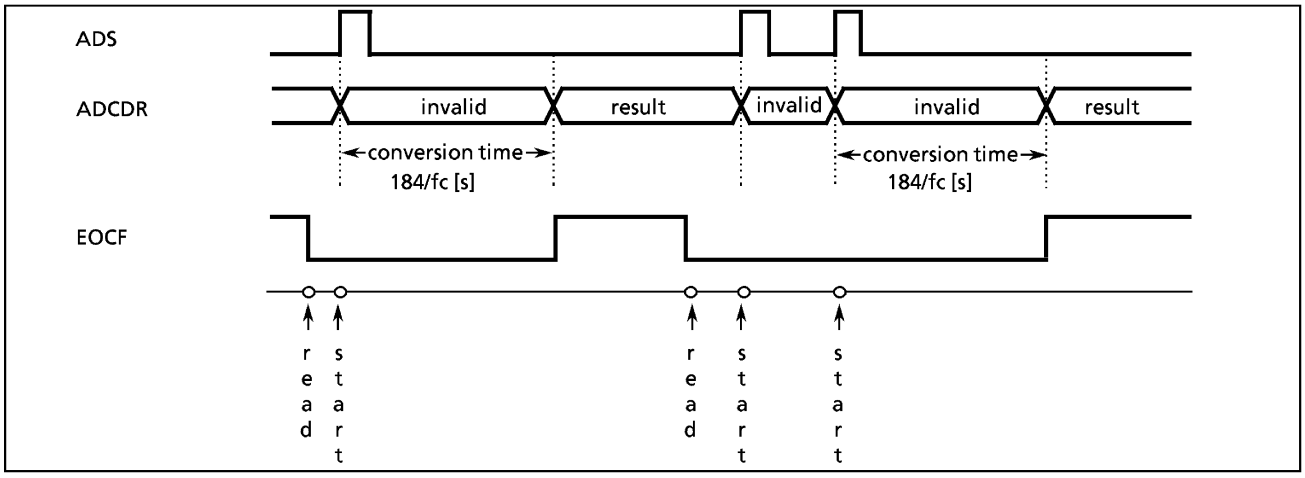


Figure 2-48. A/D Conversion Timing Chart

Example:

```
                ; AIN SELECT
                LD          (ADCCR) , 00000100B     ;   selects AIN4
                ; A/D CONVERT START
                SET         (ADCCR) . 6             ;   ADS = 1
SLOOP    :      TEST        (ADCCR) . 7             ;   EOCF = 1 ?
                JRS         T, SLOOP
                ; RESULT DATA READ
                LD          (9EH), (ADCDR)
```

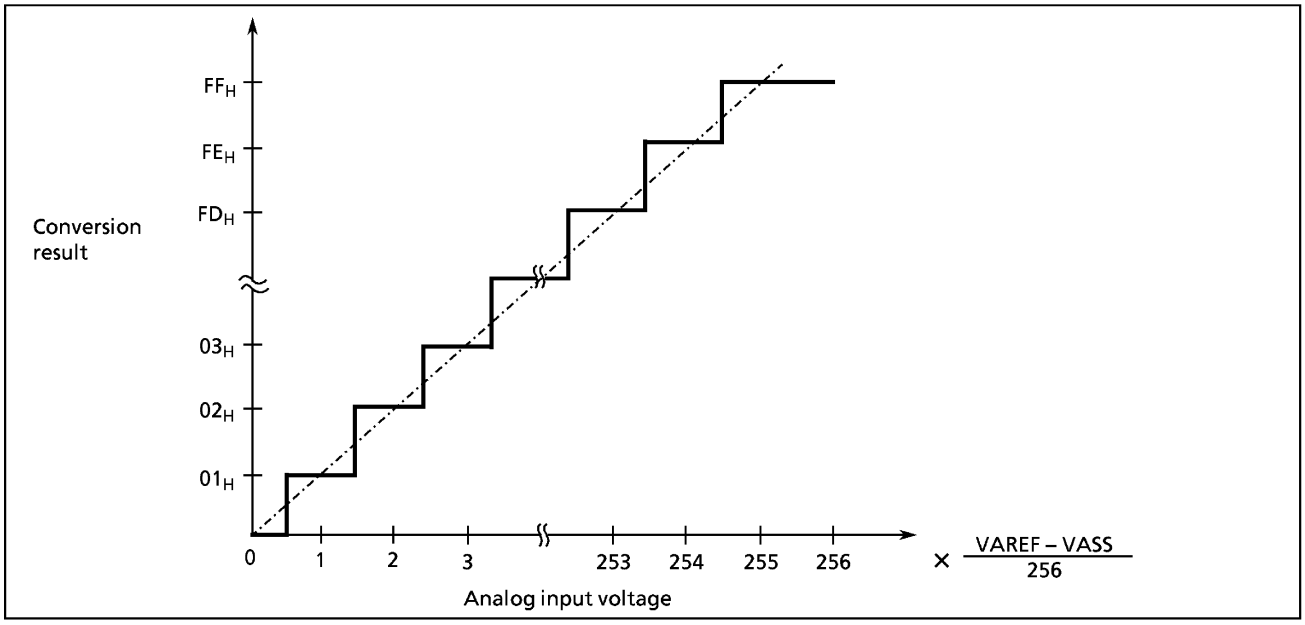


Figure 2-49. Analog Input Voltage vs A/D Conversion Result (typ.)

2.11 Pulse Width Modulation Circuit Output

87C814/H14/K14/M14 have14 built-in pulse width modulation (PWM) channel. D/A converter output can easily be obtained by connecting an external low-pass filter. PWM output is multiplexed with general purpose I/O port as; P33 ($\overline{PWM14}$). When P33 port is used PWM output, the corresponding bit output latch should be set to "1".

2.11.1 Configuration

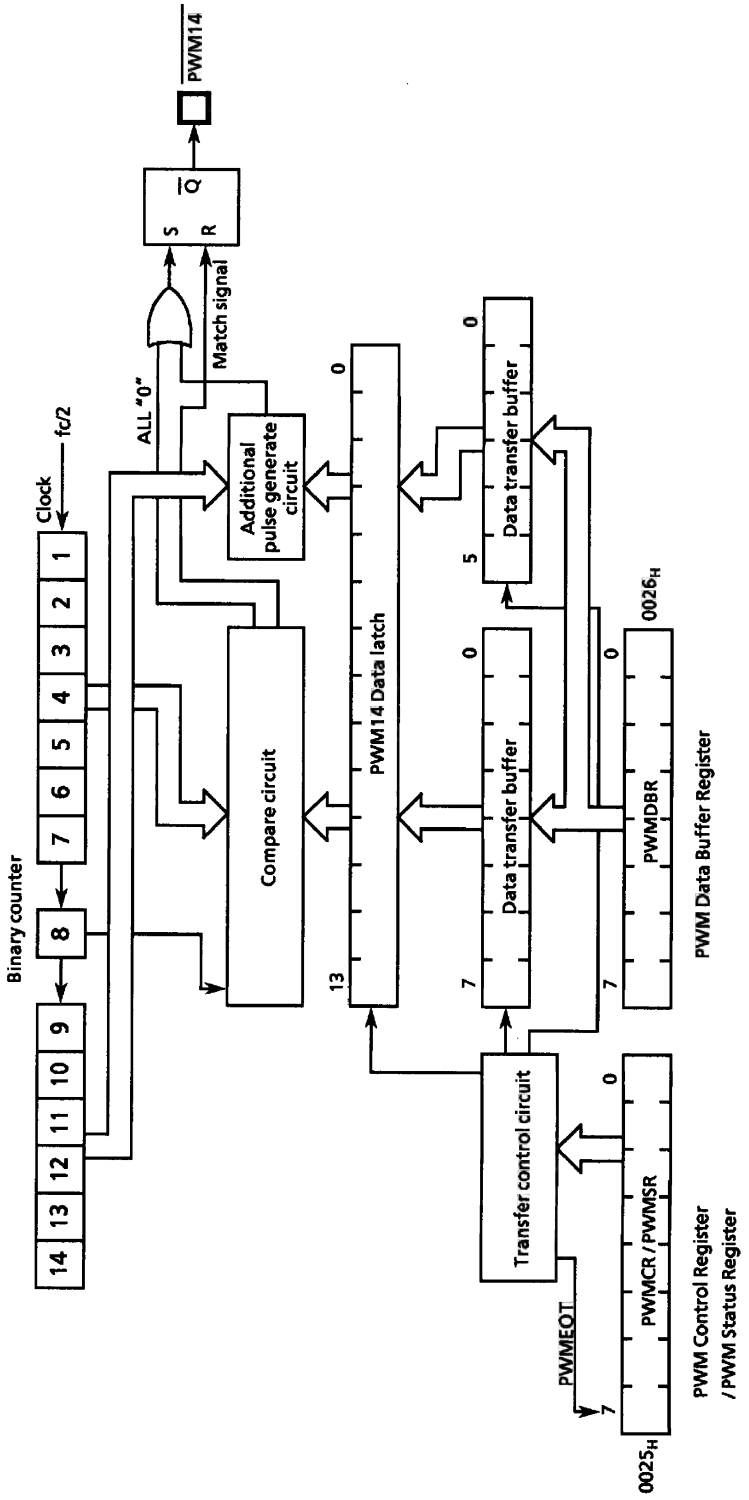


Figure 2-50. Pulse Width Modulation Circuit

2.11.2 PWM Output Wave Form

(1) $\overline{PWM14}$ output

This is 14-bit resolution PWM output and one period is $T_M = 2^{15}/fc$ [s].

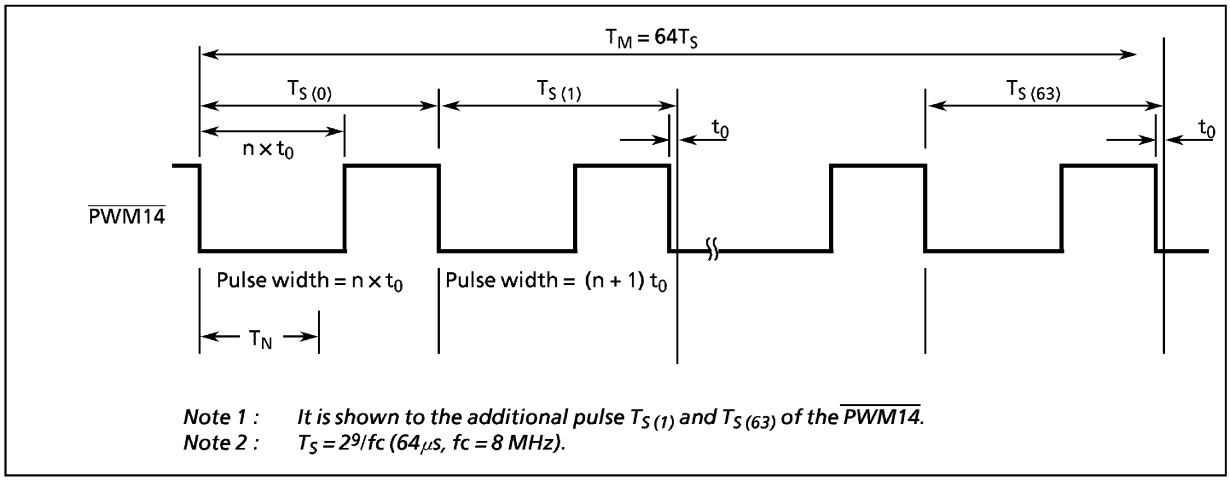
The 8 high-order bits of the PWM data latch control the pulse width of the pulse output with a period of T_S ($T_S = T_M/64$). When the 8-bit data are decimal n ($0 \leqq n \leqq 255$), this pulse width becomes $n \times t_0$, where $t_0 = 2/fc$.

The lower 6-bit of 14 bit data are used to control the generation of additional to wide pulse in each T_S period. When the 6-bit data are decimal m ($0 \leqq m \leqq 63$), the additional pulse is generated in each of m periods out of 64 periods contained in a T_M period. The relationship between the 6 bits data and the position of TS period where the additional pulse is generated is shown in Table 2-10.

Table 2-10. Correspondence between 6 Bits Data and the Additional Pulse Generated TS Period

Bit position of 6 bits data	Relative position of Ts where the output pulse is generated. (No. i of $T_{S(i)}$ is listed)
Bit 0	32
Bit 1	16, 48
Bit 2	8, 24, 40, 56
Bit 3	4, 12, 20, 28, 36, 44, 52, 60
Bit 4	2, 6, 10, 14, 18, 22, 26, 30, ·····, 58, 62
Bit 5	1, 3, 5, 7, 9, 11, 13, 15, 17, ·····, 59, 61, 63

Note : When the corresponding bit is "1", it is output.



Note 1 : It is shown to the additional pulse $T_{S(1)}$ and $T_{S(63)}$ of the $\overline{PWM14}$.
Note 2 : $T_S = 2^9/fc$ (64 μs, fc = 8 MHz).

Figure 2-51. PWM Output Wave Form

2.11.3 Control

PWM output is controlled by PWM Control Register (PWMCR) and PWM Data Buffer Register (PWMDBR). The status of transfer PWM data from PWMDBR to PWM data latch is read by PWMEOT of PWM status register (PWMSR).

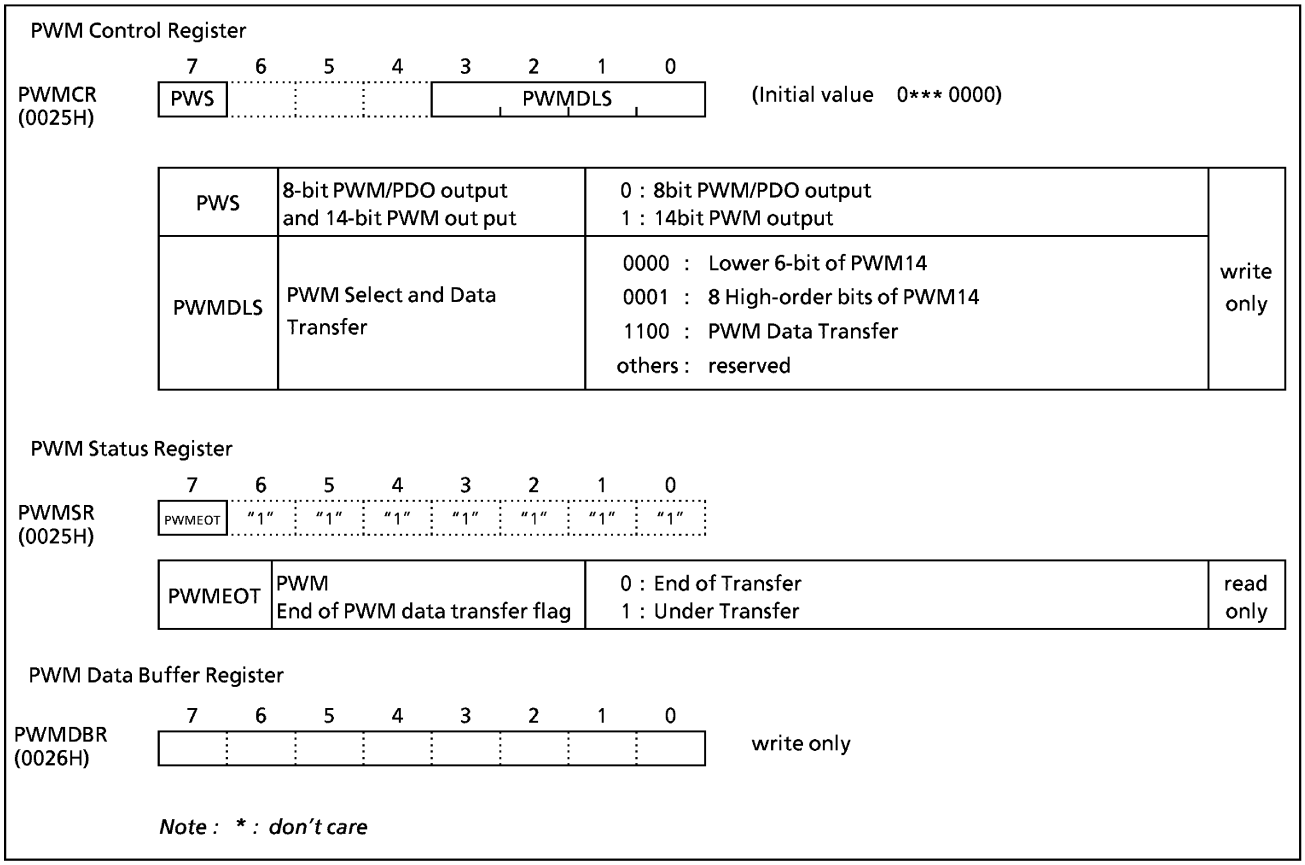


Figure 2-52. PWM Control Register / PWM Status Register / PWM Data Buffer Register

(1) Programing of PWM Data

PWM output is controlled by PWM writing the output data to data latches.
For the writing the output data are divided using the PWM Control Register (PWMCR).

1. Select lower 6-bit of the data latch by PWMDLS.
2. Write PWM output data to the PWMDBR.
3. Select 8 high-order bits of the data latch by PWMDLS.
4. Write PWM output data to the PWMDBR.
5. Write "8CH" to the PWMCR.
   When switching of the output data is completed, the end of PWM data transfer flag becomes "0", indicating that the next data can be written. Do not write PWM data when the end of PWM data is "1" because write errors can occur in this case.

Note : When writing the output data to PWM, write "8CH" to the PWMDLS after writing of the 14-bits output data is completed.

While the output data are being written to the data latch, the previously written data are being output. The maximum time from the point at which "8CH" is written to the data latch until PWM output is switched is 2^{15}/fc (4.096 ms, at fc = 8 MHz).

Example : $\overline{\text{PWM14}}$ pin outputs 32 μs pulse width without the additional pulse.
Note : at fc = 8 MHz

```
        LD      (PWMCR), 80H      ; Select lower 6-bit of PWM14
        LD      (PWMDBR), 00H     ; Without the additional pulse
        LD      (PWMCR), 81H      ; Select 8 high-order bits of PWM14
        LD      (PWMDBR), 80H     ; 32 μs ÷ 2/fc = 80H
        LD      (PWMCR), 8CH      ; PWM Data Transfer
WAIT0 : TEST    (PWMSR). 7        ; PWMEOT = 0?
        JRS     F, WAIT0
```

2.12 Vacuum Fluorescent Tube (VFT) Driver Circuit

The 87C814/H14/K14/M14 features built-in high-breakdown voltage output buffers for directly driving fluorescent tubes, and a display control circuit used to automatically transfer display data to the output port.

2.12.1 Functions

(1) 24 high-breakdown voltage output buffers built-in
- Digit output : 8 to 16 (G0 to G15)
- Segment output : 8 to 16 (S0 to S15)

S0 to S7 and G0 to G7 can be selected by program (in units of bits).
The VKK pin is provided for VFT drive power supply.

(2) Dynamic display, 8 to 16 segment x 1 to 16 digits, can be selected by program.
(3) Pins not used for the VFT driver can be used as general-purpose ports.
(4) Display data (32 bytes in DBR) are automatically transferred to the segment ports.
(5) Brightness level can be adjusted in eight steps using the dimmer function.
(6) Four types (fc/2^{12} to fc/2^{9}) of digit times (duty) can be selected.

2.12.2 Configuration

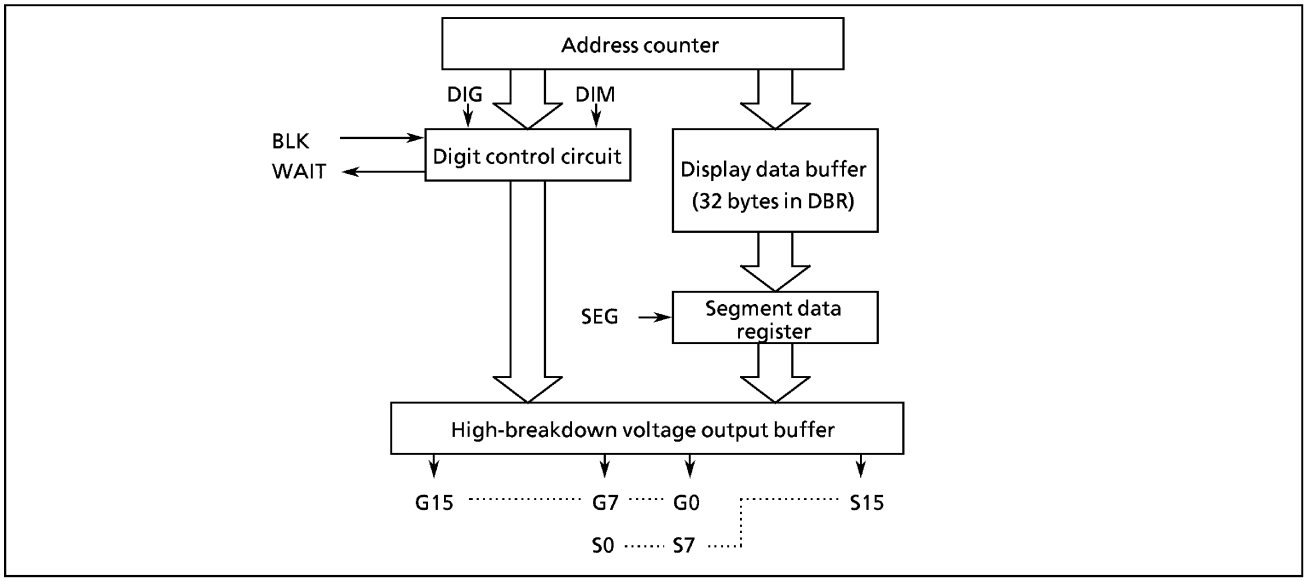


Figure 2-53. VFT

2.12.3 Control

The VFT driver circuit is controlled by the VFT control registers (VFTCR1, VFTCR2 and VFTCR3. Reading VFTSR determines the VFT operating status.
Switching the mode from NORMAL1/2 to SLOW or STOP puts the VFT driver circuit into blanking state (BLK is set to "1" and EXEY is cleared to "0" ; values set in the VFT control registers except BLK and EKEY are maintained), and sets segment outputs and digit outputs are cleared to "0". Thus, ports P5, P7 and P8 function as general-purpose ports.

VFT control register 1

VFTCR1 (0029_H)

7	6	5	4	3	2	1	0
BLK				DIM		SDT	

(initial value: 1**0 0000)

BLK	VFT display control	0 : Display enable 1 : Disable	write only
DIM	Dimmer time select	000 : (14/16) × tseg[s] 001 : (12/16) × tseg[s] 010 : (10/16) × tseg[s] 011 : (8/16) × tseg[s] 100 : (6/16) × tseg[s] 101 : (4/16) × tseg[s] 110 : (2/16) × tseg[s] 111 : (1/16) × tseg[s]	
SDT	Digit time (tseg) select	00 : 2^9 / fc [s] 01 : 2^{10} / fc [s] 10 : 2^{11} / fc [s] 11 : 2^{12} / fc [s]	

Note 1 : fc ; high frequency clock
*Note 2 : * do'nt care*
Note 3 : VFTCR1 is a write-only registers which cannot access any of in read-modify-write instruction such as bit operate, etc.

VFT control register 2

VFTCR2 ($002A_H$)

7	6	5	4	3	2	1	0
DIGH				DIGL			

(initial value: 0000 0000)

DIGH	Number of digits select (on high side) GX (low side) is set by DIGL	0000 : Outputs G0 0001 : Outputs GX to G1 0010 : Outputs GX to G2 0011 : Outputs GX to G3 0100 : Outputs GX to G4 0101 : Outputs GX to G5 0110 : Outputs GX to G6 0111 : Outputs GX to G7 1000 : Outputs GX to G8 1001 : Outputs GX to G9 1010 : Outputs GX to G10 1011 : Outputs GX to G11 1100 : Outputs GX to G12 1101 : Outputs GX to G13 1110 : Outputs GX to G14 1111 : Outputs GX to G15	write only
DIGL	Number of digits select (on low side)	0000 : G0 0001 : G1 0010 : G2 0011 : G3 0100 : G4 0101 : G5 0110 : G6 0111 : G7 1000 : G8 1001 to 1111 : reserved	

Note 1 : If ports used for both digits and segments are used for segments and not for digits.
Note 2 : Even when G0 only is used, DIGH and DIGL must be specified.
Note 3 : VFTCR2 is a write-only registers which cannot access any of in read-modify-write instruction such as bit operate, etc.

Figure 2-54. VFT Control Register 1, 2

DIG/SEG Select

Port P7 is an 8-bit segment/digit output port which can be configured as either a segment or a digit in one-bit until under VFT control register (VFTCR3) control.

Port P7 is configured as a segment output if its corresponding VFTCR3 bit is cleared to "0", and as a digit output if it corresponding VFTCR3 bit is set to "1".

VFT Control Register

VFTCR3 ($002B_H$)

7	6	5	4	3	2	1	0
G0	G1	G2	G3	G4	G5	G6	G7
S7	S6	S5	S4	S3	S2	S1	S0

(Initial value 0000 0000)

Figure 2-55. VFT Control Register 3

VFT Status Register

VFTSR (0029_H)

7	6	5	4	3	2	1	0
WAIT							

WAIT	display operational status momiter	0 : VFT display in operation 1 : VFT display operation disabled	read only

Figure 2-56. VFT Status Register

(1) Display mode setting

Select digit time using the VFT control register 1 (VFTCR1) and the number of digits and the SEG/DIG port using VFT control register 2 (VFTCR2) and number of segments port using port when BLK in VFTCR1 is 1.

Select dimmer time (digit output time) with DIM in VFTCR1.

(2) Display data setting

Data are converted into VFT display data by instructions. The converted data stored in the display data buffer (addresses 0F80 to 0F9F in DBR) are automatically transferred to the VFT driver circuit, then transferred to the high-breakdown voltage output buffer. Thus, to change the display pattern, just change the data in the display data buffer.

Bits in the VFT segment (dot) and display data area correspond one to one. When data are set to 1, the segments corresponding to the bits light. The display data buffer is assigned to the DBR area shown in Figure 2-57. (The display data buffer can not be used as data memory) .

bit 0 1 2 3 4 5 6 7	0 1 2 3 4 5 6 7	digit
0F80	0F81	G0
0F82	0F83	G1
0F84	0F85	G2
0F86	0F87	G3
0F88	0F89	G4
0F8A	0F8B	G5
0F8C	0F8D	G6
0F8E	0F8F	G7
0F90	0F91	G8
0F92	0F93	G9
0F94	0F95	G10
0F96	0F97	G11
0F98	0F99	G12
0F9A	0F9B	G13
0F9C	0F9D	G14
0F9E	0F9F	G15
segment S0 — S7	S8 — S15	

Figure 2-57. VFT Display Data Buffer Memory (DBR)

2.12.4 Display Operation

Clearing BLK in VFTCR1 to 0 after setting the display mode and storing display data starts VFT display. Figures 2-58 and 2-59 show the VFT drive waveforms.

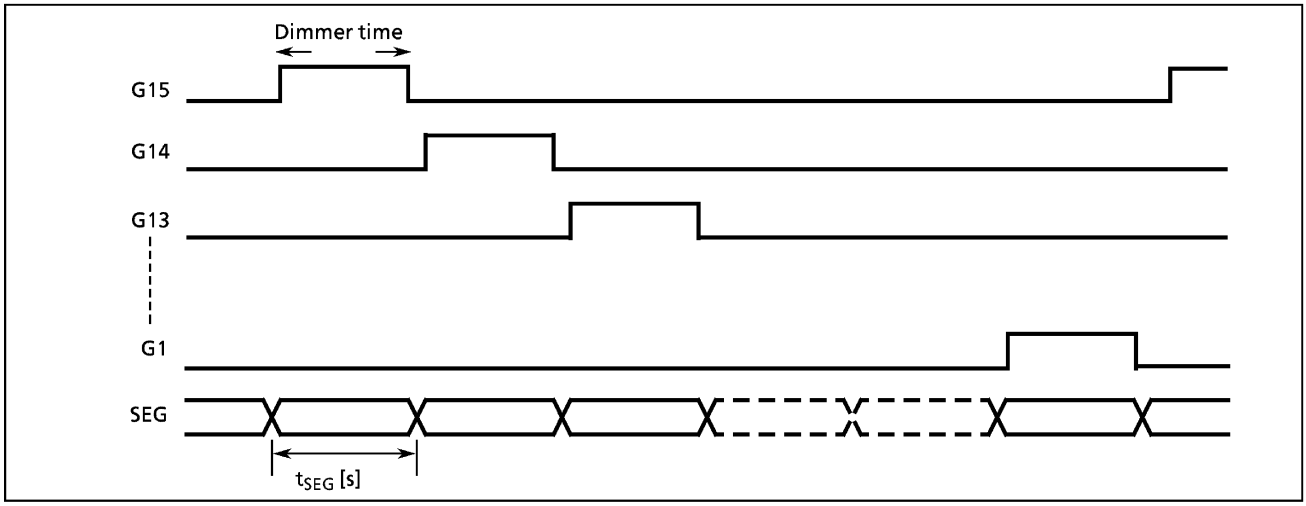


Figure 2-58. VFT Drive Waveform (with 9 Segments and 15 Digits)

Digit cycles change depending on the number of digits set. (Example: with Gn to Gm)

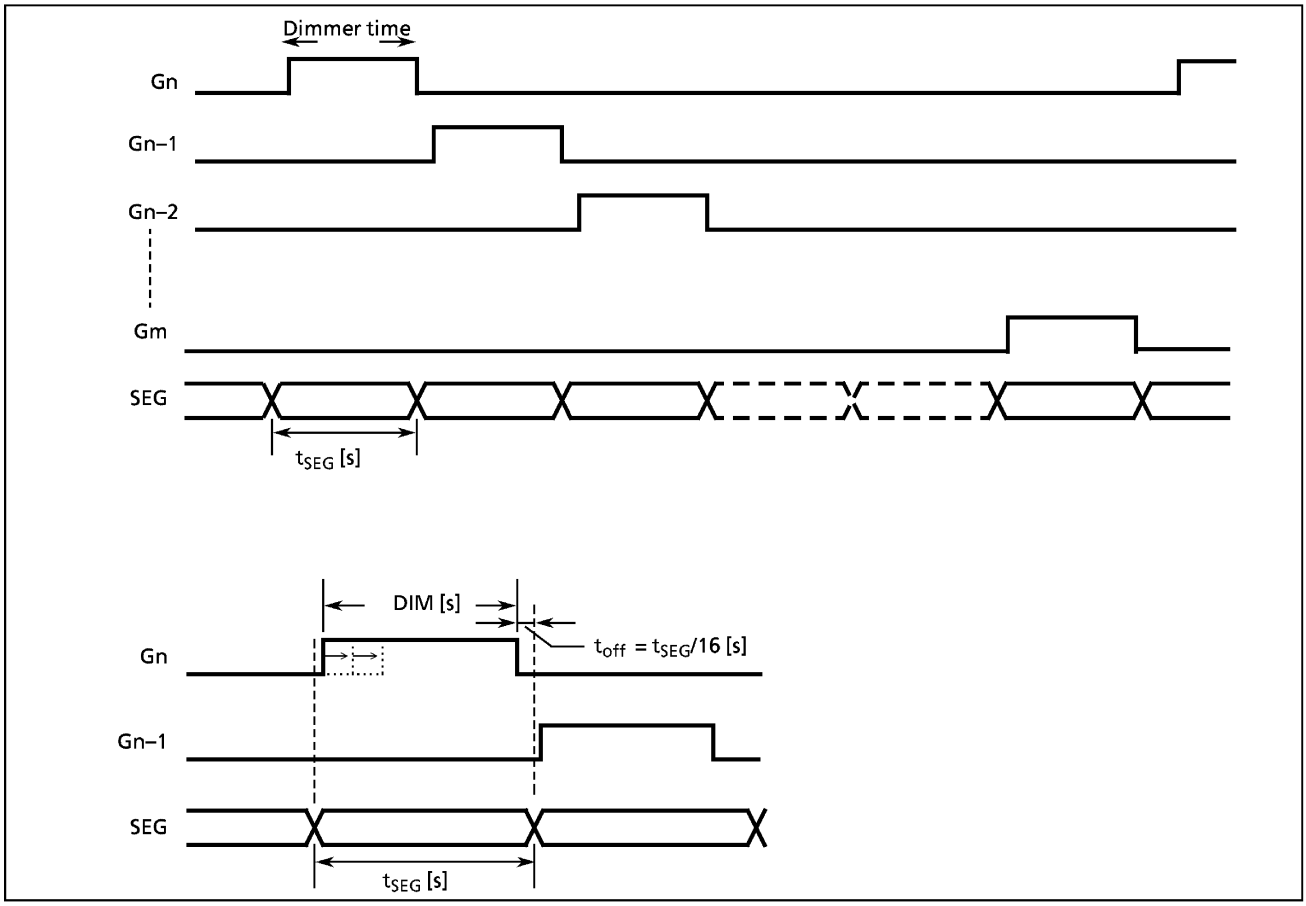


Figure 2-59. VFT Drive Waveform (with XX Segment and Gn to Gm) and Switchover Time

2.12.5 Port Function

(1) High-breakdown voltage buffer

To drive fluorescent display tube, clears the port output latch to 0. The port output latch is initialized to 0 at reset.

It is recommended that ports P5, P7 and P8 should be used as VFT driver output. Precavtion for using as general-purpose I/O pins are follows.

① P7, P8 ports

When ports P7, P8 are used as general purpose I/O pins, the data buffer memory (DBR) which correspond to the pins as also used as segments shoud be clead to "0".

When that pins are pulled down to the VKK pin internally (Rk = 80 kΩ typ. Pins P87 to P84 can be connected pull-down resisters by mask option.) which using as general purpose I/O pin, caution is required.

(a) At output:

For low-level output, the port pulled down to the VKK pin becomes VKK pin voltage. Thus, to prevent VKK pin voltage from being applied to the external circuit, clamp using a diode as shown in Figure 2.60 (a).

(b) At input:

For external data input, clear the port output latch to 0.

The input threshold value is the same as those of other general-purpose I/O ports; however, the port is pulled down to the VKK pin. Therefore, use a suffienctly large Rk (typical: 80 kΩ).

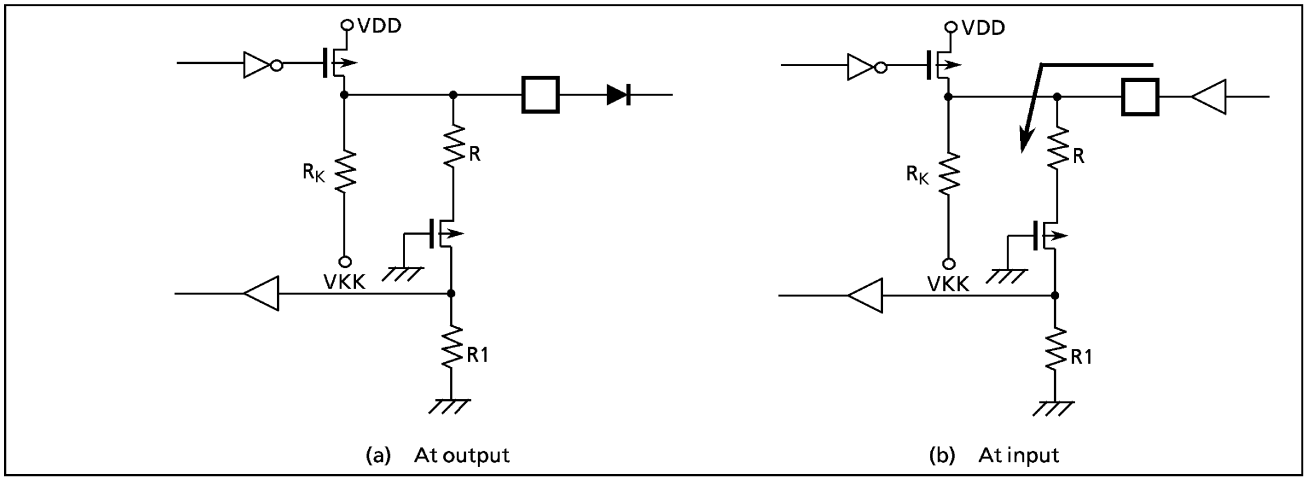


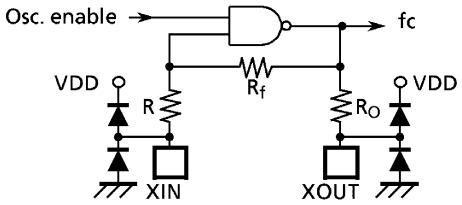
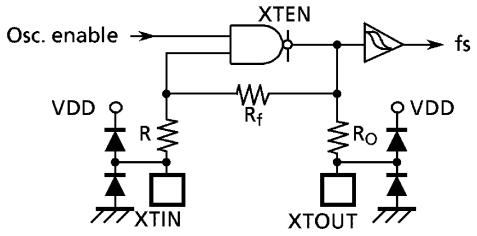
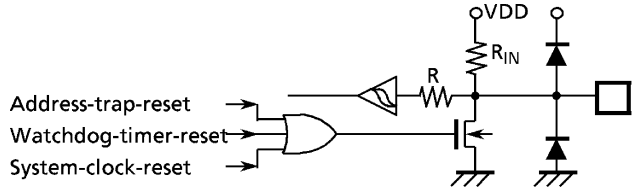
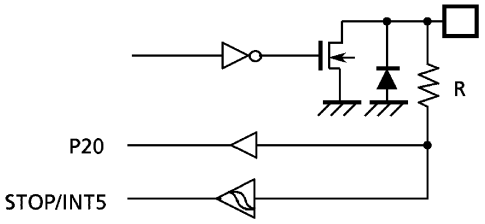
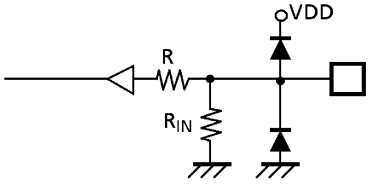
Figure 2-60. External Circuit Interface

INPUT/OUTPUT CIRCUITRY

(1) Control pins

The input/output circuitries of the 87C814/H14/K14/M14 control pins are shown below.

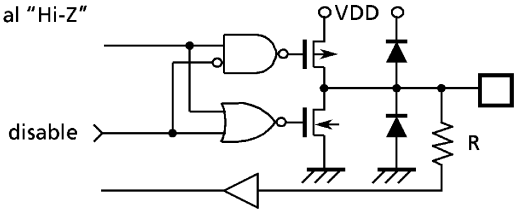
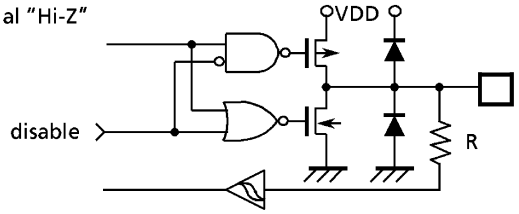
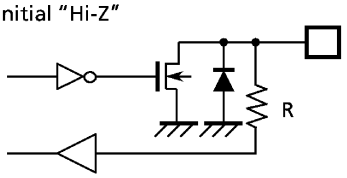
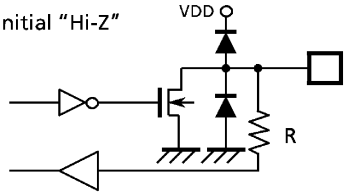
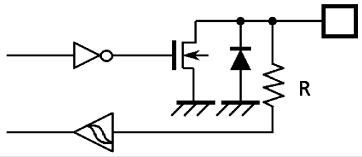
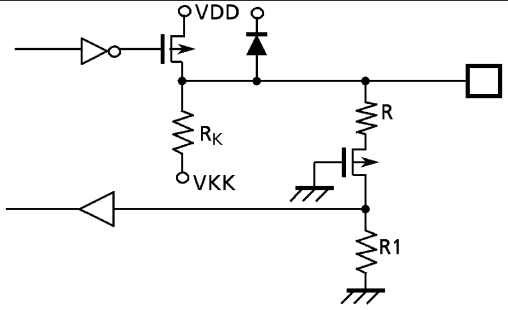
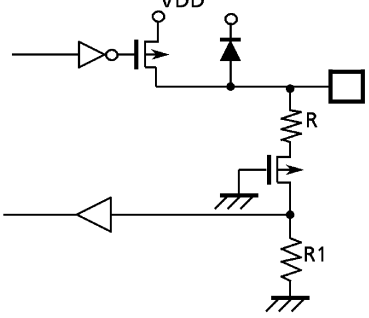
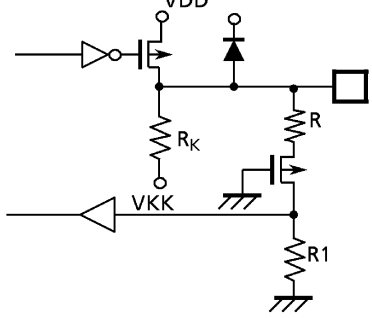
Please specify either the single-clock mode (oscillation only XIN/XOUT) or the dual-clock mode (oscillation both XIN/XOUT and XTIN/XTOUT) by a code (NM1 or NM2) as an option for an operating mode during reset.

CONTROL PIN	I/O	INPUT/OUTPUT CIRCUITRY and code	REMARKS
XIN XOUT	Input Output		Resonator connecting pins (high-frequency) R_f = 1.2 MΩ (typ.) R_O = 1.5 kΩ (typ.)
XTIN XTOUT	Input Output	NM1: Refer to port P2 NM2: 	Resonator connecting pins (low-frequency) R_f = 6 MΩ (typ.) R_O = 220 kΩ (typ.)
$\overline{\text{RESET}}$	I/O		Sink open drain output Hysteresis input Pull-up resister R_{IN} = 220 kΩ (typ.) R = 1 kΩ (typ.)
$\overline{\text{STOP}}$ / $\overline{\text{INT5}}$	Input		Hysteresis input R = 1 kΩ (typ.)
TEST	Input		Pull-down resister R_{IN} = 70 kΩ (typ.) R = 1 kΩ (typ.)

Note1 : The TEST pin of the 87PM14 does not have a pull-down resister. Be sure to fix the TEST pin to low.
Note2 : The 87PM14 is placed in the single-clock mode during reset, and the input/output circuitries are the code NMI type.

(2) - ① Input/Output Ports

The input/output circuitries of the 87C814/H14/K14/M14 input / output ports are shown below, any one of the circuitries can be chosen by a code (A, B) as a mask option.

PORT	I/O	INPUT / OUTPUT CIRCUITRY	REMARKS
P0 P6	I/O	initial "Hi-Z" 	Tri-state I/O R = 1 kΩ (typ.)
P1	I/O	initial "Hi-Z" 	Tri-state I/O Hysteresis input R = 1 kΩ (typ.)
P2	I/O	P20: initial "Hi-Z"  P21, P22: initial "Hi-Z" 	Sink open drain output R = 1 kΩ
P3	I/O	initial "Hi-Z" 	Sink open drain output Hysteresis input R = 1 kΩ (typ.)
P5 P7 P80 to P83	I/O	initial "Hi-Z" 	Source open drain output High-break down voltage R_K = 80 kΩ (typ.) R = 1 kΩ (typ.) R1 = 200 kΩ (typ.)
P84 to P87	I/O	A: initial "Hi-Z"  B: initial "Hi-Z" 	Source open drain output High-breakdown voltage R_K = 80 kΩ (typ.) R = 1 kΩ (typ.) R1 = 200 kΩ (typ.)

Electrical Characteristics

Absolute Maximum Ratings (V_{SS} = 0 V)

Parameter	Symbol	Pins	Ratings	Unit
Supply Voltage	V_{DD}		– 0.3 to 6.5	V
Input Voltage	V_{IN}		– 0.3 to V_{DD} + 0.3	V
Output Voltage	V_{OUT1}	P0, P1, P2, P3, P6, XOUT, $\overline{RESET}$	– 0.3 to V_{DD} + 0.3	V
	V_{OUT2}	Source open drain ports	V_{DD} – 40 to V_{DD} + 0.3	
Output Current (Per 1 pin)	I_{OUT1}	P0, P1, P2, P3, P6	3.2	mA
	I_{OUT2}	P8	– 12	
	I_{OUT3}	P5, P7 (digit outputs)	– 25	
Output Current (Total)	ΣI_{OUT1}	P0, P1, P2, P3, P6	120	mA
	ΣI_{OUT2}	P5, P7, P8	– 120	
Power Dissipation [Topr = 25°C]	PD		600	mW
Soldering Temperature (time)	Tsld		260 (10 s)	°C
Storage Temperature	Tstg		– 55 to 125	°C
Operating Temperature	Topr		– 30 to 70	°C

Note: The absolute maximum ratings are rated values which must not be exceeded during operation, even for an instant. Any one of the ratings must not be exceeded. If any absolute maximum rating is exceeded, a device may break down or its performance may be degraded, causing it to catch fire or explode resulting in injury to the user. Thus, when designing products which include this device, ensure that no absolute maximum rating value will ever be exceeded.

Recommended Operating Conditions (V_{SS} = 0 V, Topr = – 30 to 70°C)

Parameter	Symbol	Pins	Conditions		Min	Max	Unit
Supply Voltage	V_{DD}		fc = 8 MHz	NORMAL 1, 2 modes	4.5	5.5	V
				IDLE1, 2 modes			
			fs = 32.768 kHz	SLOW mode	2.7		
				SLEEP mode			
				STOP mode	2.0		
Output Voltage	V_{OUT2}	Source open drain ports			V_{DD} – 38	V_{DD}	V
Input High Voltage	V_{IH1}	Except hysteresis input	$V_{DD} \geqq 4.5$ V		$V_{DD} \times 0.70$	V_{DD}	V
	V_{IH2}	Hysteresis input			$V_{DD} \times 0.75$		
	V_{IH3}		$V_{DD} < 4.5$ V		$V_{DD} \times 0.90$		
Input Low Voltage	V_{IL1}	Except hysteresis input	$V_{DD} \geqq 4.5$ V		0	$V_{DD} \times 0.30$	V
	V_{IL2}	Hysteresis input				$V_{DD} \times 0.25$	
	V_{IL3}		$V_{DD} < 4.5$ V			$V_{DD} \times 0.10$	
Clock Frequency	fc	XIN, XOUT	V_{DD} = 4.5 V to 5.5 V		0.4	8.0	MHz
	fs	XTIN, XTOUT			30.0	34.0	kHz

Note 1: The recommended operating conditions for a device are operating conditions under which it can be guaranteed that the device will operate as specified. If the device is used under operating conditions other than the recommended operating conditions (supply voltage, operating temperature range, specified AC/DC values etc.), malfunction may occur. Thus, when designing products which include this device, ensure that the recommended operating conditions for the device are always adhered to.

Note 2: Clock frequency fc: Supply voltage range is specified in NORMAL 1/2 mode and IDLE 1/2 mode.

D.C. Characteristics

(V_{SS} = 0 V, Topr = – 30 to 70°C)

Parameter	Symbol	Pins	Conditions	Min	Typ.	Max	Unit
Hysteresis Voltage	V_{HS}	Hysteresis input		–	0.9	–	V
Input Current	I_{IN1}	TEST	V_{DD} = 5.5 V V_{IN} = 5.5 V / 0 V	–	–	± 2	μA
	I_{IN2}	Open drain ports, Tri-state ports					
	I_{IN3}	$\overline{RESET}$, $\overline{STOP}$					
Input Resistance	R_{IN2}	$\overline{RESET}$		100	220	450	kΩ
Pull-down Resistance	R_1	Source open drain ports	V_{DD} = 5.5 V, V_{OUT} = 5.5 V	–	200	–	
	R_K		V_{DD} = 5.5 V, V_{KK} = – 30 V	–	80	–	
Output Leakage Current	I_{LO1}	Sink open drain ports	V_{DD} = 5.5 V, V_{OUT} = 5.5 V	–	–	2	μA
	I_{LO2}	Source open drain ports	V_{DD} = 5.5 V, V_{OUT} = – 32 V	–	–	– 2	
	I_{LO3}	Tri-state ports	V_{DD} = 5.5 V, V_{OUT} = 5.5 V / 0 V	–	–	± 2	
Output High Voltage	V_{OH2}	Tri-state ports	V_{DD} = 4.5 V, I_{OH} = – 0.7 mA	4.1	–	–	V
	V_{OH3}	P8	V_{DD} = 4.5 V, I_{OH} = – 8 mA	2.4	–	–	
Output Low Voltage	V_{OL}	Except XOUT	V_{DD} = 4.5 V, I_{OL} = 1.6 mA	–	–	0.4	V
Output High current	I_{OH}	P5, P7	V_{DD} = 4.5 V, V_{OH} = 2.4 V	–	– 20	–	mA
Supply Current in NORMAL 1, 2 modes	I_{DD}		VDD = 5.5 V fc = 8 MHz	–	10	16	mA
Supply Current in IDLE 1, 2 modes			fs = 32.768 kHz VIN = 5.3 V / 0.2 V	–	4.5	6	
Supply Current in SLOW mode			V_{DD} = 3.0 V fs = 32.768 kHz	–	30	60	μA
Supply Current in SLEEP mode			V_{IN} = 2.8 V / 0.2 V	–	15	30	
Supply Current in STOP mode			V_{DD} = 5.5 V V_{IN} = 5.3 V / 0.2 V	–	0.5	10	μA

Note 1: Typical values show those at Topr = 25°C , V_{DD} = 5 V.

Note 2: Input Current I_{IN1},I_{IN3} ; The current through resistor is not included, when the input resistor (pull-up/pull-down) is contained.

Note 3: Input Current I_{IN4} ; The current when the pull-down register (Rk) is not connected by the mask option.

A/D Conversion Characteristics

(V_{SS} = 0 V, V_{DD} = 4.5 to 5.5 V, Topr = – 30 to 70°C)

Parameter	Symbol	Conditions	Min	Typ.	Max	Unit
Analog Reference Voltage	V_{AREF}	$V_{AREF} - V_{ASS} \geqq 2.5$ V	V_{DD} – 1.5	–	V_{DD}	V
	V_{ASS}		V_{SS}	–	1.5	
Analog Input Voltage	V_{AIN}		V_{ASS}	–	V_{AREF}	V
Analog Supply Current	I_{REF}	V_{AREF} = 5.5 V, V_{ASS} = 0.0 V	–	0.5	1.0	mA
Nonlinearity Error		V_{DD} = 5.0 V, V_{SS} = 0.0 V V_{AREF} = 5.000 V V_{ASS} = 0.000 V	–	–	± 1	LSB
Zero Point Error			–	–	± 1	
Full Scale Error			–	–	± 1	
Total Error			–	–	± 2	

Note: Total errors includes all errors, except quantization error.

A.C. Characteristics

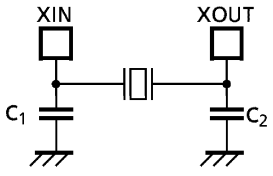
(V_{SS} = 0 V, V_{DD} = 4.5 to 5.5 V, Topr = – 30 to 70°C)

Parameter	Symbol	Conditions	Min	Typ.	Max	Unit
Machine Cycle Time	t_{cy}	In NORMAL1, 2 modes	0.5	–	10	μs
		In IDLE 1, 2 modes				
		In SLOW mode	117.6	–	133.3	
		In SLEEP mode				
High Level Clock Pulse Width	t_{WCH}	For external clock operation (XIN input), fc = 8 MHz	50	–	–	ns
Low Level Clock Pulse Width	t_{WCL}					
High Level Clock Pulse Width	t_{WSH}	For external clock operation (XTIN input), fs = 32.768 kHz	14.7	–	–	μs
Low Level Clock Pulse Width	t_{WSL}					

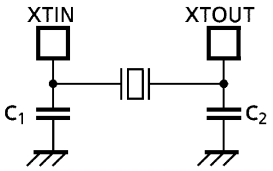
Recommended Oscillating Conditions

(V_{SS} = 0 V, V_{DD} = 4.5 to 5.5 V, Topr = – 30 to 70°C)

Parameter	Oscillator	Oscillation Frequency	Recommended Oscillator		Recommended Constant C_1	Recommended Constant C_2
High-frequency Oscillation	Ceramic Resonator	8 MHz	KYOCERA	KBR8.0M	30pF	30pF
		4 MHz	KYOCERA	KBR4.0MS		
			MURATA	CSA 4.00MG		
	Crystal Oscillator	8 MHz	TOYOCOM	210B 8.0000	20pF	20pF
		4 MHz	TOYOCOM	204B 4.0000		
Low-frequency Oscillation	Crystal Oscillator	32.768 KHz	NDK	MX-38T	15pF	15pF



(1) High-frequency Oscillation



(2) Low-frequency Oscillation

Note: An electrical shield by metal shied plate on the IC package should be recommend able in order to prevent the device from the high electric fieldstress applied for continuous reliable operation.